

AN AUTOMATED HARMONIC ANALYZER FOR USE IN STUDYING THE  
COMMERCIAL POWER SYSTEM

by

NORMAN MORTENSEN

BSEE, Kansas State University, 1986

-----

A MASTER'S THESIS

submitted in partial fulfillment  
of the requirements for the degree

MASTER OF SCIENCE

Department of Electrical and Computer Engineering

KANSAS STATE UNIVERSITY  
Manhattan, Kansas

1988

Approved by:

Gary Johnson  
Major Professor

LD  
21608  
.74  
EECE  
1988  
mlb7  
C. 2

## ACKNOWLEDGEMENTS

A11208 130216

The work detailed in this document was funded by the Kansas Electric Utilities Research Program (KEURP).

I would like to thank my major Professor Dr. G. Johnson, especially for putting up with missed deadlines. Also, I would like to thank Dr. Al Heber and Dr. Anil Pahwa for serving on my graduate committee.

Lastly, I would like to thank my wife Pam for her patience, understanding and, last but not least, her typing efforts.

## TABLE OF CONTENTS

### Chapter 1 : Introductory material

Introduction .....	1-1
A Brief Overview of Power System Harmonics .....	1-2

### Chapter 2 : Harmonic Analyzer Hardware

Introduction .....	2-1
Analog Section .....	2-3
Digital Section	
Signal Description Standards .....	2-8
Timing and Control Signals .....	2-8
Microprocessor and Support .....	2-10
System Memory, EPROM .....	2-13
System Memory, RAM .....	2-15
Serial Ports .....	2-17
PIA, Real Time Clock, ADC .....	2-19

### Chapter 3 : Harmonic Analyzer Software

Introduction .....	3-1
Ardvark .....	3-5
Clock.read .....	3-9
Clock.write .....	3-11
Disp.reg .....	3-11
Disp.string .....	3-14
Download .....	3-16

FFT .....	3-21
Getchar .....	3-34
Get.data .....	3-34
Get.time .....	3-42
Go .....	3-45
Hex.ascii .....	3-47
Hex.check .....	3-47
In .....	3-49
Initialize .....	3-51
Memory .....	3-55
Out .....	3-63
Putchar .....	3-65
Qk.fft .....	3-67
Sample .....	3-70
Send.data .....	3-74
Send.time .....	3-79
Trace .....	3-82

#### Chapter 4 : Testing

Testing Procedures .....	4-1
Results .....	4-4

#### Chapter 5 : Further Considerations and Conclusions

Further Considerations .....	5-1
Conclusions .....	5-3

Appendix A :

A Note on the Harmonic Analyzer Schematics .....	A-1
Parts List .....	A-2
Memory Map .....	A-6

Appendix B :

Harmonic Analyzer Software	
Source Code Listings .....	B-1

Appendix C :

Harmonic Analyzer Interface Program for	
Z-171 Source Code Listings .....	C-1

Appendix D :

Explanation of Signal Flow Graphs .....	D-1
---	-----

References .....	R-1
------------------	-----

## TABLE OF FIGURES

2.1	Block Diagram of Harmonic Analyzer .....	2-2
2.2	Analog Circuitry .....	2-4
2.3	Timing and Control Circuitry .....	2-9
2.4	Microprocessor and Support Circuitry.....	2-12
2.5	System Memory, EPROM .....	2-14
2.6	System Memory, RAM .....	2-16
2.7	Serial Ports and Support Circuitry .....	2-18
2.8	PIA, Real Time Clock and ADC .....	2-20
3.1	The Programming Model for the 68000 .....	3-3
3.2	Flowchart of the Control Program, ARDVARK ...	3-7
3.3	Flowchart for the Routine CLOCK.READ .....	3-10
3.4	Flowchart for the Routine CLOCK.WRITE .....	3-12
3.5	Flowchart for the Routine DISP.REG .....	3-13
3.6	Flowchart for the Routine DISP.STRING .....	3-15
3.7	Flowchart for the Routine DOWNLOAD .....	3-17
3.8	Eight Point DFT Divided Into Two Four Point DFT's .....	3-25
3.9	A Four Point DFT Divided Into Two Two Point DFT's .....	3-26
3.10	A Two Point DFT .....	3-26
3.11	The Relationship Between the Output of the FFT and the DFT .....	3-27
3.12	Signal Flow Graph for Eight Point FFT .....	3-28
3.13	Flowchart for the Routine FFT .....	3-29
3.14	Flowchart for the Routine GETCHAR .....	3-35

3.15	Flowchart for the Routine GET.DATA .....	3-36
3.16	Flowchart for the Routine GET.TIME .....	3-43
3.17	Flowchart for the Routine GO .....	3-46
3.18	Flowchart for the Routine HEX.ASCII .....	3-48
3.19	Flowchart for the Routine HEX.CHECK .....	3-50
3.20	Flowchart for the Routine IN .....	3-52
3.21	Flowchart for the Routine INITIALIZE .....	3-53
3.22	Flowchart for the Routine MEMORY .....	3-57
3.23	Flowchart for the Routine OUT .....	3-61
3.24	Flowchart for the Routine PUTCHAR .....	3-63
3.25	Flowchart for the Routine QK.FFT .....	3-65
3.26	Flowchart for the Routine SAMPLE .....	3-68
3.27	Flowchart for the Routine SEND.DATA .....	3-72
3.28	Data Storage Format .....	3-74
3.29	Flowchart for the Routine SEND.TIME .....	3-77
3.30	Flowchart of the Exception TRACE .....	3-81

## CHAPTER 1 : INTRODUCTORY MATERIAL

### Introduction

The increasing use of sensitive electronic equipment has led to the need for higher quality commercial power. The quality of power can be determined in a number of ways. One method would be to examine a power system for transient surges and other such phenomena. Another technique would be to investigate the steady state spectral purity of an electrical utility's voltages and currents. This latter method is the one used in this study.

In 1985 Dr. Gary Johnson proposed conducting a study of the harmonic content of the Kansas electrical grid to the Kansas Electric Utilities Research Program (KEURP). The problem in conducting this study was the lack of a cost effective harmonic analyzer with the appropriate capabilities on the commercial market. For this reason Dr. Johnson proposed the development of a suitable harmonic analyzer as part of the project. This proposal was accepted by KEURP.

The harmonic analyzer that emerged from this project is a portable self contained unit with all the capabilities needed to monitor power system harmonics. Its user interface has been simplified by making use of current trends in portable personal computers. This harmonic analyzer is described in depth in the remainder of this document.



## A Brief Overview of Power System Harmonics

Ideally, commercial power should possess a frequency spectrum in which all the energy is concentrated at 60Hz. In practice some of the energy is associated with other frequencies. The purpose of this study is to measure this distribution of energy.

If a signal is a periodic waveform then it can be expressed as the sum of sinusoidal waveforms whose frequencies are the frequency of the waveform and its multiples. The frequencies which are multiples of the fundamental frequency of a waveform are called harmonics. Since the waveforms on a commercial power system are generally periodic most of the energy associated with a waveform will be concentrated at the fundamental frequency and its harmonics. The harmonic analyzer described in this document measures the harmonics of the voltage and current waveforms from which the distribution of energy can be calculated.

Harmonics are caused by a number of elements on the system such as generators, transformers, solid state motor controllers and inverters. Due to the fact that the windings of a generator are not perfectly aligned the generated voltage is not a perfect 60Hz sinusoid. Additional harmonics are caused by various nonlinearities of the system. Magnetic

saturation in the generators and transformers accounts for part of this. Other inherently nonlinear devices such as motor speed controllers and inverters also generate harmonics on the power system.

The presence of harmonics on a utility can pose problems to both the utility and the consumer of commercial power. These problems include degraded metering accuracy, interference with the telephone system and the heating of motors and generators. If a circuit is resonant at a specific frequency the magnitude of the voltage and current in the circuit may reach damaging levels. Harmonics can also cause saddles in the speed torque curves of induction machines. Additionally, the wavelength of harmonics also becomes much shorter than the 60Hz wavelength. This may cause the power system to begin to exhibit transmission line effects at harmonic frequencies. These effects of harmonics make it beneficial to both the utility and the consumer of commercial power to have a low harmonic content on the power system.

## CHAPTER 2 : HARMONIC ANALYZER HARDWARE

### Introduction

The harmonic analyzer described in this document was designed to meet a number of criteria needed to make it a practical device for measuring power system harmonics. These include the ability to measure numerous harmonics at once, independent operation, long term recording capability and power failure protection. Other features that would be desirable are portability, low power consumption and ease of use. These requirements point to a special purpose data acquisition system with onboard data processing.

A data acquisition system such as this would require a fast, easy to use processor for signal processing and control. For this reason the harmonic analyzer was built around the Motorola 68000 microprocessor. The 68000 is a high performance, versatile processor that is easy to use. It provides the speed and computing power necessary for signal processing work. The 68000 is also simple to program and ideal for use as an imbedded controller in an independently operating system.

The resulting system provides the features of a data acquisition system coupled with some of the signal processing capabilities of a general purpose computer. Some of

the system's features include:

- \* 64 KBytes of RAM storage for use as scratch pad and data storage.
- \* 16 KBytes of ROM for control program storage.
- \* Real Time Clock to monitor data collection times.
- \* Sampling ports for the current and voltage waveforms designed to minimize the influence on the system.
- \* Data protection in the event of a power failure.
- \* Two serial ports for external communications.

A block diagram of the harmonic analyzer is shown in Figure 2.1. It shows the basic subsystems of the harmonic analyzer.

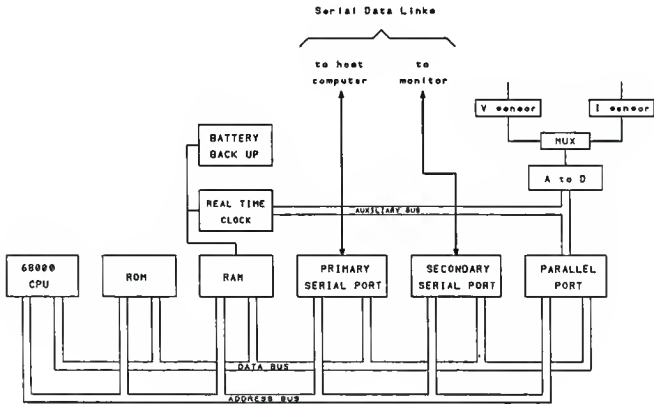


Figure 2.1 : Block diagram showing the basic subsystems of the harmonic analyzer.

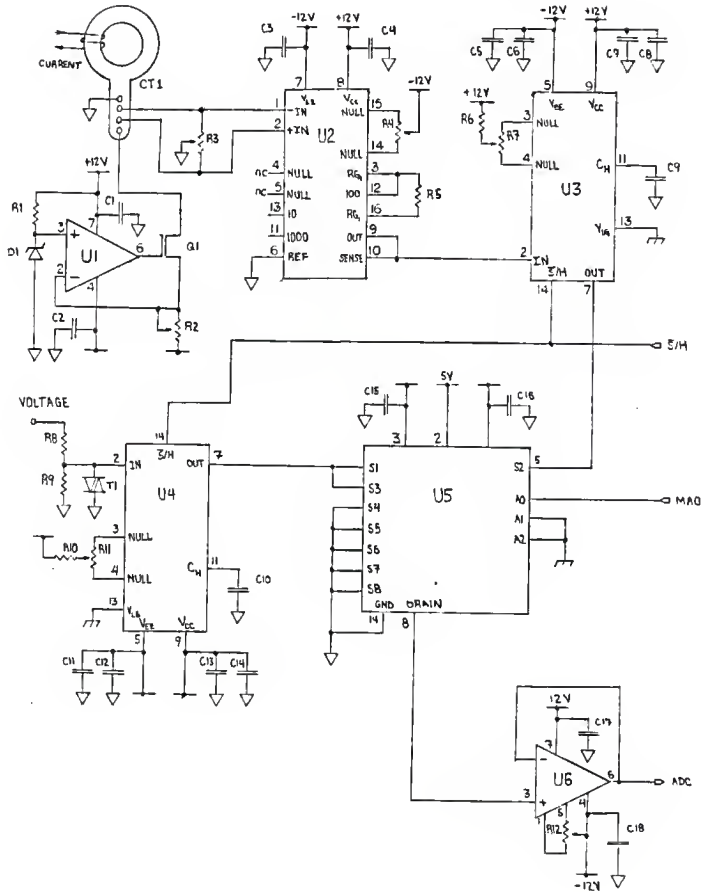
## Analog Section

The analog section of the harmonic analyzer is made up of the voltage and current sensors and related circuitry. The schematics of this portion of the harmonic analyzer are shown in Figure 2.2. The list of parts is given in Appendix A. These sensors convert the line level voltages and currents to analog voltages in the +/- 10 volt range. An analog to digital conversion is then performed on these voltages and the rest of the signal processing is done digitally.

As can be seen from the schematics no antialiasing filter is included to process the signal before it is digitized. This was decided upon for a number of reasons. The amount of energy expected to be found at frequencies that would be aliased was expected to be small based on available information. The aliasing of these frequencies should then cause little error in the measurements. A filter with very steep skirts would have been necessary to properly filter the signal. This would have caused phase distortion in the signals and taken up extra board space. For these reasons the antialiasing filter was omitted.

The current sensor CT1 is a hall effect current sensor. It produces a differential output voltage proportional to the current that flows through it. The voltage is also

Figure 2.2 : Analog Circuitry



proportional to an excitation current used to drive the device. Since the voltage produced by an expected current of 5A is very small (on the order of 3-5 mV ) the conductor carrying the current is coiled so that it passes through the sensor 10 times. This increases the output voltage from the sensor by an order of magnitude. The sensor itself is seen as a short by the system since the current conductor only carries the current through the sensor. This should lessen the effect of the analyzer on the system that it is measuring.

The excitation current for the current sensor is provided by opamp U1 and MOSFET Q1. These form an almost ideal current source. The opamp is setup in a negative feedback loop to maintain the gate voltage of Q1 so that the voltage across R2 is equal to the voltage across the reference diode D1. The amount of current flowing through Q1 can then be controlled by the resistance of R2, a multiturn trimmer potentiometer. Since the output voltages on the current sensor will be between the voltages on the current supply terminals the current source was designed to work between ground reference and the negative supply. This causes the outputs of the sensor to be close to ground.

The output voltages produced by the current sensor are amplified to the desired signal level by U2, a precision

instrumentation amplifier. The trimmer resistor R3 across the outputs of the current sensor serves two purposes. It provides the bias currents for the inputs of the instrumentation amplifier. It can also be used to null the offset produced by the current sensor. The instrumentation amplifier is set up for a gain of 200. This was done by placing R5 in parallel with one of the internal resistors of the instrumentation amplifier. This configuration provides better temperature stability.

The signal flows into a sample and hold amplifier, U3. The hold capacitor C9 is a 5000 pF polypropylene capacitor. This was chosen for its low dielectric absorption and its temperature stability. A value of 5000 pF was chosen since the sample and hold is optimized for best performance with this value of a hold capacitor. The sampling control signal S/H is obtained from the timing and control circuitry discussed in the next section.

The voltage sensor is made up of a resistive divider that converts the 120 Vrms line voltages to voltages that lie within the +/- 10V range. The input impedance of the voltage sensor is 100K. This should produce little effect on the system being monitored. A transistor T1 protects the input of the sample and hold amplifier. The sample and hold, U4, is configured the same for the voltage sensor as



for the current sensor.

The signals from the sample and holds for the current and voltage sensors are then routed through U5, an 8 channel multiplexer. The address line A0 is controlled by the signal MA0 which is generated by the PIA (U40, Figure 2.8) under program control. This determines which sensor signal flows through to the ADC.

The output of the multiplexer is buffered by U6, a voltage follower. This is an opamp specifically designed for voltage follower applications. Its output impedance is very low, less than 5 ohms. This is needed to drive the analog to digital converter since it has a dynamically changing input impedance during a conversion. A large enough impedance on the driving element could cause conversion errors.

The signal then flows into the analog to digital converter to be quantized. The remainder of the signal conditioning and processing is done in software.

## Digital Section

The digital section of the harmonic analyzer makes the bulk of its circuitry. It performs the communication, storage, signal processing and general purpose control tasks of the harmonic analyzer.

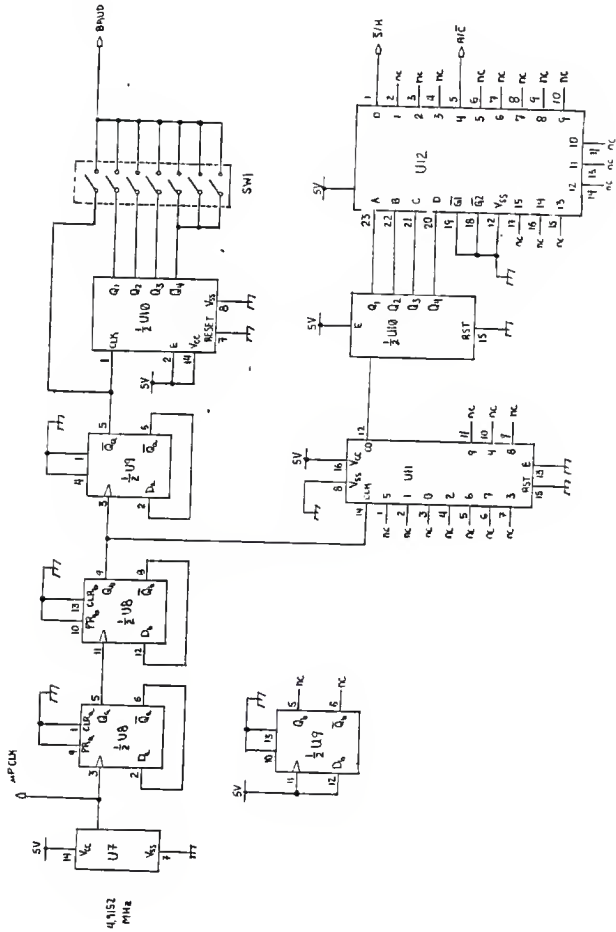
### Signal Description Standards:

The signal description standards used in this document will be as follows. An active low signal will be referred to with an asterisk preceding it, e.g. \*VPA, \*SM1. A signal not preceded by an asterisk is assumed to be active high. Certain signals whose state determines function, such as the read-write line, will be written so that the state to cause a function is marked using a bar to symbolize active low. No symbol is needed to symbolize active high for a state. The read-write line is written as  $R/\bar{W}$ . Since many of the signals in this system are active low, a signal will be said to be active or inactive instead of high or low in order to make this discussion easier to follow.

### Timing and Controls:

Figure 2.3 shows the circuitry for generating the various timing and control pulses used in the harmonic analyzer. These include the uP clock, sampling control and

Figure 2.3 : Timing and Control Circuitry



rate generation.

Timing for the entire system is derived from a 4.9152 MHz clock module, U7. This is used directly for the microprocessor clock. The 4.9152 MHz clock is then divided by 4 by the D flip-flops in U8. This results in a 1.2288 MHz clock signal.

The baud rate is generated by further dividing the frequency by two with half of U9. The resulting 614.4 KHz signal is then 9600 baud x 64. U10 generates the other baud rates by successively dividing the signal by two to get 4800x64, 2400x64, 1200x64 and 600x64. The frequency used is selected by SW1.

To generate the timing for the data sampling operation the 1.2288 MHz signal is divided by 10 using U11, a sequencer/divide by ten chip. This signal then drives the other half of U10, a 16 state counter whose output in turn drives a 16 state decoder, U12. The sample control S/H is the state 0 output line and the conversion control R/C is the state 5 output line. These signals provide all the timing needed by the various systems of the harmonic analyzer.

#### **Microprocessor and Support:**

The 68000 microprocessor forms the heart of the har-

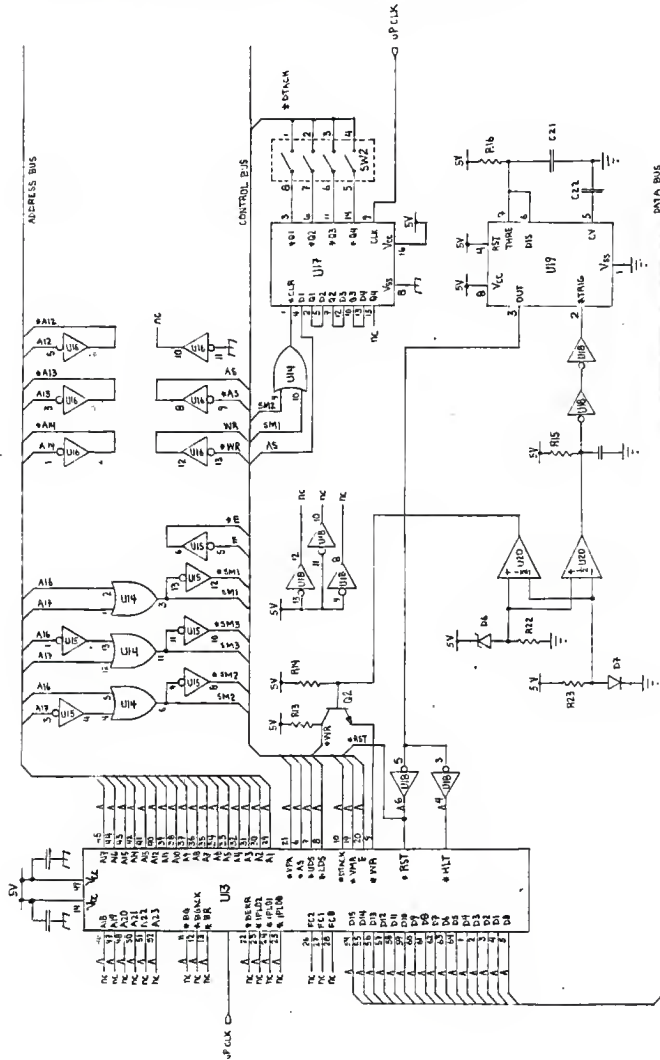
monic analyzer's digital section. It provides the basis for almost all the control signals of the digital section. In addition it provides the control and signal processing capabilities needed to meld the rest of the components together. Figure 2.4 shows the microprocessor and its related circuitry.

The microprocessor is clocked at 4.9152 MHz. Its I/O lines are connected to the 5V supply through a set of 4.7K pullup resistors. This allows the processor to reliably interface with the CMOS circuitry that forms the majority of the analyzer's digital section.

Almost all of the control signals in the digital section of the analyzer are derived from the microprocessor. Those provided directly by the processor include \*AS, \*UDS, \*LDS, \*VMA, E and R/W. Signals provided indirectly include SM1, SM2, SM3, \*VPA and \*DTACK.

The control signals SM1, SM2 and SM3 are decoded from the address lines A17 and A16 by U14 and U15. These signals and their complements decode the system memory into 64KB regions. SM1 is used to enable ROM. SM2 is used to enable the I/O devices and is fed back to the \*VPA input of the microprocessor to signal that this region uses a synchronous bus protocol instead of the asynchronous bus used by the rest of this system. This enables lower speed devices to be

Figure 2.4 : Microprocessor and Support Circuitry



interfaced to the 68000. SM3 is used to enable RAM.

The signal \*DTACK is derived from \*AS, SMI and SM3 by U14 and U17. This signal marks the end of a bus cycle for the asynchronous bus. It is only used for the regions of memory reserved for RAM and ROM. The time for a data transfer can be set by switch, SW2. This allows the timing to be adjusted for various memory chips.

The reset signal, \*RST, is not derived from any micro-processor signal. It is independently asserted on power up or whenever the supply voltage drops below 4.7V. This prevents the microprocessor from being put in a nonfunctional state by a power glitch. The comparator U20 performs the voltage sensing and the 555 timer and inverter chips, U19 and U18, provide the reset pulse of an appropriate length.

#### **System Memory, EPROM:**

The harmonic analyzer possesses 16K bytes of memory space devoted to ROM. The space is currently occupied by two 4Kx8 EPROMs, U21 and U22 as shown in Figure 2.5. These can be replaced by two 8K EPROMs if more program storage space is required.

The EPROMs block of memory is decoded by U23, a quad OR gate. The OR gates select the EPROM when \*SM1 and \*AS are

active. The output of the individual EPROMs are enabled

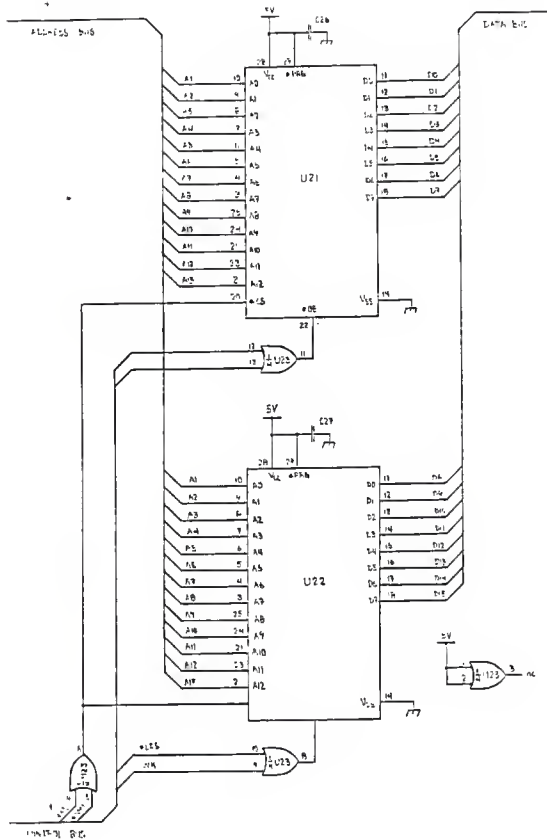


Figure 2.5 : System Memory, EPROM



when the data strobe to a chip is active and R/W is in the read mode. This prevents the EPROMs from placing data on the data bus and causing a bus collision if the EPROMs are inadvertently written to.

### **System Memory, RAM:**

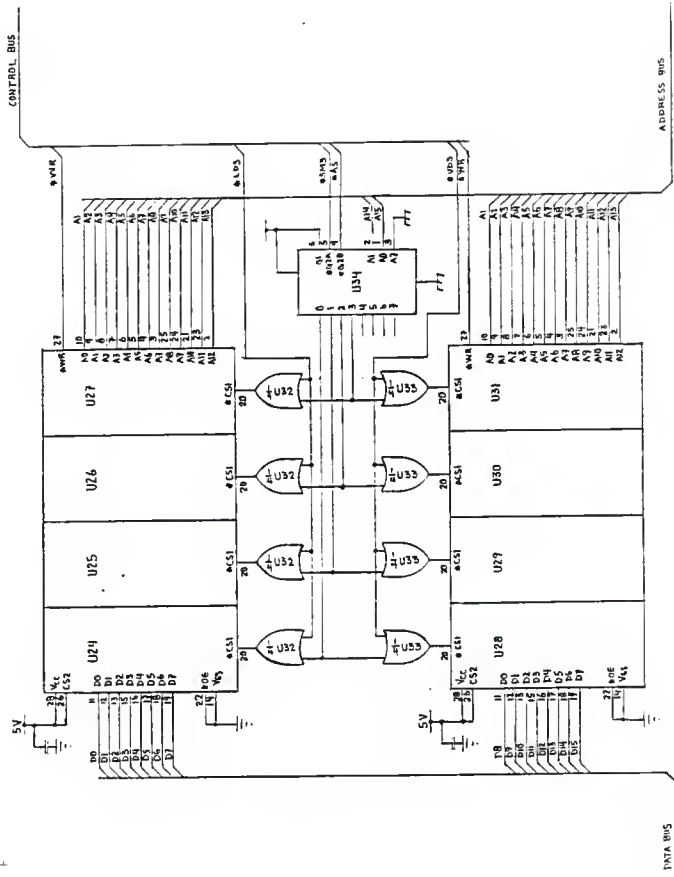
The harmonic analyzer possesses 64K bytes of static random access memory (RAM). It is structured as two 32K words due to the system's sixteen bit data bus. The RAM is battery backed, and so is nonvolatile as far as system power is concerned. Figure 2.6 shows the system RAM and associated memory decoding circuitry. Timing diagrams are shown in Appendix A.

The 64K bytes of RAM are made up of eight 6264 8Kx8 static RAM chips, U24 to U31. The memory block occupied by the RAM is \$02xxxx.

The decoding is done using the signals \*SM3 and \*AS to enable an eight state decoder, U34, which decodes A14 and A15 using its low four states. This gives a control line for each 16K bytes in the \$02xxxx block of memory addresses. These signals are then OR'ed with the data strobes, \*UDS and \*LDS, by U32 and U33 to get a chip enable for every 8K block of memory.

The grounds for the RAM chips are tied to the auxiliary

Figure 2.6 : System Memory, RAM



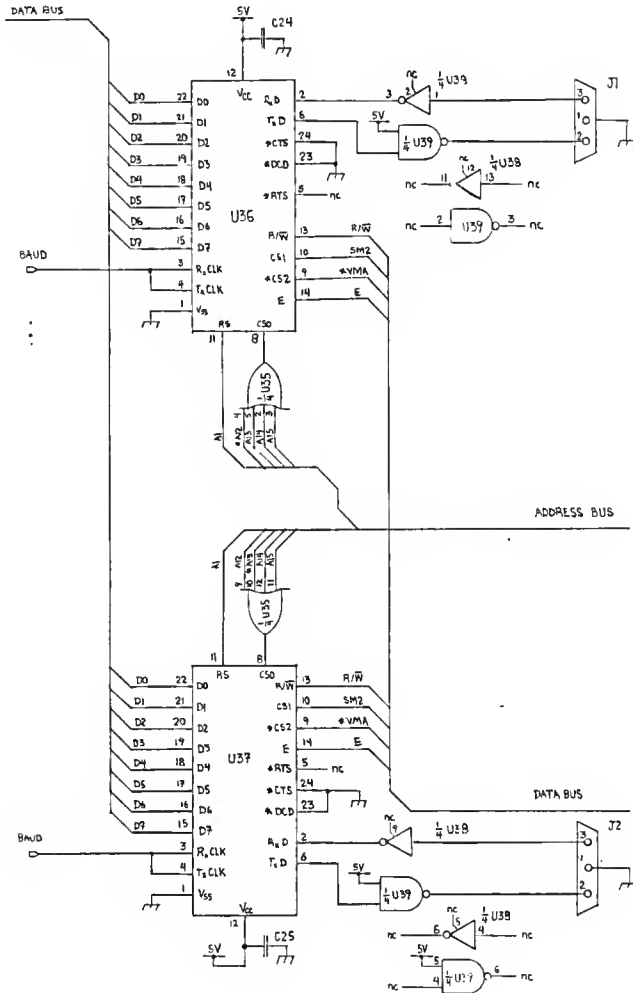
ground. This ground is only tied to the system ground when power is applied to the circuit. When power is down only devices on this ground possess a path for current flow and so can be backed up with a battery.

### **Serial Ports:**

The harmonic analyzer possesses two channels of serial communications. The schematics for the serial ports are shown in Figure 2.7. This is the only method of communicating with the external world on the harmonic analyzer. They run under an asynchronous protocol at variable baud rates. The output signals are at standard RS-232 levels. RS-232 control signals such as Clear To Send (CTS) and Request To Send (RTS) are not used. Provisions have been made on the board so that these signals may be used if desired for at least one channel. This would allow the analyzer to control a modem.

The serial ports, U36 and U37, are 6850 ACIAs. The ACIAs have their memory space decoded by U35, a dual 4 input OR gate. The outputs of the OR gates are connected to one of the three chip selects of the ACIAs. The remaining two chip selects of the ACIAs are connected to \*VMA and SM2. These are connected so that when all three signals are active an ACIA is enabled.

Figure 2.7 : Serial Ports and Support Circuitry



The RS-232 level shifting is done by U38 and U39, a 1489 quad receiver and a 1488 quad transmitter respectively. These have been set up so that the transmit data and receive data lines for each ACIA are converted to RS-232 levels. The other control signals are not presently used. Provisions have been made so that the control signals for one channel can be set up.

### **PIA, Real Time Clock, ADC**

The peripheral interface adapter (PIA), U40, is the port through which communication with the real time clock (RTC) and the analog to digital converter (ADC) is done. Its bidirectional ports are used to form a software controlled auxiliary bus. This is shown in Figure 2.8. Table 2.1 shows the pin designations of the PIA.

The ADC, U42 and RTC, U43 were connected to the system in this way for a number of reasons. The RTC was attached to the system in this manner because its bus timing is too slow to operate on the actual microprocessor bus. The ADC is configured this way so that it can sample data at a frequency determined by an independent crystal clock and not interfere with bus activity.

The auxiliary bus is controlled via software to produce the signals to enable, address and read or write to the RTC



or ADC. The only notable feature about the interface is the line controlling the \*WR of the RTC. When power goes down the WR\* of the RTC is forced inactive due to Q3 and an internal pullup on the \*WR pin of the RTC. This prevents false writes to the clock when system power is going down.

The RTC is also battery powered so it will continue to function when system power is cutoff. It is disconnected from the system ground by Q3 when power fails. This keeps the backup battery from powering the rest of the system. The system RAM is also tied to this auxiliary ground so that it is provided with power when power fails.

**TABLE 2.1 : PIA PIN DEFINITIONS**

---

**Control Pins:**

CA1 (input)  
conversion status from the A/D. A negative transition signals the end of conversion.

CA2 (output)  
mux address control. low selects the voltage sensor, high selects the current sensor.

CB1 (unused)

CB2 (unused)

**Port B Pins:**

b7 (i) : a/d data, d3  
b6 (i) : a/d data, d2  
b5 (i) : a/d data, d1  
b4 (i) : a/d data, d0

b3 (o) : a/d chip select, active low  
b2 (o) : clock write strobe, active low  
b1 (o) : clock read strobe, active low  
b0 (o) : clock select, active low

**Port A Pins:**

a7 (i/o) : A/D data, d11; clock data, d3  
a6 (i/o) : A/D data, d10; clock data, d2  
a5 (i/o) : A/D data, d9 ; clock data, d1  
a4 (i/o) : A/D data, d8 ; clock data, d0

a3 (i/o) : A/D data, d7 ; clock address, a3  
a2 (i/o) : A/D data, d6 ; clock address, a2  
a1 (i/o) : A/D data, d5 ; clock address, a1  
a0 (i/o) : A/D data, d4 ; clock address, a0

**PIA Register Locations:**

01 f001 - port a data/direction register  
01 f003 - port a control register  
01 f005 - port b data/direction register  
01 f007 - port b control register

---



## CHAPTER 3 : HARMONIC ANALYZER SOFTWARE

### Introduction

As with all microprocessor based pieces of equipment the software for the harmonic analyzer gives function to the form provided by the hardware. It provides the control for the sampling port, the signal processing capabilities and the communication protocol necessary to relay this information.

The control program for the harmonic analyzer, ARDVARK, is written in 68000 assembly code. It was assembled using the Avocet MAC68K macroassembler. This assembler possesses many features that simplified software development. The most important of these was the ability to create relocatable linkable object modules. These could then be linked to produce the final object code. This allowed the development of ARDVARK in its present modular structure.

The structure of ARDVARK follows the form recommended for software development. The main program, ARDVARK, is a short piece of code that makes use of numerous subroutines to accomplish its tasks. These subroutines in turn call other subroutines which further divide the task until it has been reduced to simplest terms. Development of each module in this fashion made debugging of the program an easier

task.

The subroutines tend to follow a general form. The register used for passing most arguments to and from a subroutine is d7 whenever possible. This was done in some cases where passing the argument in d7 was not the optimal solution in order to create a regular subroutine interface. Most of the routines save the contents of the microprocessor's register on the stack on entry. Any registers whose contents are changed by the routine are returned to their original state in exiting the routine. This makes the routines easy to utilize and prevents the occurrence of side effects. This in effect makes the I/O registers global variables and all other registers local variables in a routine.

Some of the routines that make up the harmonic analyzer's software package are not actual subroutines. They are instead program fragments. They return execution to a specific point rather than back to the routine that called them. The descriptions of these routines make note of this characteristic.

The remainder of this chapter discusses the various functions of ARDVARK. Table 3.1 shows a list of the various routines and a basic description of their functions. The routines are then discussed in more detail in the remainder

of this chapter. To aid in this discussion the 68000 micro-processor's architecture is shown in Figure 3.1. For further information on the 68000 see [1]. This discussion deals with the routines on a flow chart level. The theory needed for understanding the functioning of a routine is included in the description of the routine when necessary. For even greater detail on the routines consult the actual source code listings in Appendix B.

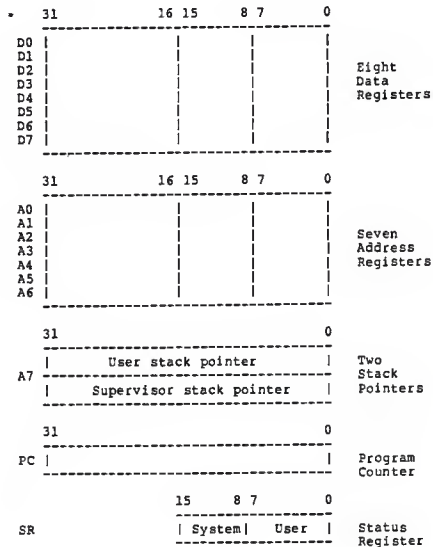


Figure 3.1 : The Programming Model For The 68000

**TABLE 3.1 : ARDVARK routines and functions**

---

ARDVARK:	main program.
ASCII.HEX:	convert ASCII coded hex digits to hexadecimal form.
CLOCK.READ:	reads a register of the real time clock.
CLOCK.WRITE:	writes to a register of the real time clock
DISP.REG:	outputs the contents of a uP register in ASCII coded hexadecimal digits through the secondary serial port.
DISP.STRING:	outputs characters in memory through the secondary serial port until an ASCII null is reached.
DOWNLOAD:	routine for downloading S-record files from a host computer.
FFT:	performs an FFT on integer data stored in memory.
GETCHAR:	gets a character from the secondary serial port.
GET.DATA:	gets one cycle of current and voltage data and stores it in memory
GET.TIME:	reads the time from the real time clock.
GO:	begins execution at a specified memory location.
HEX.ASCII:	converts a hex nybble to its equivalent ASCII code.
HEX.CHECK:	checks to see if a ASCII character corresponds to a hex nybble.

**TABLE 3.1 (continued)**

---

<b>IN:</b>	gets a character from the primary serial port and echoes it back.
<b>INITIALIZE:</b>	sets the time of the real time clock to a specified time and initializes system variables.
<b>MEMORY:</b>	allows the examination and alteration of memory contents.
<b>OUT:</b>	outputs a character through the primary serial port and waits for the character to be echoed back.
<b>PUTCHAR:</b>	outputs a character through the second serial port.
<b>QK.FFT:</b>	gets data, performs FFT and outputs results through the primary serial port.
<b>SAMPLE:</b>	controls the process of sampling and storing data.
<b>SEND.DATA:</b>	sends the stored data in the harmonic analyzers memory out the primary serial port.
<b>SEND.TIME:</b>	sends a four character string that gives the time read from harmonic analyzer's real time clock.
<b>TRACE:</b>	single steps through a program while allowing examination of all microprocessor registers.

---

**ARDVARK:**

ARDVARK is the main routine for the harmonic analyzer.

It performs all the control, computation and communication functions of the harmonic analyzer. It performs these functions by making calls to lower level routines. For this reason ARDVARK is basically a shell that handles the routing of program flow.

Figure 3.2 shows a flowchart of ARDVARK. On system bootup execution begins at the beginning of ARDVARK. ARDVARK first resets and configures the serial ports (6850 ACIA's) and then configures the parallel port (6821 PIA).

The cyclic portion of ARDVARK begins at this point. The status of the first serial port is checked to see if a byte has been received. If so the port is read and the character is decoded to find which function the harmonic analyzer should perform. Program flow is then routed to the appropriate subroutine. If the character received corresponds to no function then execution proceeds as if no code were received.

If no character were received ARDVARK then reads the time from the real time clock. If the clock setting is midnight (00:00) then the variable sys.go is made true. Sys.go is a variable that tells the system that sampling is enabled. Sampling is enabled at midnight this way. ARDVARK then checks the clock to see if the time is on the hour. If it is then ARDVARK waits until the next minute to take a

Figure 3.2: Flowchart of the control program ARDVARK.

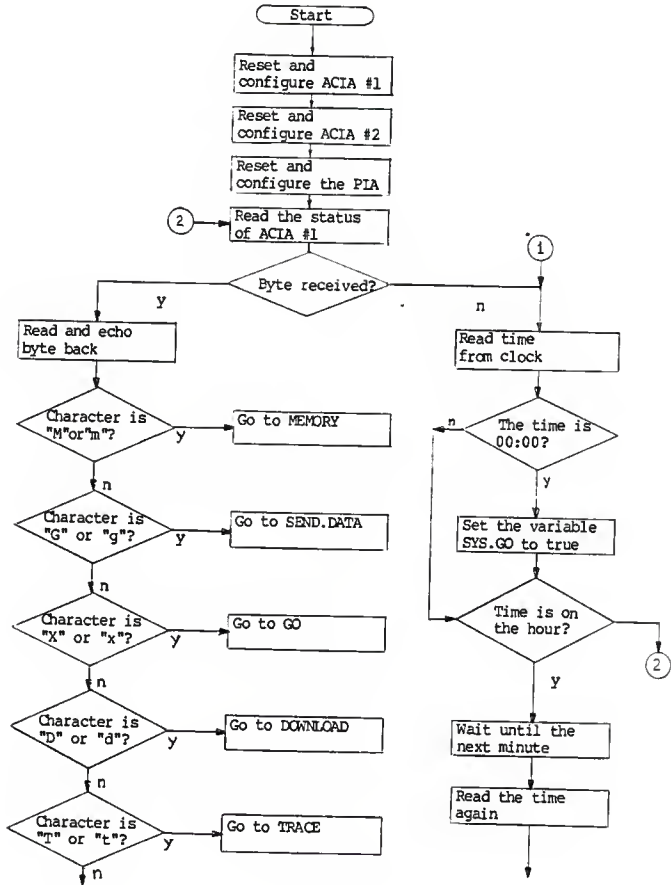
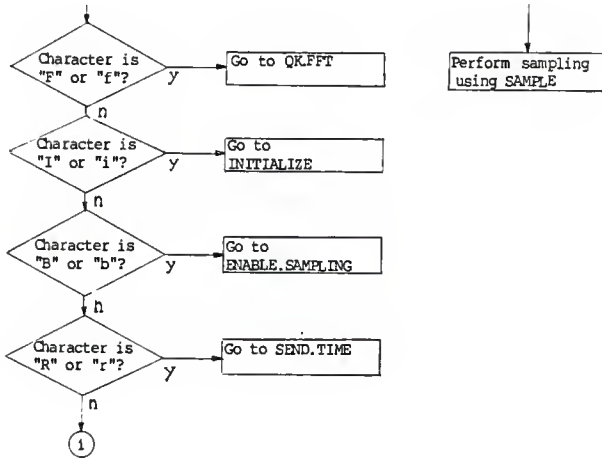


Figure 3.2: Flowchart of the control program ARDVARK.





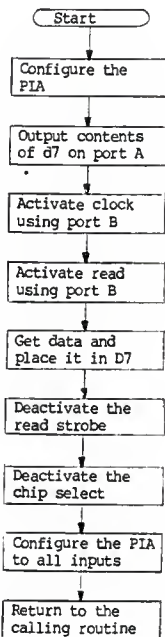
sample. ARDVARK is unavailable for external communication for this minute. This technique prevents overruns in the data sampling. If the time is not on the hour ARDVARK then goes back to the point where it reads the status of the serial port.

#### **CLOCK.READ:**

CLOCK.READ is a primitive that is used to read the data in a clock register. This routine can be used repetitively to read the time from the clock. It provides the control for the auxiliary bus needed to read from the real time clock. The number of the clock register to be read is passed in register d7 of the microprocessor. The data read is returned in d7 also. No other register contents are altered. Figure 3.3 flowcharts the operation of CLOCK.READ.

CLOCK.READ first configures the parallel port to set up the appropriate input and output lines. The register number is then output through the parallel port so that it is on the clock's address lines. CLOCK.READ then runs the control strobes, which are generated by the parallel port, to get the data from the clock. Since the data is read into bits 4-7 CLOCK.READ shifts it down to put it in bits 0-3. CLOCK.READ then configures the parallel port as it was on entry and returns to the calling routine.

Figure 3.3: Flowchart for the routine CLOCK.READ



**CLOCK.WRITE:**

This routine is another primitive. It is used to write data to a register of the real time clock. It can be used repetitively to set the clock to a given time. It works in much the same way as CLOCK.READ. A flowchart of its operation is shown in Figure 3.4.

The PIA is configured so that the A port is all outputs. The register number is passed in d7 output so that it is on the address lines of the clock. The data is output so that it is on the clock's data lines. The control strobes for the clock are then toggled so that the data is latched into the clock. The A port of the PIA is then configured as all inputs again. Execution then returns to the calling routine.

**DISP.REG:**

DISP.REG outputs the contents of a register through the second serial port. The contents of the register to display are placed in d7 for the call to DISP.REG. The register's contents are converted to ASCII coded hexadecimal and output through the secondary serial port. The data can be displayed on a terminal. No register contents are altered by this routine. A flowchart of DISP.REG is shown in Figure 3.5.

Figure 3.4: Flowchart for the routine CLOCK.WRITE

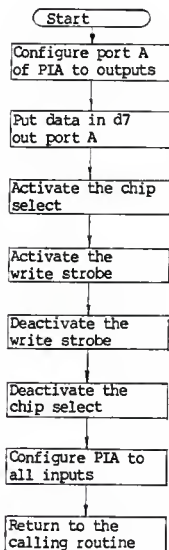
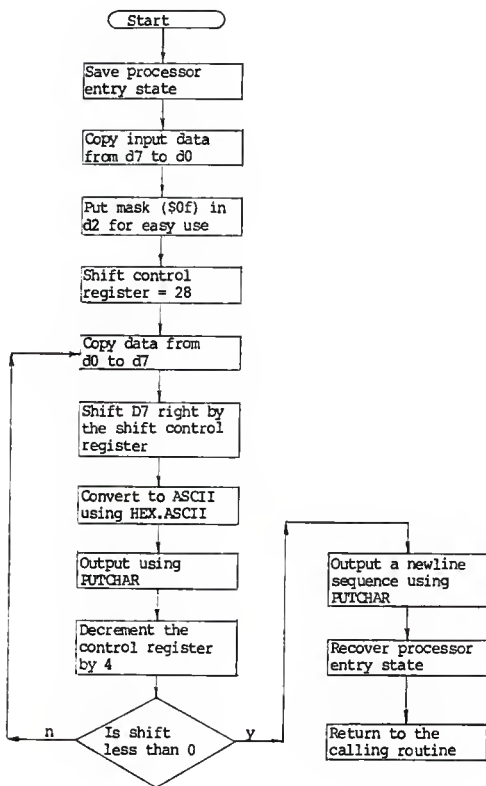


Figure 3.5: Flowchart for the routine DISP.REG

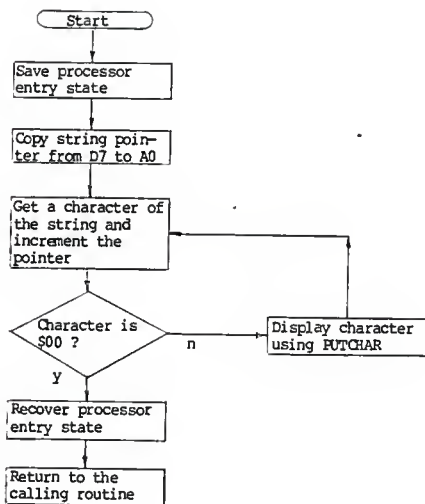


The contents of the microprocessor's registers are saved on entry to DISP.REG. The input data is saved in another register. Each successive nybble from the most to least significant is treated in the following manner. The input data is copied from the storage register back into d7. The nybble is shifted to the least significant position in d7. The nybble is then converted to ASCII by calling the routine HEX.ASCII. The resulting ASCII character is then sent out to the second serial port using the routine PUTCHAR. The shift control register is then adjusted so the next most significant nybble is shifted to the bottom of the register. When all of the input data has been converted and sent a carriage return and a linefeed are sent out using PUTCHAR. The processor's entry state is then restored and DISP.REG returns to the calling routine.

#### **DISP.STRING:**

DISP.STRING is a routine to send the contents of memory from a starting memory location, passed in d7, to an ending point signaled by an ASCII null, \$00. The data is sent out the second serial port. The data sent in this fashion is generally textual, hence the subroutines name. No register contents are altered by this routine. Figure 3.6 shows a flow chart of DISP.STRING.

Figure 3.6: Flowchart for the routine DISP.STRING



The entry state of the microprocessor's registers is saved. The starting location is then copied into an address register. A byte is read from memory into a data register and the address register is incremented to point at the next byte of memory. If the byte is not an ASCII null (\$00) it is sent out the second serial port using PUTCHAR. The process continues until a null is encountered. If an ASCII null is encountered then processor's entry state is restored and program flow returns to the calling routine.

#### **DOWNLOAD:**

DOWNLOAD allows S-record files to be downloaded from a host computer to the harmonic analyzer via a serial communications link. The S-record file format is the format used to store object code produced as an assemblers output. This format is the one specified by Motorola. For further information see [2]. The data is stored in hexadecimal form at the locations specified for it in the file itself. Using this routine programs can be downloaded to the harmonic analyzer and tested. Figure 3.7 shows a flow chart of DOWNLOAD.

DOWNLOAD operates in the following manner. It first searches for an S in the incoming data stream. When an S is found the next character is checked to see if it is a 1,2 or



Figure 3.7: Flowchart for the routine DOWNLOAD

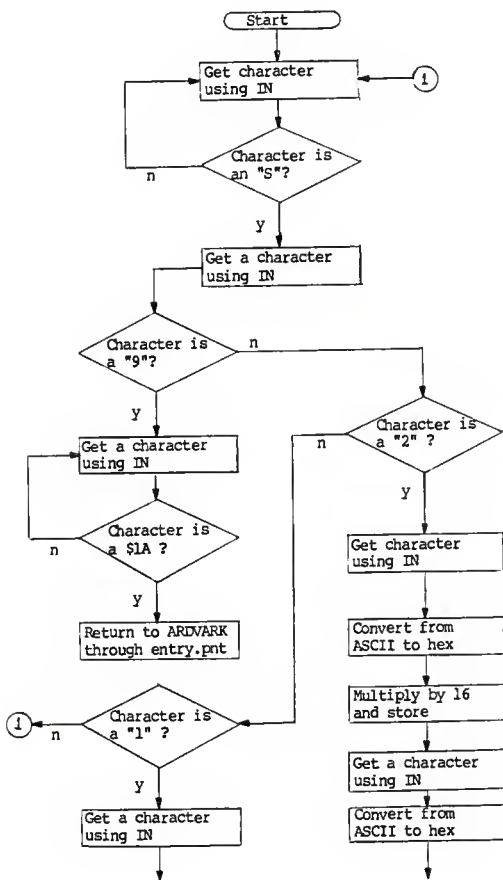


Figure 3.7: Flowchart for the routine DOWNLOAD

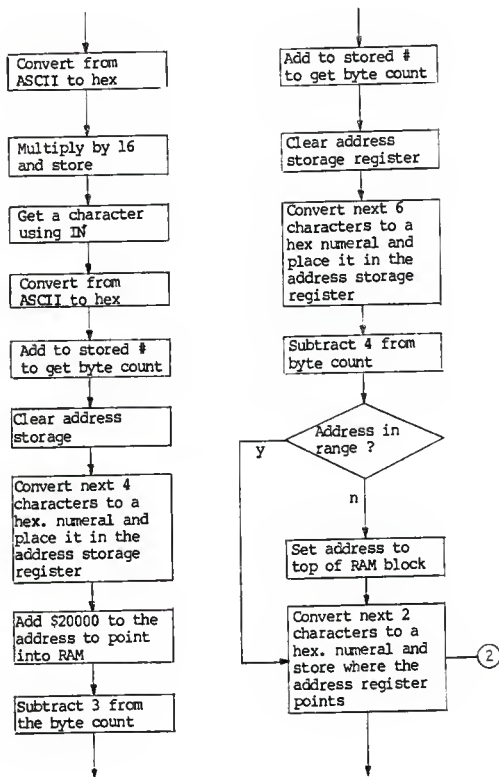
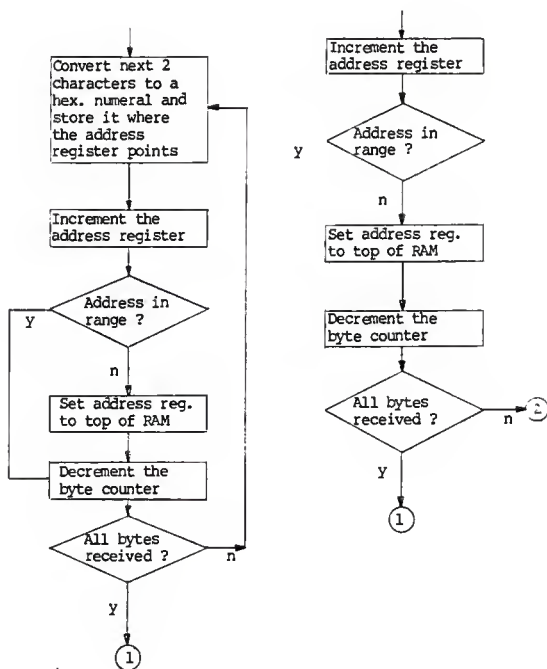


Figure 3.7: Flowchart for the routine DOWNLOAD



9 and program flow is routed to the S1, S2 or S9 line decoding portion of the routine, respectively. If the character after the S is not one of these three then download resumes looking for an S again in the same manner.

The S1 line decoding section functions as follows. The first two ASCII characters are read in and converted to a hexadecimal numeral with the first character being considered the most significant digit. This gives the number of character pairs or data bytes remaining in the line. The next four characters received are the ASCII coded address. These are similarly converted into a hexadecimal number and stored as the data pointer. The data byte counter is then reduced by three, two for the address and one for the checksum at the end of the line. The next pair of characters is converted to a hexadecimal numeral and stored at the position pointed at by the data pointer. The data pointer is then incremented and checked to see if it points at physical memory. If not, the data pointer is loaded with the last byte in RAM. The data counter is decremented and if there are more characters left in the line the process continues. When the line is completed program flow returns to the start of the routine.

The S2 line data line decoding routine functions as follows. As in the S1 decoding routine the first two charac-

ters are converted to a hexadecimal numeral that is the number of data bytes represented by the rest of the line. This is stored as the byte counter. The next six characters are converted from an ASCII coded hexadecimal number to binary. This is the address at which the data placement is to start. This is saved as the data pointer. If the address is out of range of the harmonic analyzers memory then the address of the last byte of RAM is placed in the data pointer. This allows data to be placed anywhere in the 68000's addressing range. Data is converted and stored in the same way as in the S1 decoding routine. When the end of the line is reached program flow returns to the beginning of the routine.

The S9 line is the terminating line of an S-record file. The routine for decoding it merely waits until all the data has been sent and only filler data from a disk file is being sent. This is sent as \$1A's on the host computer used to develop the harmonic analyzer. When this state is reached program flow returns to ARDVARK through the node entry.pnt.

#### **FFT:**

The routine FFT is the heart of the harmonic analyzer's software package. It performs a decimation in frequency fast Fourier transform (FFT) on a complex input sequence. The input data is assumed to be in offset integer. A pointer to

the memory block storing the samples is passed in register a6 of the processor. No registers are altered by this routine.

The fast Fourier transform (FFT) is actually a discrete Fourier transform (DFT) calculated by a fast algorithm. The DFT is calculated by equation 1.

$$X(n) = \sum_{m=0}^{N-1} x(m) e^{-j2\pi nm/N} \quad \dots(3.1)$$

where:  $n = 0, 1, \dots, N-1$   
 $m = 0, 1, \dots, N-1$   
 $N =$  number of points  
 $X(n) =$  sequence of Fourier coefficients  
 $x(n) =$  input sequence

The DFT shows the distribution of the energy of a signal over a set of discrete frequencies. For further information on the properties of the DFT see references [3], [4] and [5]. A fast algorithm can be used to compute the DFT with a large time savings over the straight forward computation of the equation above. This computation is called the fast Fourier transform or FFT.

The fast algorithm used to compute the DFT on the harmonic analyzer is the Cooley-Tukey decimation in frequency FFT. The decimation in frequency algorithm functions only for sequences of length 2. If a sequence is not of length 2 then it must be zero padded or truncated so that it is.

The decimation in frequency algorithm can be derived from the DFT in the following manner. For a more indepth treatment see [3], [4] and [5]. The equation for calculating the DFT can be rewritten by dividing the input sequence in two.

$$X(n) = \sum_{m=0}^{N/2-1} x(m) e^{-j2\pi nm/N} + \sum_{m=N/2}^{N-1} x(m) e^{-j2\pi nm/N} \quad \dots(3.2)$$

$$X(n) = \sum_{m=0}^{N/2-1} x(m) e^{-j2\pi nm/N} + e^{-j(2\pi/N)(N/2)n} \sum_{m=0}^{N/2-1} x(m+N/2) e^{-j2\pi nm/N}$$

$$\text{But } e^{-j(2\pi/N)(N/2)n} = e^{-j\pi n} = (-1)^n$$

$$X(n) = \sum_{m=0}^{N/2-1} [x(m) + (-1)^n x(m+N/2)] e^{-j2\pi nm/N}$$

This can be rewritten for the odd and even DFT coefficients.

For the even coefficients this becomes:

$$X(2c) = \sum_{m=0}^{N/2-1} [x(m) + x(m+N/2)] e^{-j(2\pi/N)2cn}$$

$$X(2c) = \sum_{m=0}^{N/2-1} [x(m) + x(m+N/2)] e^{-j2\pi nc/(N/2)}$$

... (3.2a)

For the odd coefficients this becomes:

$$X(2c+1) = \sum_{m=0}^{N/2-1} [x(m) - x(m+N/2)] e^{-j(2\pi/N)(2c+1)n}$$

$$X(2c+1) = \sum_{m=0}^{N/2-1} [x(m) - x(m+N/2)] e^{-j2\pi m/(N/2)} e^{-j2\pi nc/(N/2)}$$

where  $c = 0, 1, \dots, N/2-1$

... (3.2b)

The two sequences can be seen to be DFT's themselves. The even DFT coefficients can be seen to be the result of taking the DFT of the  $N/2$  point sequence  $\{x(m) + x(m+N/2)\}$  where  $m=0, \dots, N/2-1$ . The odd DFT coefficients are the re-



result of taking the DFT of the  $N/2$  point sequence  $\{ [x(m) - x(m+N/2)] \exp(-j(2\pi m)/(N/2)) \}$  where  $m=0, \dots, N/2-1$ . These DFT's can be broken up in the same manner. This continues as many times as necessary to reduce the sequences to one point sequences. The one point sequences are the DFT coefficients,  $X(n)$ .

The problem with this technique is that the results are not in the correct order. Consider an eight point DFT done via this technique. The results of splitting the sequence yields a sequence whose DFT is  $X(0), X(2), X(4), X(6)$  and another sequence whose DFT is  $X(1), X(3), X(5), X(7)$  of the original sequence. This is shown in Figure 3.8. The figure used to explain this is a signal flow graph. An explanation of signal flow graphs is given in Appendix D.

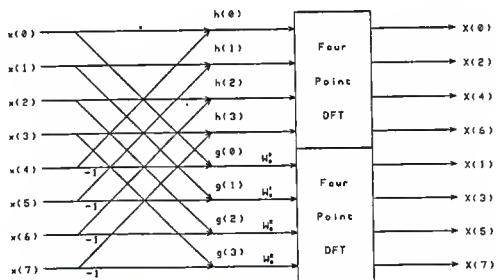


Figure 3.8: Eight point DFT divided into two four point DFT's.

Performing the division of the sequences again on the four point sequences further scrambles the results. This is shown in Figure 3.9 for the subsequence marked  $h(0)$ ,  $h(1)$ ,  $h(2)$  and  $h(3)$  above.

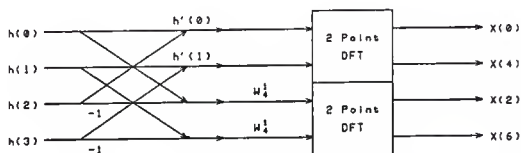


Figure 3.9 : A four point DFT divided into two two point DFT's.

The two point FFT does not further scramble the results. This is because the sequences have only one point so division into the odd and even sequences does not reorder the elements. Also the two point FFT is actually the DFT. A two point DFT is shown below in Figure 3.10.

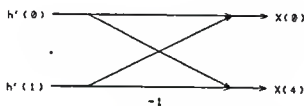


Figure 3.10 : A two point DFT

An unscrambling action is needed to place the resulting coefficients in the proper order. The sequence resulting from the FFT,  $R(n)$ , are actually the scrambled results of the DFT,  $X(n)$ . The relationships for an eight point FFT are shown in Figure 3.11. Due to the symmetrical relationships of the unscrambling action it is often referred to as the "butterfly transfers."

The output of the FFT can be placed in order by swapping some of the elements. Swapping  $R(1)$  with  $R(4)$  and  $R(3)$  with  $R(6)$  will put the output of an eight point FFT into the proper order. (See Figure 3.11.) The pairs of elements to be swapped can be determined by bit reversing the indices of the FFT coefficients.

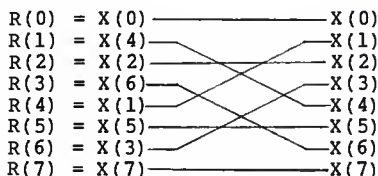


Figure 3.11 : The relationship between the output of the FFT,  $R(n)$ , and the DFT,  $X(n)$  for an eight point sequence.

The bit reversal is performed in the following way. Express the index of a coefficient in binary form. The number of bits used is determined by the length of the

sequence. For an eight point DFT 3 bits are needed to express all possible indices. For  $R(1)$  the index would be 001. Bit reversing this gives 100, or 4.  $R(1)$  would then be swapped with  $R(4)$  to unscramble the output. Some numbers, such as 2 and 5 in this case, have the same value when bit reversal is performed. This means that the element is already in its correct location in the sequence.

This process continues until all related pairs of elements are swapped. Once all the pairs are swapped the output of the FFT is in the correct order.

The signal flow graph showing the calculations needed for an eight point FFT are shown in Figure 3.12. Figure 3.13 shows a flow chart of the routine FFT used to calculate a 128 point FFT. For comparison the listing of a FORTRAN program that performs a decimation in frequency FFT is given in Listing 3.1.

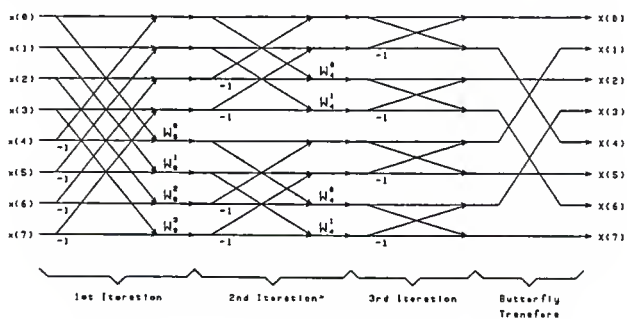


Figure 3.12: Signal flow graph of an 8 point DFT

Figure 3.13: Flowchart of the routine FFT

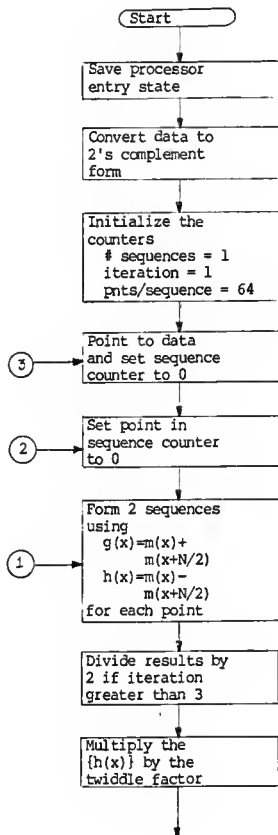


Figure 3.13: Flowchart of the routine FFT

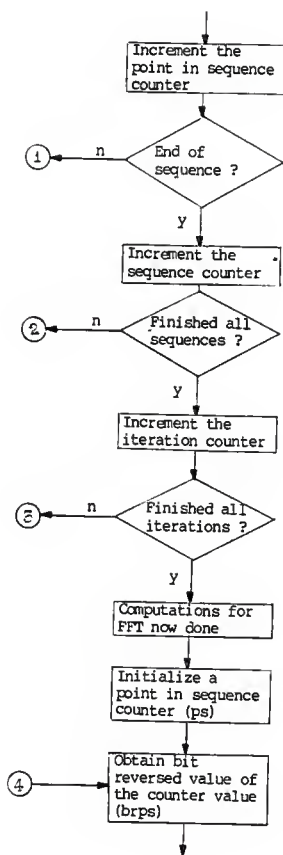
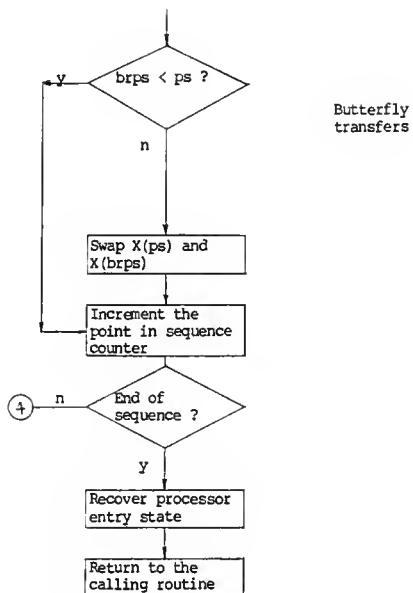


Figure 3.13: Flowchart of the routine FFT



### Listing 3.1: FFT routine in FORTRAN

```

C*****
      FAST FOURIER TRANSFORM
C
C      DG FORTRAN 5 SOURCE FILENAME:          CFFT.FR
C
C      DEPARTMENT OF ELECTRICAL ENGINEERING   KANSAS STATE UNIVERSITY
C
C      REVISION          DATE                PROGRAMMER
C      -----          ----                -
C      00.0              FEB 02, 1980       FRED RATCLIFFE
C
C*****
      CALLING SEQUENCE
C
C              CALL FFT(X,N,INV)
C
      PURPOSE
C
C              FAST FOURIER TRANSFORMATION
C
      ROUTINE(S) CALLED BY THIS ROUTINE
C
C              NONE
C
      ARGUMENT(S) REQUIRED FROM THE CALLING ROUTINE
C
C              X      -      COMPLEX VECTOR TO BE TRANSFORMED
C              N      -      NUMBER OF POINTS TO BE TRANSFORMED
C                          (MUST BE A POWER OF TWO)
C              INV    -      INV=0 -> FORWARD TRANSFORM
C                          INV=1 -> INVERSE TRANSFORM
C
      ARGUMENT(S) SUPPLIED TO THE CALLING ROUTINE
C
C              X      -      COMPLEX TRANSFORMED VECTOR
C
C*****
      NOTE 1: This subroutine makes no checks on the validity
              of the data supplied by the calling routine.
C
      NOTE 2: Argument(s) supplied by the calling routine are
              not modified by this subroutine.
C
      NOTE 3: This subroutine was obtained from 'ORTHOGONAL TRANSFORMS FOR
              DIGITAL SIGNAL PROCESSING' by N. Ahmed.
C
C*****
      SUBROUTINE FFT(X,N,INV)
      COMPLEX X(1),W,T
      ITER=0
      IREM=N
10      IREM=IREM/2
      IF (IREM.EQ.0) GO TO 20

```



Listing 3.1: FFT routine in FORTRAN

```

ITER=ITER+1
GO TO 10
CONTINUE
S=-1
IF (INV.EQ.1) S=1
NXP2=N
DO 50 IT=1, ITER
NXP=NXP2
NXP2=NXP/2
WFWR=3.141592/FLOAT(NXP2)
DO 40 M=1, NXP2
ARG=FLOAT(M-1)*WFWR
W=CMPLX(COS(ARG), S*SIN(ARG))
DO 40 MXP=NXP, N, NXP
J1=MXP-NXP+M
J2=J1+NXP2
T=X(J1)-X(J2)
X(J1)=X(J1)+X(J2)
40 X(J2)=T*W
50 CONTINUE
N2=N/2
N1=N-1
J=1
DO 65 I=1, N1
IF (I.GE.J) GO TO 55
T=X(J)
X(J)=X(I)
X(I)=T
65 K=N2
IF (K.GE.J) GO TO 65
J=J-K
K=K/2
GO TO 60
65 J=J+K
IF (INV.EQ.1) GO TO 75
DO 70 I=1, N
70 X(I)=X(I)/FLOAT(N)
75 CONTINUE
RETURN
END

```

#### **GET.CHAR:**

This routine gets a character from the second serial port. The character read in is returned in the low byte of d7. No other registers are altered by this routine. Figure 3.14 shows a flowchart of GET.CHAR.

The status of the second serial port is read into d7. If no character has been received the status is checked until one has been received. When a character has been received it is read into d7 and GET.CHAR returns to the calling routine.

#### **GET.DATA:**

This routine controls the data sampling process. It samples one cycle of voltage waveform and then one cycle of current waveform. The data resulting from this process is stored in two sequential blocks of memory set aside for this purpose. (See the system memory map, Appendix C. ) Execution returns to the calling routine at the end of the sampling process. No registers are altered by this routine. Figure 3.15 shows the flow chart of GET.DATA's operations.

GET.DATA operates in the following manner. The registers whose contents are changed in this routine are saved on the stack for recovery later. The PIA is configured to read data from the analog to digital converter (ADC) . A pointer

Figure 3.14: Flowchart of the routine GETCHAR

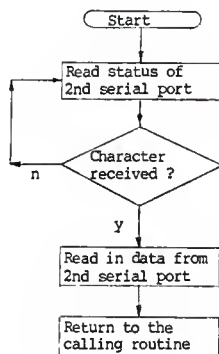


Figure 3.15: Flowchart of the routine GET.DATA

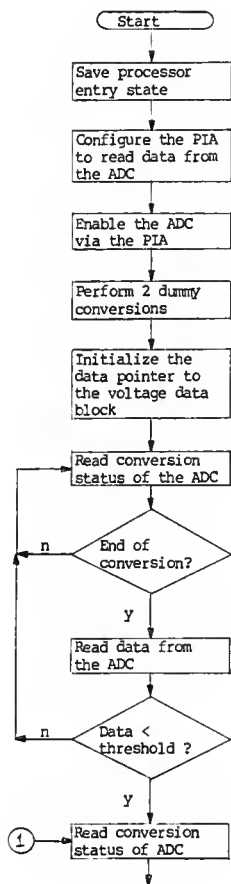


Figure 3.15: Flowchart of the routine GET.DATA

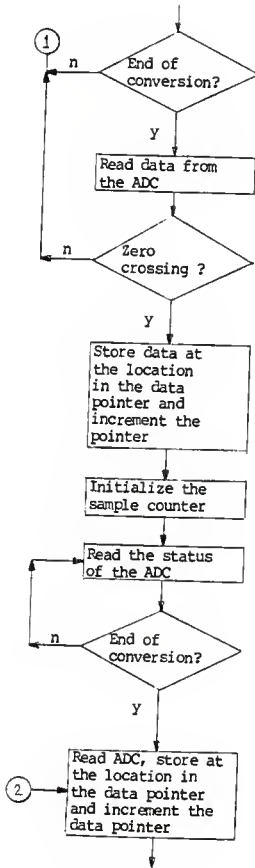


Figure 3.15: Flowchart of the routine GET.DATA

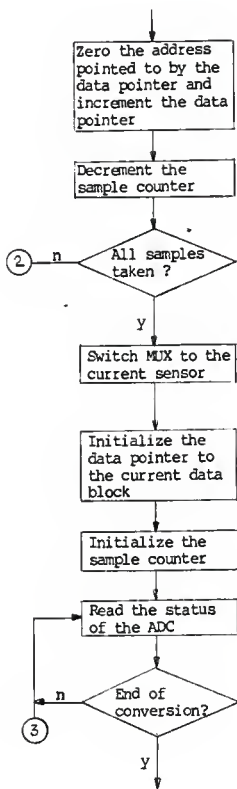
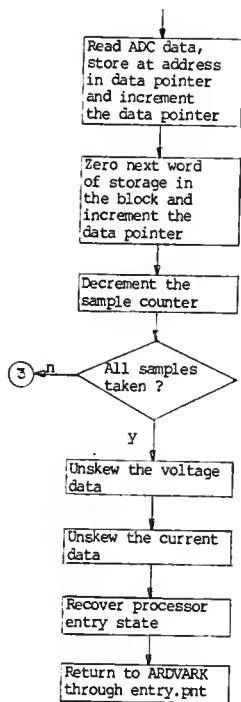


Figure 3.15: Flowchart of the routine GET.DATA



is initialized to point to the voltage data storage block. The ADC is then enabled and two dummy conversions are performed.

GET.DATA then functions as a software zero crossing detector. Data storage does not begin until a zero crossing is detected. The zero crossing detector first waits until the voltage waveform has reached a negative threshold. This ensures that the detector will reliably monitor the zero crossing. It also means that sampling will not occur until an AC voltage of a large enough magnitude is monitored. The waveform is then monitored until a zero crossing is detected. One hundred and twenty eight samples are then obtained and stored in the block beginning where the data pointer pointed. The PIA is then used to switch the channel of the multiplexer. This connects the current sensor to the ADC. One hundred and twenty eight samples of current data are then taken. These are stored in a block of memory directly after the voltage data.

The output of the 12 bit ADC is stored in the high 12 bits of the 16 bit word, not the low twelve bits. For this reason the data is read back, shifted down four bits and all unessential bits cleared. This is done for both the voltage and current data.

Once the data is shifted GET.DATA prepares to return



to the calling routine. The entry values of altered registers are recovered from the stack. Execution then returns to the calling routine.

#### GET.TIME:

This routine reads the time from the harmonic analyzers real time clock. It returns a word made up of four BCD digits that represent the hours and minutes. This word is returned to the calling routine in register d7. No registers other than d7 are altered by this routine. Figure 3.16 shows a flow chart of GET.TIME.

GET.TIME operates as follows. The registers changed in the routine are saved on the stack. The units of minutes register is read using the routine READ.CLOCK. The value is stored for later use. A register is then cleared to store the time. The tens of hours register is read from the clock and added to the time storage register. Since the time data is in the low nybble of the I/O register, d7, this places the time data in the low nybble of the time storage register. The time storage register is then shifted left by four bits. This clears off the low nybble of the time storage register so that another nybble can be added into that location. The same procedure is used to read in the data from the units of hours and tens of minutes registers of the clock. The units of minutes information is treated the same way except that the time storage register is not shifted once the data is added in. The value read in from the units of minutes register is then compared with its initial value

Figure 3.16: Flowchart of the routine GET.TIME

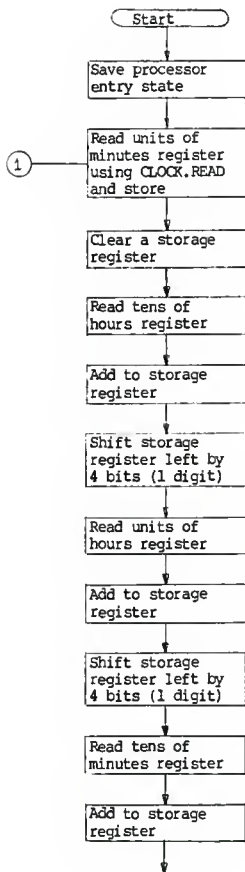
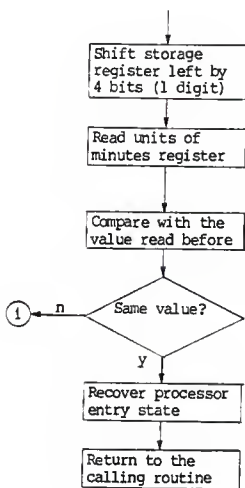


Figure 3.16: Flowchart of the routine GET.TIME



that was stored at the beginning of the routine. If these are different then the read was performed during a clock transition on the registers of interest and the read is reperformed. If the value had not changed then the time is placed in the I/O register, d7, and the registers are returned to their entry state. Execution then returns to the calling routine.

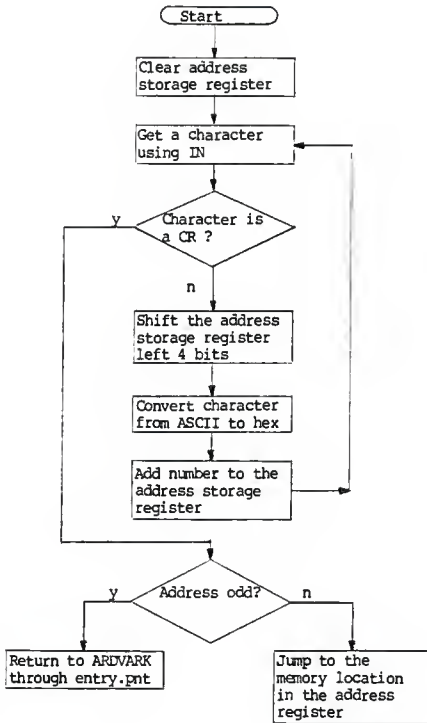
#### GO:

The routine GO provides a means for executing programs stored in the harmonic analyzers memory. The memory address is specified by the user on the host computer. A flow chart of GO is shown in Figure 3.17

The operation of the routine GO is fairly straightforward. The address storage register, d0, is cleared. A character is then fetched from the primary serial port. If the character is a carriage return then the program checks the address that is stored to see if it is even. If it is the program jumps to that memory location. If the address is odd then GO jumps to the entry point of ARDVARK.

If the character received is not a carriage return (CR) then it is converted from ASCII to a hexadecimal digit by the routine ASCII.HEX. The storage register is multiplied by 16 to shift the value in it left one hex digit. The just

Figure 3.17: Flowchart of the routine GO



received number is then added into the opening caused by the shift. Another character is fetched and the process repeats until a CR is received.

#### **HEX.ASCII:**

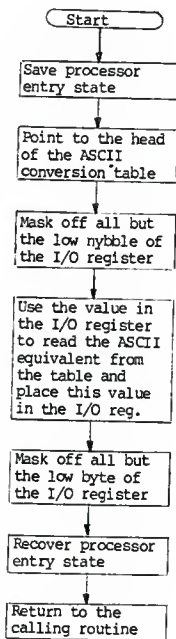
HEX.ASCII performs the conversion of a hex digit to its ASCII equivalent representation. The hexadecimal digit is passed in the low nybble of d7. The returned ASCII character is in the low byte of d7. Other than d7 no register contents are altered by this routine. Figure 3.18 shows a diagram of the HEX.ASCII routine.

On entry to HEX.ASCII the registers used by the routine are saved on the stack. A pointer to the conversion table is then initialized. The register d7 is then masked so that all but the low four bits are cleared. The resulting number is then used as the offset into the conversion table. For each hex number its ASCII equivalent is stored as an entry in the table. The entry pointed to is then retrieved and placed in d7. The registers altered in the routine are then recovered from the stack where they were saved on entry. Execution then returns to the calling routine.

#### **HEX.CHECK:**

HEX.CHECK examines an ASCII character and determines if

Figure 3.18: Flowchart of the routine HEX.ASCII





it is the equivalent of a hex number. If it is an equivalent then the carry bit is cleared, otherwise it is set. The character to be checked is passed to the routine in register d7. No registers are altered by this routine. The flowchart for HEX.CHECK is shown in Figure 3.19.

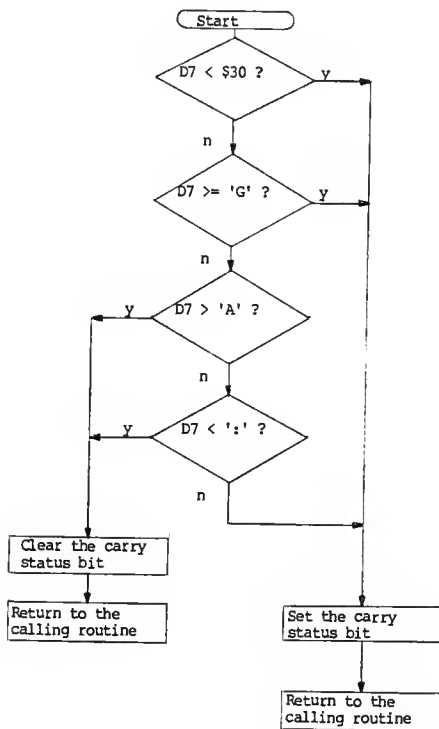
The character is first compared with ASCII zero, \$30. If it is less than \$30 then the carry is set and the program returns to the calling routine. If greater than \$30 then it is compared with ASCII "G", or \$47. If it is greater than \$47 then the carry is set and the routine returns to the calling routine.

If the character is between "0" and "F" it can still be a number of nonhexadecimal equivalent ASCII characters. An additional check is then done to see if the character is less than ":", \$3A, or greater than "@", \$40. If this is the case then the carry is set, otherwise the carry is cleared to signal a nonhex equivalent character. Execution then returns to the calling routine.

#### **IN:**

The routine IN controls input to the harmonic analyzer via the primary serial port. It echoes back whatever character it receives to show that the character has been received. The character received is returned in register d7.

Figure 3.19: Flowchart fo the routine HEX.CHECK



No other registers are altered by this routine. Figure 3.20 shows a flowchart of IN.

The first action of the routine IN is to read the status of the primary serial port. If no character has been received then the status continues to be read until one is received. The character is then read in from the serial port. The status of the port is read again to determine whether the transmit register is ready. If not the status is read until it is. The received character is then echoed out the primary serial port. Execution then returns to the calling routine.

#### **INITIALIZE:**

INITIALIZE is a routine to initialize the harmonic analyzer. It sets the harmonic analyzer's real time clock and initializes the system variables. No registers are altered by this routine. A flowchart of INITIALIZE is shown in Figure 3.21.

INITIALIZE begins by saving the contents of the processor registers that are changed by the routine on the stack. It then reads a character string that gives the time to set the clock to. The system variables are then initialized. These include the data pointer, sample counter and sampling enabled flag. The rest of the routine configures and sets

Figure 3.20: Flowchart of the routine IN

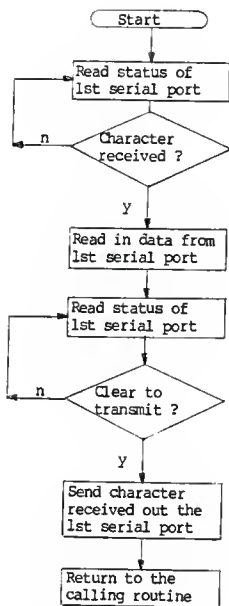


Figure 3.21: Flowchart of the routine INITIALIZE

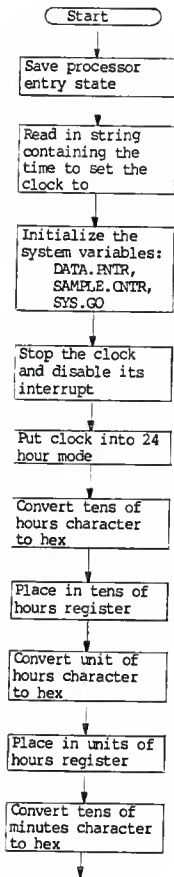
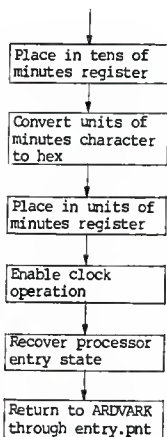


Figure 3.21: Flowchart of the routine INITIALIZE



the clock.

The clock is set using the routine CLOCK.WRITE. The clock is first halted with its interrupts masked. The clock is then put into the twenty four hour mode with the interrupts disabled. The first character in the time string is then converted to hexadecimal using ASCII.HEX. This number is then written to the tens of hours register of the harmonic analyzer's clock. The second character of the string is then converted and placed in the units of hours register of the clock. The third and fourth characters of the time string are converted and then placed in the tens and units of minutes registers respectively. The clock is then started. This concludes the functions of the routine and the return sequence begins.

The registers altered by this routine are placed in their original state. Program flow then returns to ARDVARK at the entry node entry.pnt. Since program flow does not return to the calling routine this is not a true subroutine.

**MEMORY:**

MEMORY is the routine that controls the examination and alteration of memory locations. It is not a subroutine as it is called by a normal jump and returns to ARDVARK through the entry point entry.pnt (see listings in Appendix B).

Figure 3.22 shows the flowchart of memory.

Memory functions in the following manner. It first clears a register for address storage. A character is then fetched from the primary serial port. If it is a carriage return (CR) then execution proceeds by testing the received address. If the character is not a CR then it is converted to hexadecimal using ASCII.HEX. The address storage register is shifted left by four bits to cause a shift of the stored address one hex digit to the left. The digit just received is then added in as the least significant digit. This continues until a CR is received.

The received address is then tested. It is first checked to see if the address is in the range of the actual physical memory of the harmonic analyzer. If not an error code is sent out the primary serial port and the execution returns to ARDVARK through entry.pnt. If the address is in range then it is checked to see if it is an even address. If not even then an error code is sent and execution returns to ARDVARK. If it is even then the code for no error is transmitted.

MEMORY now displays the data at an address. First the address is sent in ASCII coded hexadecimal digits. The data at the address is then fetched and sent in the same manner.



Figure 3.22: Flowchart of the routine MEMORY

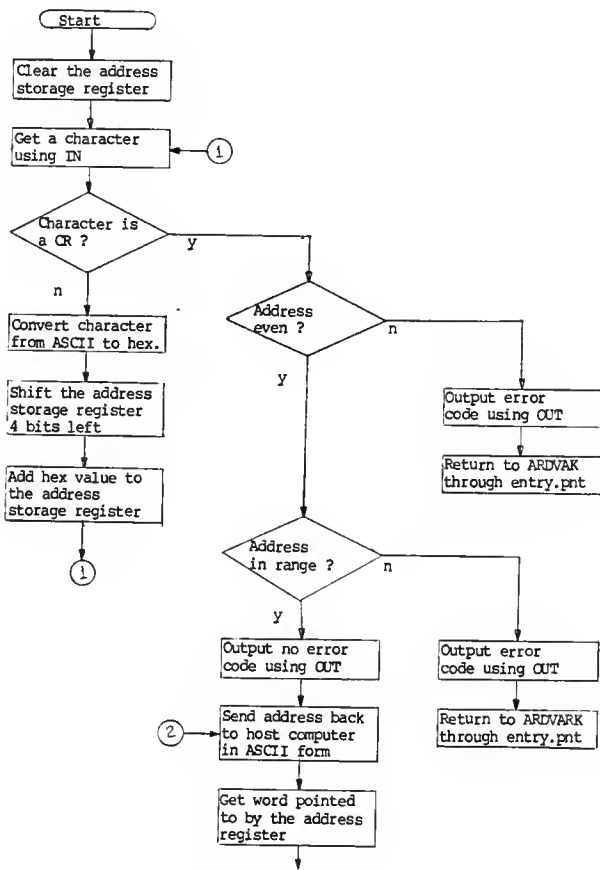


Figure 3.22: Flowchart of the routine MEMORY

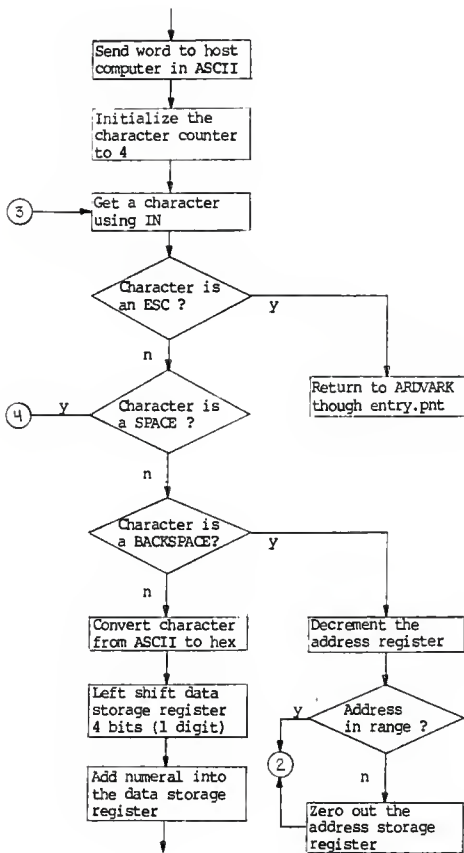
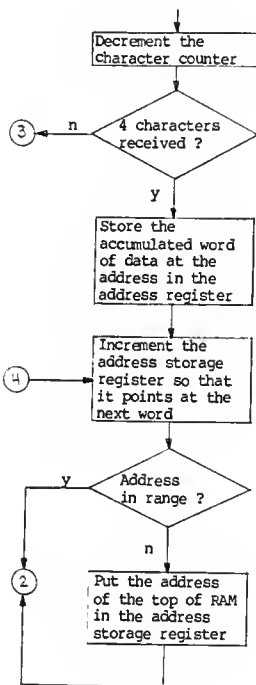


Figure 3.22: Flowchart of the routine MEMORY

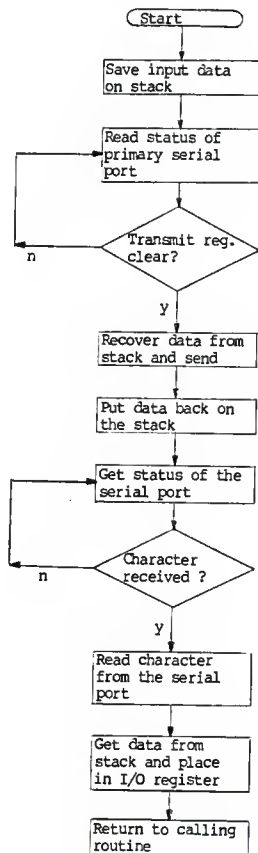


The next section of memory allows the alteration of the displayed memory location. It also allows the user to increment or decrement the address without altering the contents of the present location. This section is set up so that when four ASCII coded hexadecimal digits are received the value they represent is stored in the memory location addressed. If a space or back space is received at any point in this sequence the address is incremented or decremented, respectively. If an escape character, \$1B, is received then execution is returned to ARDVARK through entry.pnt. When the address is incremented or decremented it is checked to see if it still is in the memory range of the harmonic analyzer. If the address is not in range it is set to \$0000 or the top of RAM depending on whether the fault occurred while the address was decremented or incremented, respectively.

#### OUT:

The routine OUT is the analog of the routine IN. It controls the transmission of data out the primary serial port. Once a character has been sent out the port the routine waits for it to be echoed back. This keeps the transmitter from overrunning the receiver. The character to be sent is passed to the routine in the low byte of d7. No register contents are altered by this routine. Figure 3.23 shows a flowchart detailing the operation of OUT.

Figure 3.23: Flowchart of the routine OUT



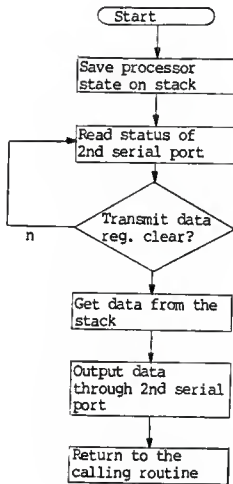
OUT begins operation by saving the character to send on the stack. It then reads in the status of the primary serial port. If the transmit data register is not ready then status is read until it is. When the port is ready to transmit, the data is recovered from the stack and sent out the port. The data is saved once again on the stack. The routine now waits for an echo. This prevents data from being sent out faster than it can be received. The status of the port is then read. If no character has been received then the status is read until one is. Once a character has been received the port is read. The data is recovered from the stack again and execution returns to the calling routine.

**PUTCHAR:**

PUTCHAR is a I/O primitive. It transmits a character out the secondary serial port. Unlike the routine OUT, PUTCHAR does not wait for an echo to be returned from the receiver. The character to be transmitted is passed in the low byte of d7. No registers are altered by this routine. Figure 3.24 shows the operation of PUTCHAR.

PUTCHAR operates in the following manner. It places the data to be sent on the stack. The status of the second serial port is then read to see if the transmit register is ready. If the transmit register is not ready then the status

Figure 3.24: Flowchart of the routine PUTCHAR



is read until it is. PUTCHAR gets the data from the stack and sends it out the secondary serial port. Execution then returns to the calling routine.

#### **QK.FFT:**

The routine QK.FFT samples the voltage and current waveforms for one cycle, performs an FFT on them and sends the results to the host computer. This allows a quick check of line conditions as well as being a method of testing the operation of the harmonic analyzer. No registers are changed by this routine. Figure 3.25 shows the operation of QK.FFT.

The following description details the operation of QK.FFT. QK.FFT first saves the contents of registers that are altered by this routine so that they can be recovered when exiting the routine. It then samples the data using the routine GET.DATA which samples the voltage and current waveforms for one cycle. It performs an FFT on the voltage and current data by calling the routine FFT. The data is then sent out the primary serial port. The data is sent by placing a pointer to the beginning of the voltage data. The byte pointed to is transmitted via the routine OUT. The pointer is incremented to point to the next byte. Since the current data follows immediately after the voltage data (see the memory map in Appendix D) this can continue until the



Figure 3.25: Flowchart for the routine QK.FFT

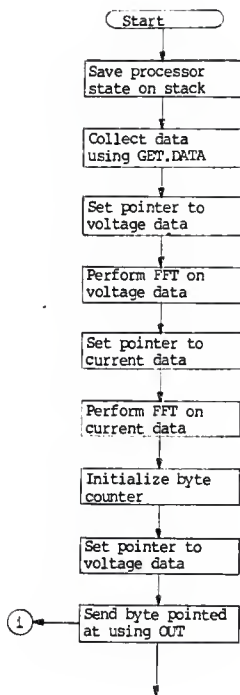
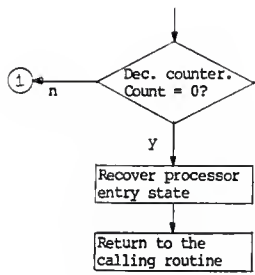


Figure 3.25: Flowchart for the routine QK.FFT



end of the current data. Transmission ends at the end of the block of current data.

This concludes the functions of QK.FFT. The program then goes into the return phase. The registers that were altered in this routine are recovered from the stack and placed in their original condition. Execution then returns to ARDVARK through the entry node entry.pnt. For this reason QK.FFT is not a true subroutine.

#### **SAMPLE:**

SAMPLE controls the acquisition and storage of data by the harmonic analyzer. It checks to see if sampling should occur. If it should then the data is collected, an FFT performed and the results are stored in memory. The time is passed to this routine in the low word of d7. Figure 3.26 shows a flow chart of SAMPLE.

SAMPLE first saves the contents of registers that will be altered on the stack. The variable SYS.GO is then checked to see if sampling is enabled. This is followed by a check of the sample counter. If sampling is not enabled or the sample counter is equal to 168 then SAMPLE goes to the exit phase, otherwise it samples the voltage and current waveforms by calling the routine GET.DATA. It then performs an FFT on the voltage data followed by a FFT on the current

Figure 3.26: Flowchart of the routine SAMPLE

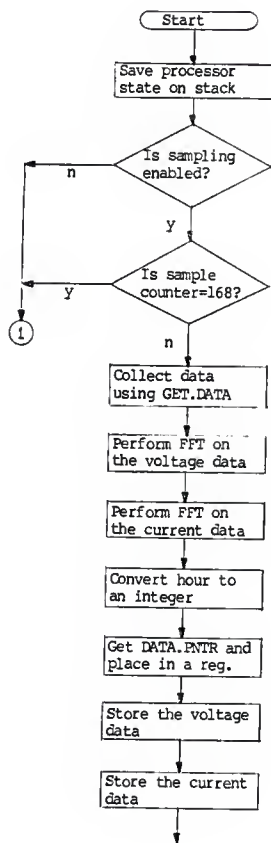
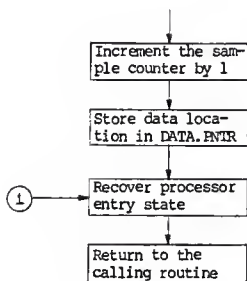


Figure 3.26: Flowchart of the routine SAMPLE



data by calling the routine FFT.

The clock data that was passed to SAMPLE is then converted from a set of BCD numbers representing hours and minutes to a hexadecimal number representing just the hours. This is done by shifting the time data right eight bits. This puts the BCD coded tens and units of hours in the low byte of a register. The nybble for the tens of hours is multiplied by ten to produce its hexadecimal equivalent. The nybble representing the units of hours is then added to this. The results in a hexadecimal word length integer that gives the hour when the data was taken.

SAMPLE stores all the data next. The data pointer is fetched from the system variable data.pntr. The time integer is then stored followed by the voltage data and current data. Not all of the harmonics are stored. The even harmonics from 24 to 62 are left aside. These harmonics are generally negligible and omitting them allowed longer data collection runs without increasing the amount of memory required. The new value of the data pointer resulting from the storage of the data is placed back in the variable data.pntr.

The contents of the altered registers are returned to their entry state by recovering these values from the stack where they were saved. Program flow then returns to ARDVAR

through the node entry.pnt.

#### **SEND.DATA:**

SEND.DATA sends out the data accumulated by the harmonic analyzer to the host computer. The data is sent out as word length integers in Intel format. The least significant byte of the integer is sent first followed by the most significant byte. No registers are altered by this routine. Figure 3.27 shows the operation of SEND.DATA in flowchart form. The following paragraphs detail the functioning of SEND.DATA.

SEND.DATA sends the data as packets of related information. Each sample block, or data packet, is made up of the hour and the harmonic information for that hour for both the current and voltage waveforms. The time is a word length integer that gives the number of the hour the data was taken. The harmonics stored are 0-23 and 25-63 odd. The harmonic information itself is in integer form. Each harmonic takes up 4 bytes of storage, a word integer for the real part and a word integer for the imaginary part. Figure 3.28 shows the data block in graphic form. The length of each block is 177 words.

A checksum is appended to the end of this packet for

Figure 3.27: Flowchart of the routine SEND.DATA

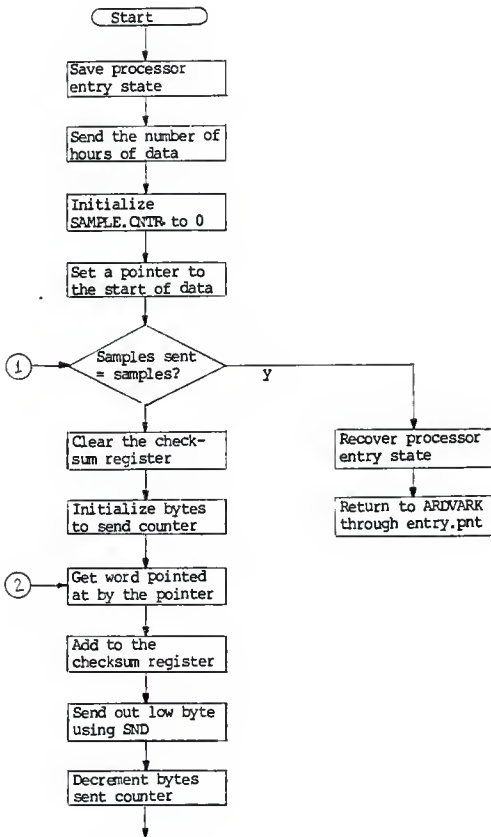
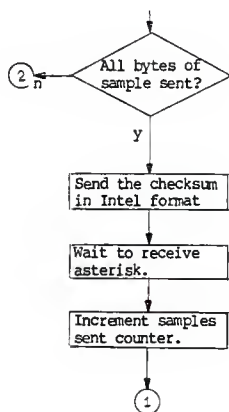




Figure 3.27: Flowchart of the routine SEND.DATA



error checking purposes. The checksum is a word length integer. It is computed by adding each byte in the data packet modulo 65535. The transmission of the checksum is the final action in sending the data packet.

The I/O performed in this routine is nonstandard for the primary serial port. Instead of using the standard routines for serial I/O, IN and OUT, SEND.DATA uses the routines SND and RCV. These routines do not perform the echoing that IN and OUT do. This enables the data to be sent at a faster rate. The listings for SND and RCV are included

Hour
Real part of the 0th voltage harmonic
Imaginary part of the 0th voltage harmonic
Real part of the 1st voltage harmonic
Imaginary part of the 1st voltage harmonic
⋮
Real part of the 63rd voltage harmonic
Imaginary part of the 63rd voltage harmonic
Real part of the 0th current harmonic
Imaginary part of the 0th current harmonic
⋮
Real part of the 63rd current harmonic
Imaginary part of the 63rd current harmonic

Figure 3.28: Data storage format for each hour of data in the harmonic analyzer and format of data transmission packet.

with those of SEND.DATA in Appendix B. The functional descriptions of SND and RCV are the same as those for PUTCHAR and GETCHAR except that SND and RCV deal with the primary serial port instead of the second.

SEND.DATA operates in the following manner. SEND.DATA first saves the registers that are altered in the routine on the stack. The variable sample.cnt is read and sent out as an integer. It then clears the counter for the number of samples sent. The data pointer is then set to the beginning of the data storage block of memory. SEND.DATA then enters a cyclic state to send out the samples.

Each cycle begins with a check of the number of samples sent. If this equals the number of samples stored in the harmonic analyzer then SEND.DATA enters the exit mode covered below. If the number of samples sent does not equal the number of samples stored then the following actions occur. The checksum register is cleared and the 'words to send' counter is initialized. The word of data pointed to by the data pointer is fetched and the data pointer is incremented. This value is added into the checksum register. The low byte is sent followed by the high byte of the word. This is done using the routine snd. If the end of the data block is not reached then the next word is fetched and action is continued until the end of the data block is reached. When

the end occurs the accumulated checksum of the block is sent out, least significant byte first. The harmonic analyzer now waits for the reception of an asterisk to signal that the packet has been received. When this is received the cycle begins again until all the samples have been sent.

SEND.DATA then recovers the contents of registers that were changed by the routine from the stack. Execution then returns to ARDVARK through entry.pnt.

#### **SEND.TIME:**

The routine SEND.TIME sends the time read from the harmonic analyzers clock out the primary serial port in ASCII coded form. The time is sent as four characters representing hours and minutes. No registers are changed by this routine. Figure 3.29 shows the functioning of this routine.

As in most of the ARDVARK subroutines SEND.TIME places the contents of the registers that will be altered by the routine on the stack so the original contents can be recovered when the routine is concluded. SEND.TIME then reads the data from the clock using the routine GET.TIME and saves it in a storage register.

The time is then copied back into the I/O register and shifted right. This moves the most significant nybble in the

Figure 3.29: Flowchart for the routine SEND.TIME

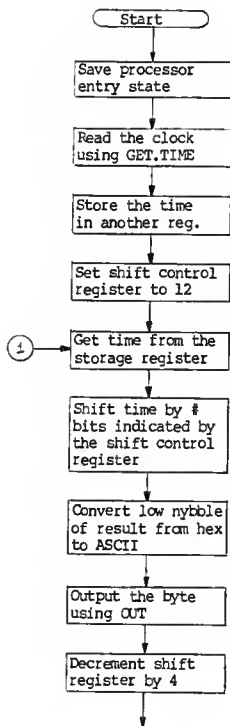
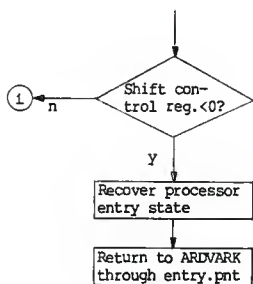


Figure 3.29: Flowchart for the routine SEND.TIME



place of the least significant nybble. This is then converted to ASCII and sent out using OUT. The shift control register is decremented so that the next most significant nybble will be shifted to the low nybble location. This cycle continues until all four nybbles have been sent.

SEND.TIME then recovers the entry state of any registers changed in the routine. Execution returns to ARDVARK through the entry point entry.pnt.

#### **TRACE:**

TRACE allows the user of the harmonic analyzer to single step through programs. It requires the use of a terminal attached to the secondary serial port. It is not really a subroutine attached to ARDVARK. Instead it is an exception routine that is called every time an instruction is executed. This requires that the trace enable bit of the microprocessor be active. This is taken care of in TRACE itself.

Using TRACE is very simple. In order to enter the trace mode a "T" is sent to the harmonic analyzer via the primary serial port. This causes the execution of a trap to the trace exception. This causes the contents of all the processor registers to be displayed on the screen. The message "trace off" is then displayed. A character must be entered

to continue. If the character is a carriage return (CR) then the state of the trace bit remains the same. If any other character is entered then the trace bit is toggled. When first entering the TRACE routine enter any character but CR to get into the "trace on" mode. Once a character has been entered the message "pc:" will appear on the next line. This allows the alteration of the program counter. Type the address of the next statement to execute. If the first character entered is a CR then the current value of the program counter is used. No checking is made for odd addresses. Once the address has been entered the next instruction is executed. If the trace bit is enabled then the trace is returned to and the results are the same as above except that the message displayed after the registers is "trace on." If the trace bit is off execution continues in the normal fashion.

The exception trace, shown in Figure 3.30, operates in the following way. All the processor registers are saved onto the stack. A pointer is set to the top of the stack. The registers are then copied into the I/O register, d7, and displayed using the routine DISP.REG. A label is displayed for the register using the routine DISP.STRING. This continues until all the registers have been displayed.

The next section of trace controls the toggling of the



Figure 3.30: Flowchart for the exception TRACE

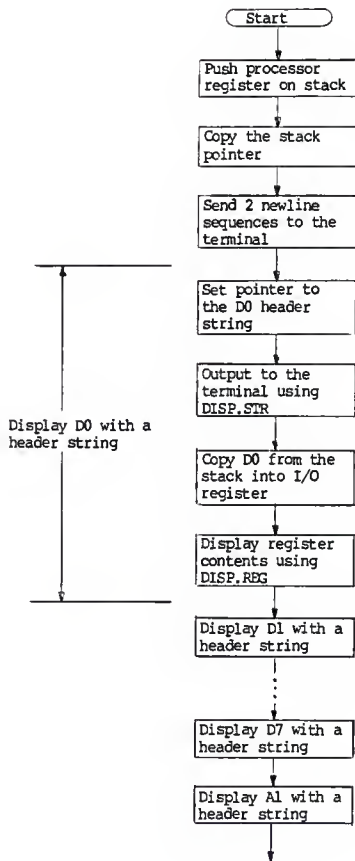


Figure 3.30: Flowchart for the exception TRACE

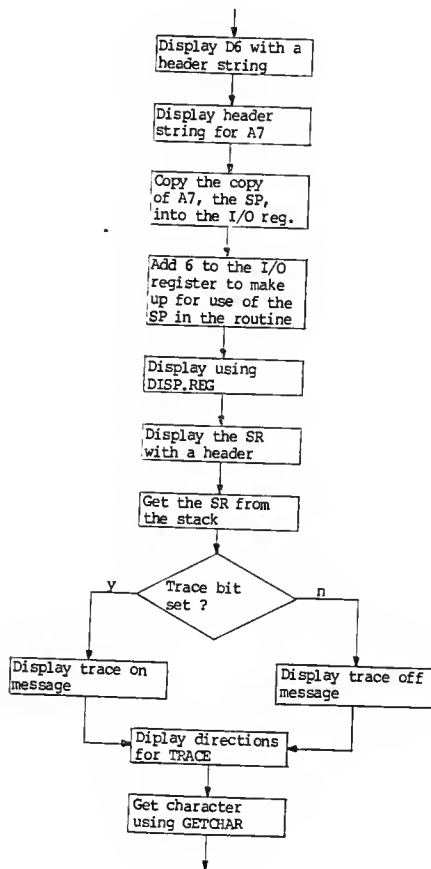


Figure 3.30: Flowchart for the exception TRACE

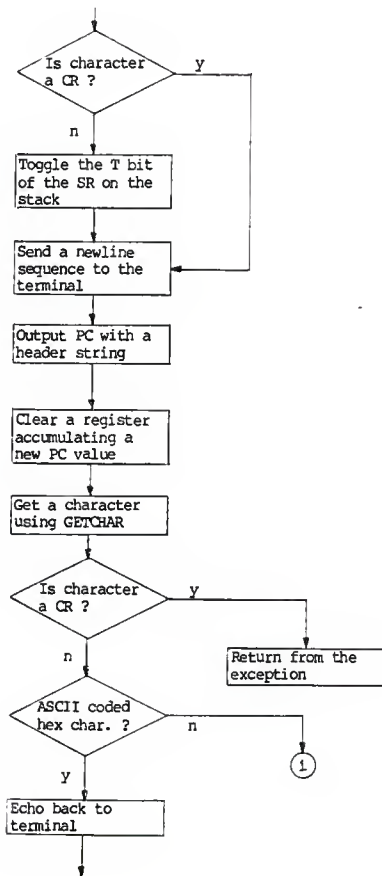
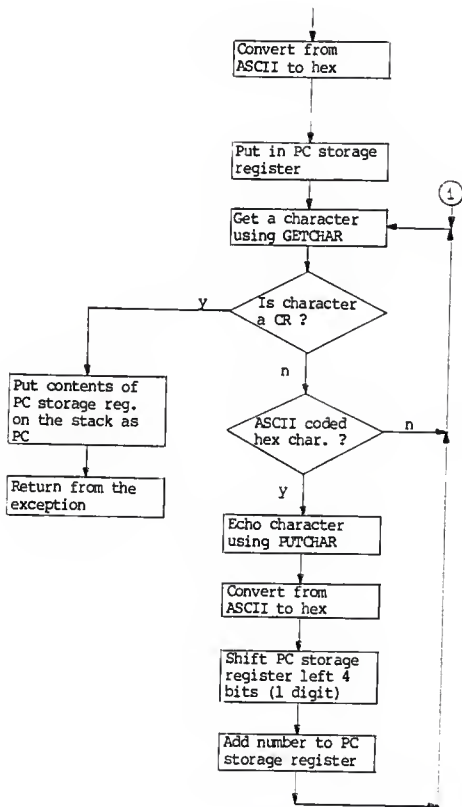


Figure 3.30: Flowchart for the exception TRACE



trace bit. First the status of the trace bit is checked and depending on the state the message "trace on" or "trace off" is displayed using the routine DISP.STRING. Next a character is fetched from the secondary serial port using GETCHAR. If the character returned is not a CR then the trace bit is toggled in the copy of the status register stored on the stack.

The next section of TRACE deals with modifying the program counter. A new line sequence is sent out using the routine DISP.STRING. The label string for the program counter is displayed using DISP.STRING. A character is fetched using the routine GETCHAR. If this character is a CR then execution proceeds to the return phase of TRACE. If it is not a CR then it is checked to see if it is hex. If so it is echoed back to the screen, converted to hexadecimal and added into the address storage register. If the character is not a hexadecimal equivalent then no echo is sent to the terminal and a new character is fetched. Each succeeding character is given a similar treatment except that the storage register is shifted four bits left before the succeeding values are added in. This clears the low nybble and places the value just received in the least significant position of the address. This continues until a CR is received. Trace then goes into the return phase.

The return phase of TRACE clears two more lines on the screen using DISP.STRING. The registers are then read back from the stack. A return from exception is then executed and TRACE goes back to execute the instruction pointed at by the program counter. If the trace bit is set this causes a trace exception to be executed again. If the trace bit is clear then execution proceeds in the usual fashion.

## CHAPTER 4 : TESTING

### Testing Procedure

A testing procedure for the harmonic analyzer was developed to evaluate its performance. Due to the fact that spectrally pure 5A, 120V sources are difficult to come by, the testing procedure used bypassed the sensors for the harmonic analyzer. 5A, 120V sources were used to test the sensors in a more general sort of way but the input signals were not predictable enough to be considered adequate tests. The testing procedure used is outlined in the following paragraphs.

The tests were performed with the following physical setup. The input waveforms were obtained from an HP 3314A synthesizer in the square wave and triangle wave modes. The input signal is fed into the input of the voltage sensor's sample and hold. This bypasses the resistive divider that scales the actual line voltages down. This should not significantly change the results of the test.

The harmonic analyzer is then connected to the Zenith Z-171 microcomputer which is used to monitor the harmonic analyzer's operation and gather data. The Z-171 is running SETUP, the harmonic analyzer/Z-171 communications program.

The Z-171 is then used to cause the harmonic analyzer to perform an FFT on the waveform and send the data on the spectrum to the Z-171.

The two tests performed are the square wave test and the triangle wave test. For the square wave test a 10.0 Vpp 60.0 Hz square wave with no DC offset is used as the input signal. In the triangle wave test a 10.0 Vpp, 60.0 Hz triangle wave is used as the input signal. The output data on the spectrum of the input signals are then compared with ideal values for the appropriate waveform's spectrum.

The results of both tests for five trials each, are shown in Table 4.1 and Table 4.2. The values shown are in percentages of the amplitude of the fundamental frequency, 60.0 Hz. The sampled data was averaged and the maximum negative and positive deviations were noted. This was performed for the first 14 harmonics of 60.0 Hz in both the square wave and the triangle wave tests.

As can be seen from the tables of results the harmonic analyzer is generally accurate to about 0.1%. The error increases compared to the magnitude of the harmonic as the order of the harmonic increases. This is due to the quantization noise inherent in the analog to digital conversion. The signal to noise ratios for lower level signals are then less. In the waveforms used in the test



the amplitude of the various harmonics decreases with frequency, so the error increases.

One of the design goals for the harmonic analyzer project was to achieve an accuracy of better than 1% preferably to within 0.1%. Based on the bench mark tests this design goal was met.

**Table 4.1 Results of Square Wave Test**

harmonic	ideal % of fundamental	avg. % of fundamental	+ deviation	- deviation
1	100	100	0	0
2	0	0	0	0
3	33.3	33.4	0.22	-0.22
4	0	0	0	0
5	20.0	20.06	0.09	-0.143
6	0	0	0	0
7	14.3	14.34	0.053	-0.064
8	0	0	0	0
9	11.1	11.23	0.067	-0.064
10	0	0.016	0.062	-0.016
11	9.09	9.23	0.061	-0.061
12	0	0	0	0
13	7.69	7.84	0.055	-0.060
14	0	0	0	0

**Table 4.2 Results of Triangle Wave Test**

harmonic	ideal % of fundamental	avg. % of fundamental	+ deviation	- deviation
1	100	100	0	0
2	0	0.184	0.0025	-0.0046
3	11.11	11.23	0.207	-0.292
4	0	0	0	0
5	4.00	4.01	0.0603	-0.820
6	0	0	0	0
7	2.04	2.09	0.046	-0.033
8	0	0	0	0
9	1.23	1.22	0.012	-0.0097
10	0	0	0	0
11	0.83	0.846	0.034	-0.028
12	0	0	0	0
13	0.58	0.584	0.011	-0.024
14	0	0	0	0

## CHAPTER 5 : FURTHER CONSIDERATIONS AND CONCLUSIONS

### Further Considerations

At this point a number of future considerations and improvements can be mentioned. These can be grouped into the areas of basic system changes, hardware improvements and software improvements.

The first basic system change is to add the subsystems necessary to monitor three phases of voltage and current. This would enable the analyzer to monitor the relations between harmonics on different phases. Also in order to avoid aliasing problems an anti-aliasing filter should be added. The software system could be modified by adding some self test diagnostic routines that would test some of the harmonic analyzers systems.

There are numerous hardware improvements that could be done. The most important is to redesign the board to eliminate the noise problems in the digital section. A future design could possibly use a multilayer board with a ground plane in conjunction with better layout to reduce noise. The microprocessor clock could be separated from the sampling clock system so the microprocessor could be run at a higher speed.

Many of the chips used in the current version of the harmonic analyzer could be replaced by newer more efficient and powerful chips. The 6800 style timing used to interface with these slower peripherals could be eliminated with the newer faster chips. Some of the improvements in this area are to replace the RAM, serial ports and parallel port. The use of 43256 32kx8 RAM's, 2681 dual serial ports would produce lower chip counts and more functionality than the present chips.

The use of custom chips could significantly reduce the chip count of the harmonic analyzer. The decoding logic used for chip selects is presently implemented by discrete chips. The use of programmable array logic (PAL) chips could reduce the number of chips needed for these functions to about two. Additionally PAL's could be used to control the action of the sampling port.

The software changes mainly deal with improving the signal processing routines. The present FFT could be used to perform a Walsh technique analysis of the data. In this the incoming data would be sampled over a number of cycles,  $n$ , the FFT of this data is computed in one cycle segment and averaged. This reduces the variance of the output sequence. The reduction is  $walsh = fft/n$ . For more information on

this technique see [6]. This technique can not be performed presently due to lack of buffer memory on the harmonic analyzer.

These improvements would increase the overall system performance. They should increase the functionality to board space ratio and provide a faster more accurate and reliable system.

### **Conclusions**

A low cost spectrum analyzer for use in measuring power system harmonics has been developed. The analyzer will measure current and voltage harmonics each hour for one week. The data acquired by the analyzer can be gathered by an MS-DOS compatible computer. The accuracy of the analyzer compares to what was expected based on benchmark tests.

The harmonic analyzer does have a number of problems. The current hardware model appears to have some noise problems in the digital section. This reduces system reliability. It did not affect the project too significantly but would not be acceptable in a commercial machine. Also there exists the possibility of aliasing in sampling the input waveforms. This would decrease the accuracy of the analyzer if it were significant.

This document discussed the hardware and software systems that make up the harmonic analyzer. The hardware was discussed at the component level. Some of the decisions that were made in deciding on the form of the hardware are also discussed. The software system was examined on a routine by routine basis at the flowchart level to explain the methodology used.

## APPENDIX A

Parts list and memory map for the Harmonic Analyzer. The schematics for the Harmonic Analyzer are included in Chapter 2.

## Parts List for the Harmonic Analyzer

### Integrated Circuits

U1	: 741 operational amplifier
U2	: AD524 precision instrumentation amplifier Analog Devices
U3, U4	: SMP-10 Sample and Hold Amplifiers Precision Monolithics Inc.
U5	: MUX-08 eight channel multiplexer Precision Monolithics Inc.
U6	: LM310 voltage follower National Semiconductor
U7	: 4.9152 Mhz TTL level clock generator
U8, U9	: 74HC74 Dual D flip/flops
U10	: CD4520 Dual 4 bit counter
U11	: CD4017 Decade Sequencer/Divider
U12	: 74HC154 4 to 16 decoder, skinny package
U13	: 68000 Microprocessor
U14, U23, U32, U33	: 74HC32 Quad two input OR
U15, U16, U18	: 74HC04 Hex inverter
U17	: 74HC175 Quad D flip/flop
U19	: 555 Timer
U20	: LM393 Dual comparator
U21, U22	: 2732 4Kx8 EPROM
U24-U31	: 6264 8Kx8 static RAM
U34	: 74HC138 3 to 8 decoder
U35	: 74HC20 Dual four input NAND



U36, U37 : 6850 Asynchronous Comm. Interface Adapter  
 U38 : 1489 RS-232 to TTL quad receiver  
 U39 : 1488 TTL to RS-232 quad transmitter  
 U40 : 6821 Peripheral Interface Adapter (PIA)  
 U41 : 74HC08 Quad two input AND  
 U42 : AD574 12 bit Analog to Digital Converter  
 Analog Devices  
 U43 : MM58274 Real Time Clock  
 National Semiconductor

### Resistors

R1 : 18K 0.25 W  
 R2 : 500 ohm multiturn trimmer potentiometer  
 R3, R4, R12,  
 R20, R23 : 10K multiturn trimmer potentiometer  
 R5 : 205 ohm 1% 0.25W metal film  
 R6, R10 : 2.4K 0.25W  
 R7, R11 : 100K multiturn trimmer potentiometer  
 R8 : 90.9K 1% metal film 0.5W  
 R9 : 9.09K 1% metal film 0.5W  
 R13, R14 : 4.7K 0.25W  
 R15 : 2.0M 0.25W  
 R16 : 10.0M 0.25W  
 R17, R21,  
 R22 : 1.0K 0.25W  
 R18, R19 : 100 ohm multiturn trimmer pots

## Capacitors

C1-C5, C7,  
C11, C13,  
C15-C20,  
C23-C28,  
C31-C40 : 0.1uF ceramic disk capacitors

C6, C8,  
C12, C14 : 10uF tantulum capacitors

C9, C10 : 5000pF polystyrene or teflon capacitor

C21, C22 : 1.0uF mylar capacitor

C29 : 20 pF cermaic disk

C30 : 5-30pF trimmer capacitor

## Transistors

Q1 : BS170 N-channel enhancement mode MOSFET  
Q2, Q3, Q4 : 2N2222A

## Diodes

D1 : LM329 precision zener  
D2 : 10 V transorb or two back to back 10 V  
zeners

D3, D6 : 3.9V 1.0W Zener

D4 : 3.6V 1.0W Zener

D5, D7 : Low to Med. power general purpose Si diode

## Miscellaneous

J1, J2 : 25 pin subminiature D plug  
SW1 : 7 Pole DIP Switch  
SW2 : 4 Pole DIP Switch  
CT1 : PI-100 Hall Effect Current Sensor  
F. W. Bell  
X1 : 32.786 Khz crystal  
T1 : 10V Transorb  
SIP1-SIP8 : 4.7k Common node SIP resistors  
Panasonic

## Memory Map for the Harmonic Analyzer

	Address:	Comment:
	-----	-----
ROM:	00 0000 - 00 03FF	68000 Vector Table
	00 0400 - 00 1FFF	Ardvark Monitor Program
	00 2000 - 00 FFFF	Reflections of ROM
I/O:	01 0000 - 01 FFFF	Memory Block with 6800 timing specifications
	01 1001	ACIA #1 Status register
	01 1003	ACIA #1 Data Register
	01 2001	ACIA #2 Status Register
	01 2003	ACIA #2 Data Register
	01 F001	PIA Data/Dir. Register A
	01 F003	PIA Control Register A
	01 F005	PIA Data/Dir. Register B
	01 F007	PIA Control Register B
RAM:	02 0000 - 02 077F	System Stack 1920 bytes
	02 0780 - 02 07FF	System Variables 128 bytes
	02 0800 - 02 09FF	Temporary storage block for voltage data
	02 0A00 - 02 0BFF	Temporary storage block for current data
	02 0C00 - 02 FFFF	Memory Block for Storage of Harmonic Data 61 Kbytes

## APPENDIX B

Source code listings for ARDVARK, the harmonic analyzer's control program

```

*****
*
*      ardvark.lnk                {link specification}
*
*      Link specification for the harmonic analyzer's
*      monitor.
*
*      source file:   ardvark.lnk
*
*      author:       Norm Mortensen
*
*      data:         30Apr87
*
*****

```

```

          output motorola
pc          equ.l $400
data.pntr  equ.l $20780
sample.cntr equ.l $20784
sys.go     equ.l $20786
data.start equ.l $20C00

```

```

org pc
link vectors      this module assembles absolutely at $00
link ardvark      these link relocatably from pc up
link in
link out
link getchar
link get_data
link putchar
link dispreg
link dispstr
link asciihex
link hexascii
link hexcheck
link clk_read
link clk_wrt
link get_time
link fft
link tablel28
link download
link memory
link sample
link init
link snd_data
link snd_time
link go
link qk_fft
link trace        this compiles absolutely at $1000

```

```

*****
*
*      ardvark                                {program}
*
*      Main control routine for the harmonic analyzer.
*      It performs system reset on power up and provides
*      control for the various harmonic analyzer activities.
*
*      source file:   ardvark.a68
*
*      author:        Norm Mortensen
*
*      date:          28Apr87
*
*****

```

```

*      global label definitions
*      xdef entry.pnt

```

```

*      external references
*      xref download
*      xref get.time
*      xref go
*      xref in
*      xref initialize
*      xref memory
*      xref pc
*      xref putchar
*      xref qk.fft
*      xref send.time
*      xref sample
*      xref send.data
*      xref set.clock
*      xref sys.go

```

```

*      label definitions

```

```

stk          equ    $02fffe
sreg1        equ    $011001          acia 1 reg locations
dreg1        equ    $011003
sreg2        equ    $012001          acia 2 reg locations
dreg2        equ    $012003
a.status     equ    $01f003          a side of he pia
a.data       equ    $01f001
b.status     equ    $01f007          b side of acia
b.data       equ    $01f005

```

```

*
*      system reset and initialization
*

acia.reset      move.b  #3,sreg1      master reset of acia'S
                move.b  #3,sreg2
                move.b  #$12,sreg1    configure acia 1
                move.b  #2,sreg2      configure acia 2

pia.reset       move.b  #$30,a.status
                move.b  #0,a.data
                move.b  #$34,a.status
                move.b  #$00,b.status
                move.b  #$0f,b.data
                move.b  #$04,b.status
                move.b  #$0f,b.data

entry.pnt:
*
*      Check for char from serial port one
                move.b  sreg1,d7      get status
                andi  #1,d7          check for char ready
                beq  check.clock      if zero no char

*
*      Get data and echo back to source
                move.b  dreg1,d7      get data
                andi  #$ff,d7        mask off all but data
                move  d7,-(sp)        save

echo.rdy:      move.b  sreg1,d7      check for empty tx reg
                andi  #2,d7
                beq  echo.rdy
                move  (sp)+,d7
                move.b  d7,dreg1      send echo
                bsr  putchar         display on terminal

*
*      Decode the input character and route
                flow accordingly

                cmpi.b  #'M',d7      if m then memory
                beq  memory          examine/alter routine
                cmpi.b  #'m',d7
                beq  memory

                cmpi.b  #'G',d7      if x then send off
                beq  send.data        data routine
                cmpi.b  #'g',d7
                beq  send.data

                cmpi.b  #'X',d7      if g then program
                beq  go               execution routine

```



```

    cmpi.b #'x',d7
    beq go

    cmpi.b #'D',d7      if d then the file
    beq download        downloading routine
    cmpi.b #'d',d7
    beq download

    cmpi.b #'T',d7      if t then the instruction
    beq tracer          trace routine
    cmpi.b #'t',d7
    beq tracer

    cmpi.b #'F',d7
    beq qk.fft          get data and run an fft
    cmpi.b #'f',d7      on it
    beq qk.fft

    cmpi.b #'I',d7      harmonic analyzer init
    beq initialize      routine
    cmpi.b #'i',d7
    beq initialize

    cmpi.b #'B',d7      set sys flag to okay
    beq enable.sampling sampling
    cmpi.b #'b',d7
    beq enable.sampling

    cmpi.b #'R',d7      read clock and send out
    beq send.time
    cmpi.b #'r',d7
    beq send.time

*
*check.clock          Debugging code
bra entry.pnt

*
check.clock          Check to see if time to sample
bsr get.time         get time
cmpi #0,d7           check for midnight
bne skip.over
move #$ffff,sys.go  enable sampling

*
skip.over:          Check for on tens of seconds
andi #$ff,d7        if min=0 then sample
bne entry.pnt

*
wait:               If time for sample wait to prevent overrun
bsr get.time

```

```

        andi #$f,d7
        beq wait

*       Get time and call sampling routine
        bsr get.time
        bra sample

*       For debugging purposes
        bra entry.pnt

*-----
*
*       tracer:
*
*       instruction trace calling routine

tracer:      trap #0           force execution of the trace
*                               exception

*-----
*
*       enable.sampling:
*
*       Sets the flag SYS.GO to enable sampling

enable.sampling:
        move #$ffff,sys.go
        bra entry.pnt

```

```

*****
*
*      ascii.hex :                               {subroutine}
*
*      This routine converts an ascii byte in d7 to a hex
*      nibble in d7. No other registers are altered. Illegal
*      data flagged with a return of $FF.
*
*
*      source file : asciihex.a68
*
*
*      Revisions:
*
*      -----
*      0.00          Norm Mortensen          28 Jan 87
*
*****

*      Entry point
*      xdef  ascii.hex          * declare global variable for
*                               * the linker

ascii.hex:
    movem.l a0-a1,-(sp)        * save the entry status
    lea    table(pc),a0       * point a0 to data table
    andi.l #$ff,d7            * clear all but low byte
    andi   #$1e,CCR           * clear the carry
    subi   #$30,d7            * test for under range
    bcc    good.char          * and remove offset
    cmpi.b #$17,d7            * check for over range
    bcs    good.char

bad.input:
    move.l #$ff,d7            * return error symbol
    movem.l (sp)+,a0-a1       * return to entry state
    rts                       * go home!

good.char:
    move.l d7,a1              * put offset into pointer
    move.b (a0,a1),d7         * get hex value
    andi.l #$f,d7             * save only low nybble
    movem.l (sp)+,a0-a1       * return to entry status
    rts                       * go home!

table:
    dc.b 0,1,2,3,4,5,6,7,8,9 * hex look up table
    dc.b $ff,$ff,$ff,$ff,$ff,$ff,$ff,$ff
    dc.b $a,$b,$c,$d,$e,$f,$ff

```

```

*****
*
*      clock.read                                {subroutine}
*
*      Primitve routine to read the contents of a register
*      of the real time clock. The number of the register to
*      be read is placed the low nibble of d7. The contents
*      of the register are returned in the low nibble of d7.
*      No other register contents are altered.
*
*      source file : clk_read.a68
*
*      revisions:
*      _____
*      1.0   28Apr87           Norm Mortensen
*
*****

*      global definitions
*      xdef clock.read

*      internal definitions
*      a.status: equ $1f003
*      a.data:   equ $1f001
*      b.status: equ $1f007
*      b.data:   equ $1f005

clock.read:

*      configure pia for read
*      move.b #$30,a.status    get to direction reg
*      move.b #$0f,a.data      0-3:outputs, 4-7:inputs
*      move.b #$34,a.status    get back to the data reg

*      put address on lines
*      move.b d7,a.data

*      run strobes for read cycle
*      move.b #$e,b.data       activate clk select
*      move.b #$c,b.data       activate *cs, *rd
*      nop
*      move.b a.data,d7        get data
*      move.b #$e,b.data       only *cs active
*      move.b #$f,b.data       all inactive, read done

*      position data
*      asr #4,d7              slide data to low nibble

```

```
andi #$f,d7           mask off the rest of the reg

*   make pia all inputs
    move.b #$30,a.status  get to direction reg
    move.b #0,a.data      all lines inputs
    move.b #$34,a.status  get back data reg

*   go back home
    rts
```

```

*****
*
*      clock.write                      { subroutine}
*
*      Primitive routine to allow writing to the registers of
*      the real time clock. The data is in bits 7-4 of d7
*      the register number is in bits 3-0. No register contents
*      are changed by this routine.
*
*      source file : clk_wrt.a68
*
*      revisions:
*
*      _____
*      1.0   28Apr87       Norm Mortensen
*
*****

```

```

*****

```

```

*      global definitions
*      xdef clock.write
*      internal definitions
*      a.status:      equ $1f003      use byte addressing mode
*      a.data:        equ $1f001      only with these labels
*      b.status:      equ $1f007
*      b.data:        equ $1f005

```

```

clock.write:

```

```

*      Configure the pia for a clock write
*      move.b #$30,a.status      get to data direction reg
*      move.b #$ff,a.data        all bits outputs
*      move.b #$34,a.status      back to data reg
*
*      Put data and address on line
*      move.b d7,a.data
*
*      Go thru right sequence for the clock
*      move.b #$e,b.data         activate clock select
*      move.b #$a,b.data         activate select and wrt strobes
*      nop                       give clock time
*      move.b #$e,b.data         only clk select active
*      move.b #$f,b.data         all inactive write complete
*
*      Make the pia all inputs and exit
*      move.b #$30,a.status      get to data direction reg
*      move.b #0,a.data          all bits inputs
*      move.b #34,a.status       back to data reg
*      rts

```

```

*****
*
*      disp.reg                                {subroutine}
*
*      This routine displays the contents of d7 on the the
*      terminal device. The output shows the register's
*      contents in hexadecimal form. No registers are
*      changed in this routine
*
*      source file : dispreg.a68
*
*      revisions:
*
*      _____
*      0.00          Norm Mortensen          23 Jan 87
*
*****

```

```

*      define global label
*      xdef disp.reg

*      external references
*      xref hex.ascii      hex to ascii subroutine
*      xref putchar        display character

*      Internal Symbols
*      CR: equ $0d
*      LF: equ $0a

disp.reg:      movem.l d0-d2,-(sp)      save entry state of uP
               move.l d7,d0          put data in d0
               move.l #$f,d2         character mask
               move    #28,d1        init shift control reg.

nibble.out:   move.l d0,d7           copy data
               asr.l d1,d7          shift a nibble down
               and.l d2,d7         clear off the extra bits
               bsr    hex.ascii     convert to ascii code
               bsr    putchar       put out serial port
               subi   #4,d1         dec. shift amount by 4
               bcc   nibble.out     if shift > 0 then cont.
               move.b #CR,d7       close up the line with
               bsr    putchar       a carriage return and a
               move.b #LF,d7       linefeed
               bsr    putchar
               move.l d0,d7         put registers in order
               movem.l (sp)+,d0-d2  get back entry status
               rts

```

```

*****
*
*      disp.string:                                {subroutine}
*
*      disp.string displays a string on the output console
*      device. A pointer to the string to be displayed is
*      passed in register d7. The string should be stored in
*      increasing sequential memory positions. The end of
*      string is signaled with an ascii NULL. No registers
*      are changed.
*
*      source file : dispstr.a68
*
*      revisions:
*
*      _____
*      0.00          Norm Mortensen          30 Jan 87
*
*
*****

```

```

*      Entry point
*      xdef    disp.string

```

```

*      External subroutine references
*      xref    putchar

```

```

disp.string:      movem.l d0/d7/a0,-(sp)  * save the entry status
                  move.l  d7,a0        * put string pointer in a0

output.chars:    move.b  (a0)+,d7
                  beq     end.of.string * if an ascii null quit
                  bsr     putchar      * otherwise send out
                  bra     output.chars  * go for next char

end.of.string:   movem.l (sp)+,d0/d7/a0 * return to entry state
                  rts                * go home!

```



```

*****
*
*       download
*
*       this routine downloads files from the North Star to the
*       harmonic analyzer. The files must be in the form of
*       Motorola S records in order to be processed properly
*       by the harmonic analyzer.
*
*       source file :   download
*
*       language   :   Avocet 68000 assembler
*
*       revisions:
*
*       _____
*       1.00   6Mar87   Norman Mortensen
*
*****

*       global label definitions
*       xdef download

*       external references
*       xref ascii.hex
*       xref entry.pnt
*       xref in

*       internal label definitions
ram.top      equ $27fffe

download:    bsr in
             cmpi.b #'S',d7          search until 'S' is found
             bne download
             bsr in                   get character after 'S'
             cmpi.b #'9',d7
             beq s9.line
             cmpi.b #'2',d7
             beq s2.line              then decode as s2 line
             cmpi.b #'1',d7          if s1 then fall thru
             bne download            try again

*
*       register assignmnets for S1 line decoding
*
*       d0 : counts number of data btes remaining in line
*       d1 : data and scratchpad register
*       d2 : address character counter
*       a0 : address pointer register
*

```

```

*           get the number of data bytes is a line

byte.count:  bsr in           get 1st count char
              bsr ascii.hex  convert to hex => d7
              asl #4,d7      make it the high digit of count
              move d7,d0     put partial cnt in d0
              bsr in         get 2nd char count
              bsr ascii.hex  convert to hex => d7
              add d7,d0      d0 : counts data bytes in line

*           initialize registers for address decoding
              move.l #0,d1
              move #4,d2

*           decode the next four bytes as the address

address.bytes: bsr in
               bsr ascii.hex
               asl #4,d1
               add d7,d1
               subi #1,d2
               bne address.bytes
               addi.l #$20000,d1    loads program into RAM
               move.l d1,a0        a0 now holds the address
               subi #3,d0          adjust character counter

*           decode the remaining bytes in pairs as hex
*           data in ascii

data.bytes:   move #0,d1          init data byte storage
              bsr in             get 1st char of pair
              bsr ascii.hex      convert to hex => d7
              move d7,d1         partial data => d1
              asl #4,d1          make it the high digit
              bsr in             get 2nd char of the pair
              bsr ascii.hex      convert to hex
              add d7,d1          decoded data => d1
              move.b d1,(a0)+    store and inc addr ptr

*           check to see if still in ram
              cmpa.l #ram.top+2,a0
              bcs addr.good
              movea.l #ram.top,a0  if not pt to top ram
              subi #1,d0           decrement data byte cnt
              bne data.bytes       cont till end of line
              bra download         if eol then go back to
*                                 line decoding

```

```

*
*   register assignments for the s2 line decoding
*
*       d0 : holds the number of data bytes in the line
*       d1 : data register and temp storage
*       d2 : address character counter
*       a0 : the address pointer
*

s2.line:      bsr in
              bsr ascii.hex
              andi #$f,d7
              move d7,d0
              asl #4,d0
              bsr in
              bsr ascii.hex
              andi #$f,d7
              add d7,d0           d0 holds the line byte count

              move.l #0,d1
              move #5,d2

s2.address:   bsr in
              bsr ascii.hex
              andi.l #$f,d7
              asl.l #4,d1
              add.l d7,d1
              dbf d2,s2.address
              move.l d1,a0
              subi #4,d0         remove the counts for the
                                address and checksum bytes
                                from the byte count

*
*
              cmpa.l #ram.top+2,a0
              bcs s2.data
              move.l #ram.top,a0

              subi #1,d0        decrement the byte count
                                so that the looping
                                primitive db can be used

*
*

s2.data:     bsr in
              bsr ascii.hex
              andi.l #$f,d7
              move.l d7,d1
              asl #4,d1

```

```
bsr in
bsr ascii,hex
andi #0x07,d7
add d7,d1
move.b d1,(a0)+

cmpa.l #ram.top+2,a0
bcs s2.address.valid
move.l #ram.top,a0
```

```
s2.address.valid:
    dbf d0,s2.data
    bra download
```

```
s9.line:
    bsr in
    cmpi #0x0a,d7
    bne s9.line
    bra entry.pnt
```

```

*****
*
*      fft                                {subroutine}
*
*      Routine to perform the fast fourier transform on
*      128 samples of data. The pointer to the beginning
*      of the data is passed in a6. The data is stored as
*      follows:
*          #0      real      (2 bytes)
*                imag      (2 bytes)
*          #1      real      (2 bytes)
*                imag      (2 bytes)
*                :
*                :
*          #127   ""        ""
*
*      The data format is complex rectangular fixed point.
*      The results are calculated in place. No registers
*      are altered by the routine.
*
*      source file : fft.a68
*
*      revisions:
*      -----
*      1.1   Norm Mortensen   Feb 87   debugged
*      1.2   "                "      May 87   fixed error in
*                                               the butterfly swaps
*
*****
*
*      define global label
*      xdef fft
*
*      external references
*      xref sintbl
*
*      parameters that arte dependent on the length of the FFT
*
samples      equ      128          number of samples
iterations   equ      7           samples = 2**iterations
cos          equ      64          offset of cos in sin table
*                                               also cos = samples/2
*
*      parameters dependent upon the a/d
*
bits        equ      12           bits of a/d
offset      equ      2048        offset of the a/d

```

```

*      standard parameters

im      equ      2      offset to im part
c.next  equ      4      next complex number
mask    reg      d0-d7/a0-a6      (all registers but sp)

*
*-----*
*
*      Save the initial machine status
fft:    movem.l  mask,-(sp)      save the entry state

*
*      This loop subtracts 2048 from all input values. This
*      converts the data from the a/d to 2's compliment form.
*      It also zeroes the imaginary counterpart.
*
*      move      #samples-1,d0    set the sample counter
*      moveq     #0,d2            clr d2 serve as const 0
*      move.l    a6,a0           a0 pnts to the data
init:    move    (a0),d1        get data
*      subi     #offset,d1      convert to 2's comp
*      move     d1,(a0)+        put back in location
*      move     d2,(a0)+        clr imag part of data
*      dbf      d0,init         if not last sample then
*                               branch to init

*
*
*      perform the number crunching of the FFT
*      based on three nested loops: the iteration loop,
*      the sequence loop and the pnt in sequence loop.
*
*      register assignments are as follows
*      a0 : seq counter
*      a1 : # of sequences reference
*      a2 : data pntr
*      a3 : pntr to the base of the sin table
*      a4 : cos offset in the sin table
*      a5 : half sequence offset
*      a6 : points to the start of data
*      d0 : iteration counter
*      d1 : pnt in sequence counter
*      d2 : number of pnts per sequence /2

```

```

*

*           initialize the registers
moveq    #1,d0           iteration counter
movea.l  #1,a1           # of sequences
move     #samples,d2
asr      #1,d2           # pts per sequence / 2
lea      sintbl(pc),a3   ptr to the sintable

*
iter.loop:  do #iterations times
            move.l  a6,a2           pt to data
            movea.l #0,a0           init seq counter

            move.l  #0,a5           make sure a5 is clr
            move    d2,a5           copy pts.per.seq / 2
            add     a5,a5
            add     a5,a5           half.seq * 4 since data is
*                                           4 bytes. half seq offset

*
sequence.loop:  do # sequences times
               moveq  #0,d1        init pt in seq

*
               do pts in seq / 2 times

*           calculate  $g(x) = m(x) + m(x+N/2)$ 
*           and store the results in m(x)
*           calculate  $h(x) = m(x) - m(x+N/2)$ 
*           and store the results in m(x+N/2)

pt.seq.loop:  move (a2),d4           m(x).re => d4
              move im(a2),d5       m(x).im => d5
              move (a2,a5),d6      m(x+N/2).re =>d6
              move im(a2,a5),d7    m(x+N/2).im =>d7
              add  d4,d6           add re => d6
              add  d5,d7           add im => d7

              cmpi #4,d0           if iter. cntr less
              bcs no.divide.l      then don't divide
              asr  #1,d6           take care of 1/N
              asr  #1,d7           a little each iter

no.divide.l:  move d6,(a2)         store at m(x).re
              move d7,im(a2)      store at m(x).im
              sub  (a2,a5),d4      (m(x)-m(x+N/2)).re

```

```

                                sub  im(a2,a5),d5  (m(x)-m(x+N/2)).im
                                cmpl #4,d0          if iter. cntr less
                                bcs no.divide.2     than 4 don't /2
                                asr  #1,d4          take care of 1/N
                                asr  #1,d5          a little each iter
no.divide.2:                    move d4,(a2,a5)     store re to m(x+N/2)
                                move d5,im(a2,a5)  store im to m(x+N/2)

*                                multiple the resulting h(x) by Wn
*                                get the sin and cos values to calculate
*                                Wn. arg Wn = 2pi/N * (pt in seq) *
*                                2 ** iteration.cntr. The 2pi/N is already
*                                taken into account in the formation of
*                                the sin table.

*                                calculate the offset into sin/cos table
                                move dl,-(sp)       save pt in seq cntr
                                asl  d0,d1         pt.seq **iteration
                                move dl,a4         sin offset => a4
                                move (sp)+,dl      get pt in seq back

*                                get cos(Wn) values
                                move cos(a3,a4),d4
                                move d4,d6         Wn.re in d4,d6

*                                get -sin(Wn) values
                                move (a3,a4),d5    get sin
                                neg  d5           -sin
                                move d5,d7         Wn.im in d5,d7

*                                multiply h(x) * Wn => m(x+N/2)
                                muls (a2,a5),d4    m(x+N/2).re * Wn.re
                                muls im(a2,a5),d5  m(x+N/2).im * Wn.im
                                muls im(a2,a5),d6  m(x+N/2).im * Wn.re
                                muls (a2,a5),d7    m(x+N/2).re * Wn.im
                                sub.l d5,d4       result.re => d4
                                add.l d7,d6       result.im => d6
                                move  #bits-1,d5  set shift index
                                asr.l d5,d4       rescale result.re
                                asr.l d5,d6       rescale result.im
                                move  d4,(a2,a5)  result.re => m(x+N/2)
                                move  d6,im(a2,a5) result.im => m(x+N/2)

*                                loop cntr and data ptr increment
                                addq  #1,d1       inc pt.in.seq
                                addq  #c.next,a2   next data element

*                                Test for end of sequence
                                cmp  dl,d2       cmp pt cntr with # pts

```



```

                                bne pt.seq.loop

*                               Test for the end of the iteration
                                adda a5,a2          pnt to head of nxt seqnce
                                addq #1,a0          increment the seq cntr
                                cmpa a0,a1          compare with # sequences
                                bne sequence.loop    if last seq fall thru

*                               Test for end of the fft calculations
                                add al,al          2 * # sequences
                                asr #1,d2          halve #pts.per.seq
                                addq #1,d0          incthe iter cntr
                                cmpi #iterations+1,d0 check for last iteration
                                bne iter.loop       it.cntr = last iter fall thru

*
*                               Butterfly transfers
*

*                               Register assignments for the butterfly transfers
*                               a0 : data ptr
*                               d0 : offset index to data element
*                               d1 : corresponding bit reverse offset to data ele
*                               d2 : pt sequence counter
*                               d3 : bit swap counter
*                               d4 : copy of pt seq cntr to be bit swapped
*

*                               initialization of parameters
                                move.l a6,a0          pt at the data
                                moveq #0,d2          init cntr

*                               do number of samples times
butterfly:                       move d2,d4          copy cntr to bit swap
                                moveq #0,d1          clr reversed bits reg

*                               do number of bits in index times
bit.swap:                         move #iterations-1,d3  init bit swap counter
                                roxr #1,d4          slide bit out 1 way
                                roxl #1,d1          and slide it in another
                                dbf d3,bit.swap     loop check

                                cmp d2,d1          d1<d2 if carry set. Don't
                                bcs swap.done        swap if d1<d2, already done
                                move.l d2,d0          put cntr in offset reg
                                asl.l #2,d0         *4 since data 4 bytes wide
                                asl.l #2,d1         *4 since data 4 bytes wide
                                move.l (a0,d0),d5    :

```

```

                                move.l (a0,d1),(a0,d0) : swap the data
                                move.l d5,(a0,d1)      :
swap.done:  addq #1,d2                inc pnt cntr
            cmpi #samples,d2         if d2=#samples then done
            bne butterfly

```

\* clean up after the fft

```

movem.l (sp)+,rmask    get back entry status
rts                    go home!

```

```

*****
*
*           a sin/cos lookup table must be linked in at
*           this point in the code
*
*           in order to change the length of the input
*           sequence to the fft the length of the sine
*           table must be changed.
*
*****

```

```

*
*           128 pnt FFT sine table
*

```

```

*           global label definition
*           xdef sintbl

```

```
sintbl:
```

```

dc.w      0,      100,    201,    300
dc.w     399,    497,    594,    690
dc.w     783,    875,    965,   1052
dc.w    1137,   1219,   1299,   1375
dc.w    1447,   1517,   1582,   1644
dc.w    1702,   1756,   1805,   1850
dc.w    1891,   1927,   1959,   1986
dc.w    2008,   2025,   2037,   2045
dc.w    2047,   2045,   2037,   2025
dc.w    2008,   1986,   1959,   1927
dc.w    1891,   1850,   1805,   1756
dc.w    1702,   1644,   1582,   1517
dc.w    1447,   1375,   1299,   1219
dc.w    1137,   1052,   965,    875
dc.w     783,    690,    594,    497
dc.w     399,    300,    201,    100
dc.w      0,    -100,   -201,   -300
dc.w    -399,   -497,   -594,   -690
dc.w    -783,   -875,   -965,  -1052
dc.w   -1137,  -1219,  -1299,  -1375
dc.w   -1447,  -1517,  -1582,  -1644
dc.w   -1702,  -1756,  -1805,  -1850
dc.w   -1891,  -1927,  -1959,  -1986
dc.w   -2008,  -2025,  -2037,  -2045

```

```

*****
*
*      getchar:                                {subroutine}
*
*      gets a one byte character from the second serial port.
*      The character is returned in the low byte of d7.
*
*      source file : getchar.a68
*
*      revisions:
*
*      _____
*      0.00          Norm Mortensen          20 Jan 87
*
*****

```

```

*      Entry point
*
*      xdef    getchar

```

```

*      Internal definitions

```

```

stat.reg2    equ    $012001
data.reg2    equ    $012003

```

```

getchar:
    move.b   stat.reg2,d7    * get port status
    andi.b   #1,d7          * has a byte been received
    beq      getchar        * if not check again
    move.b   data.reg2,d7   * otherwise get character
    rts                                     * go home!

```

```

*****
*
*      get_data.a68                      {subroutine}
*
*      this routine drives the analog sampling port
*      voltage samples are taken for one cycle ( 128 ) times
*      and then current samples are taken for one cycle
*      this routine doesn't alter the contents of any register
*      Call as get.data
*
*      source file : get_data.a68
*
*      revisions:
*
*      -----
*      1.00   Norm Mortensen
*
*****

```

```

*      define global labels
*      xdef get.data

*      external references
*      xref v.pntr
*      xref i.pntr

*      register assignments
*
*          d0: sample counter
*          d1: data register
*          a0: data block pointer
*

```

```

*      constant declarations

```

```

a.control.reg   equ   $01f002
a.direction.reg equ   $01f000
a.data.reg      equ   $01f000

b.control.reg   equ   $01f006
b.direction.reg equ   $01f004
b.data.reg      equ   $01f004

entry.point     equ   $000458

threshold       equ   $50
samples         equ   128

```

```

*
*----- Data Sampling Routine -----
*

get.data:      movem.l d0/d1/d2/a0,-(sp)      save entry status
               move #0,d2                clear d2

*              configure the a side of the pia
               move #$30,a.control.reg
               move #$00,a.direction.reg
               move #$34,a.control.reg

*              configure the a side of pia
               move #$00,b.control.reg
               move #$0f,b.direction.reg
               move #$04,b.control.reg
               move.b #$0f,b.data.reg

*              enable the a/d converter
               move #$07,b.data.reg      enable the a/d
               move.l #v.pntr,a0

*              Do two dummy conversions to set things up
con.1:         move.b a.control.reg+1,d1
               bpl con.1
con.2:         move.b a.control.reg+1,d1
               bpl con.2

*
*----- Zero crossing detector -----
*

chk.threshold: move.b a.control.reg+1,d1
               bpl chk.threshold        check for eoc

*              check to see if data below a certain threshold
               move.b a.data.reg+1,d1
               cmpi.b #threshold,d1
               bcc chk.threshold        if not, new sample

*              check for the - to + transition
zero.crossing: move.b a.control.reg+1,d1
               bpl zero.crossing        wait until eoc
               move.b a.data.reg+1,d1  if transition then
               bpl zero.crossing        start sampling

*              store data
               move.b d1,(a0)+         store msb

```

```

        move.b b.data.reg+1,(a0)+      store lsb
        move d2,(a0)+                  clear imag

*
*----- voltage sampling -----*
*
*           get the rest of the voltage samples
sample.v:  move #samples-2,d0                set sample cntr
           move.b a.control.reg+1,d1     check status
           bpl sample.v
           move.b a.data.reg+1,(a0)+    if eoc then grab
           move.b b.data.reg+1,(a0)+    result and store
           move d2,(a0)+                clear imag
           dbf d0,sample.v
           voltage samples done now

*
*----- current sampling -----*
*
*           set up for current sampes
           move.b #$3c,a.control.reg+1  switches mux
*                                           channel to i
*                                           pnt to start of
*                                           i block
           move.l #i.pntr,a0

*           get the current samples
sample.i:  move #samples-1,d0            init sample cntr
           move.b a.control.reg+1,d1     check status
           bpl sample.i
           move.b a.data.reg+1,(a0)+    if eoc grab 12 bit
           move.b b.data.reg+1,(a0)+    result and store
           move d2,(a0)+                clear imag
           dbf d0,sample.i
           current samples done now

           move #$0f,b.data.reg         deselect the adc

*
*----- unskew data -----*
*
*           sort out the samples and shift the data over so that
*           it's in proper format
*
           move #samples-1,d0           initialize sample counter
           move.l #v.pntr,a0            initialize the sample
*                                       pointer

```

v.sorter:	move (a0),d1	get skewed data word
	asr #4,d1	orient it properly
	andi #\$0fff,d1	mask off the unused bits
	move d1,(a0)+	put it back
	lea 2(a0),a0	skip over the imag part
	dbf d0,v.sorter	
	move #samples-1,d0	initialize sample counter
*	move.l #i.pntr,a0	initialize the sample pointer
i.sorter:	move (a0),d1	get skewed data
	asr #4,d1	orient properly
	andi #\$0fff,d1	mask off unused bits
	move d1,(a0)+	put back, next
	lea 2(a0),a0	skip over imag part
	dbf d0,i.sorter	
*	movem.l (sp)+,d0/d1/d2/a0	recover original data
	rts	
*	jmp entry.point	alter to a rts
*		when tested



```

*****
*
*      get.time                                {subroutine}
*
*      Gets the hours and minutes from the real time clock
*      and returns them in the low word of d7. The data is
*      in bcd form with tens of hours the most significant
*      nibble. No other registers are altered
*
*
*      source file : get_data.a68
*
*
*      revisions:
*      -----
*      1.00 28Apr87  Norm Mortensen
*
*****

*      global definitions
*      xdef get.time

*      external references
*      xref clock.read

*      internal definitions
*      sec.l: equ 2          clock register numbers
*      sec.l0: equ 3        i.e. 10's of seconds ...
*      min.l: equ 4
*      min.l0: equ 5
*      hrs.l: equ 6
*      hrs.l0: equ 7

get.time:

*          Save machine status
*          movem.l d0-dl,-(sp)

*          Read units of minutes
read.time  move #min.l,d7          Save the smallest unit
          bsr clock.read        of time read. If this
          move d7,d0            doesn't change other don't.

*          Clr dl for use as time storage reg
          moveq #0,dl

```

```

*           Read in tens of hours
move #hrs.10,d7
bsr clock.read
add d7,d1
asl #4,d1

*           Read in the units of hours
move #hrs.1,d7
bsr clock.read
add d7,d1
Asl #4,d1

*           Read in tens of minutes
move #min.10,d7
bsr clock.read
add d7,d1
asl #4,d1

*           Read in units of minutes
move #min.1,d7
bsr clock.read
add d7,d1

*           Make sure time hasn't changed
cmp d0,d7
bne read.time

*           If time hasn't changed then call it a clock read
move dl,d7
movem.l (sp)+,d0-d1
rts

```

```

*****
*
*          go                      {program fragment}
*
*      this routine begins program execution at a specified
*      memory location
*
*
*      source file :   go.a68
*      language:      Avocet 68000 assembler
*
*
*      revisions:
*
*      -----
*      1.00   6Mar87   Norman Mortensen
*
*****

```

```

*      global label definitions
*      xdef go

```

```

*      external labels
*      xref ascii.hex
*      xref entry.pnt
*      xref in

```

```

cr      equ      13

```

```

go:      move.l #0,d0          d0 addr pnt storage
ex.addr: bsr in

```

```

        cmpi #cr,d7
        beq ex.addr.even    stop on cr
        bsr ascii.hex
        asl.l #4,d0
        andi #$1e,ccr
        add.l d7,d0
        bra ex.addr

```

```

ex.addr.even: move d0,d7          only error protection
              asr #1,d7        against odd addresses
              bcs entry.pnt

```

```

              move.l d0,a0
              jmp (a0)

```

\*\*\*\*\*

\*  
\* hex.ascii {subroutine}  
\*

\* The hex.ascii routine converts a hex nibble in the  
\* low nibble of d7 to an ascii character code. This  
\* is returned in the low byte of d7. Since the low byte  
\* of d7 is always hex no error status is returned. The  
\* contents of no other registers are returned.  
\*

\* source file : hexascii.a68  
\*

\* Revisions:

-----  
\* 0.00 Norm Mortensen 28 Jan 87  
\*

\*\*\*\*\*

\* Entry point  
\* xdef hex.ascii \* global variable for  
\* \* the linker

hex.ascii:

movem.l a0-al,-(sp) \* save entry state of uP  
  
lea table(pc),a0 \* a0 points to the ascii table  
andi.l #\$f,d7 \* clr all but the low nibble  
move.l d7,a1 \* put hex value in a pntr reg.  
move.b (a0,a1),d7 \* pull ascii eqv. from table  
andi.l #\$ff,d7 \* clear all but the low byte  
  
movem.l (sp)+,a0-al \* return reg to entry status  
rts \* go home!

table: \* ascii code look up table  
dc.b '0123456789ABCDEF'

```

*****
*
*      hex.check                                {subroutine}
*
*      Checks to see if the low byte of d7 contains an
*      ascii code that corresponds to a hexadecimal digit.
*      If the character is valid the carry bit is cleared
*      otherwise the character is set. No registers are
*      changed.
*
*      source file : hexcheck.a68
*
*      revisions:
*      _____
*      0.00          Norm Mortensen          28 Jan 87
*
*****

```

```

*      Entry point
*
*      xdef      hex.check

```

```

hex.check:
    cmpi.b  #'0',d7      * check to see if at
    bcs     return.bad   * least $30. If not bad
    cmpi.b  #'G',d7      * check to see if less
    bcc     return.bad   * than 'G'.If >= then bad
    cmpi.b  #'A',d7      * If what's left is >= A
    bcc     return.good  * then the code is good
    cmpi.b  #'!',d7      * If what'S left is <= :
    bcs     return.good  * then it's good

return.bad  ori.b  #1,ccr  * clear the carry
            rts          * go home!

return.good andi.b  #$1e,ccr * clear the carry
            rts          * go home!

```



\*\*\*\*\*

\*

initialize (subroutine)

\*

Performs the initialization of the clock. The clock is placed in the 24 hr mode. The hours and minutes registers are then set. The time data is received via the serial communication port. No register contents are altered.

\*

\*

source file : init.a68

\*

\*

revisions:

\*

1.00 28Apr87 Norm Mortensen

\*

\*

\*\*\*\*\*

global definitions  
xdef initialize

external references  
xref ascii.hex  
xref clock.write  
xref data.pntr  
xref data.start  
xref entry.pnt  
xref in  
xref sample.cnt  
xref sys.go

internal definitions  
a.status: equ \$1f003  
a.data: equ \$1f001  
b.status: equ \$1f007  
b.data: equ \$1f005

initialize:

Save machine status and setup time  
movem.l d0-d3/d7,-(sp)

Get the time to set the clock to  
bsr in get tens hours  
move d7,d0  
bsr in get units of hours

```

move d7,d1
bsr in                               get tens of minutes
move d7,d2
bsr in                               get ubits of minutes
move d7,d3

*      Initialize the system variables
move.l #data.start,data.pntr
move #0,sample.cntr
move #0,sys.go

*      Initialize the clock
move #5,d7
bsr clock.write                      stop clock, mask int's

*      Set to 24 hr mode
move #$f1,d7
bsr clock.write

*      Put in no interrupt mode
move #7,d7
bsr clock.write
move #$f0,d7
bsr clock.write

*      Set the tens of hours nibble
move d0,d7
bsr ascii.hex
asl #4,d7
addi #7,d7
bsr clock.write

*      Set the units of hours nibble
move d1,d7
bsr ascii.hex
asl #4,d7
addi #6,d7
bsr clock.write

*      Set the tens of minutes nibble
move d2,d7
bsr ascii.hex
asl #4,d7
addi #5,d7
bsr clock.write

```



- \* Set the units of minutes nibble  
move d3,d7  
bsr ascii.hex  
asl #4,d7  
addi #4,d7  
bsr clock.write
  
- \* Start up the clock  
moveq #0,d7  
bsr clock.write
  
- \* Get back machine status and back home  
movem.l (sp)+,d0-d3/d7  
jmp entry.pnt

```

*****
*
*      memory
*
*      this routine allows the examination and alteration
*      of the harmonic analyzers memory by the North Star
*
*
*      source file :   memory.a68
*      language  :   Avocet 68000 assembler
*
*
*      revisions:
*
*      -----
*      1.00   6Mar87   Norman Mortensen
*
*
*****

```

```

*      global label definitions
*      xdef memory

```

```

*      external references
*      xref ascii.hex
*      xref disp.reg
*      xref entry.pnt
*      xref hex.ascii
*      xref in
*      xref out
*      xref putchar

```

```

*      internal label definitions
bs      equ $08      backspace
cr      equ $0d      carriage return
error   equ $00      error code
esc     equ $1b      escape
no.error equ $01      a no error code
ram.top equ $02fffe  top of the system ram
space   equ $20      space

```

```

*      Register assignments:
*      d0 : address pointer register
*      d1 : new data storage register
*      d2 : shift control register
*      d7 : i/o register

```

```

memory:          moveq #0,d0      clr the address storage reg
                 moveq #0,d1      clr temp data storage reg

addr.fetch:
    bsr in                get addr char
    cmpi.b #cr,d7        end of addr data?
    beq inrange          break out if end
    bsr ascii.hex        convert ascii to hex
    asl.l #4,d0          shift address up 1 digit
    add.l d7,d0          add in the low nibble
    bra addr.fetch       get next datum

inrange:         cmpi.l #ram.top+1,d0
                 bcs even
                 move #error,d7
                 bsr out
                 bra entry.pnt

even:           move d0,d7
                asr #1,d7
                bcc good.addr
                move #error,d7
                bsr out
                bra entry.pnt

good.addr:      move #no.error,d7
                bsr out

examine.alter:

send.addr:      move #28,d2
                move.l d0,d7
                asr.l d2,d7
                bsr hex.ascii
                bsr out
                subi #4,d2
                bcc send.addr

                move.l d0,a0
                move #12,d2
send.data:      move (a0),d7
                asr d2,d7
                bsr hex.ascii
                bsr out
                subi #4,d2

```

```

        bcc send.data

com.data:
        move #4,d3
        bsr in
        cmpi #esc,d7
        beq entry.pnt
        cmpi #space,d7
        beq increment
        cmpi #bs,d7
        beq decrement

        bsr ascii.hex
        asl #4,d1
        add d7,d1
        subi #1,d3
        bne com.data
        move d1,(a0)

*       increment the addr ptr and check to see if
*       it still pnts at system memory

increment:
        add.l #2,d0
        cmpi.l #ram.top+1,d0
        bcs examine.alter
        move.l #ram.top,d0
        bra examine.alter

*       decrement the addr ptr and check to see
*       that it still pts at system memory

decrement:
        subi #2,d0
        bcc examine.alter
        move #0,d0
        bra examine.alter

```



```

*****
*
*      putchar:                                {subroutine}
*
*      puts a one byte character out the second serial port.
*      No register contents are changed. The value to be output
*      is passed in the low byte of d7.
*
*
*      source file : putchar.a68
*
*
*      revisions:
*
*      -----
*      0.00          Norm Mortensen          20 Jan 87
*
*****

```

```

*      Entry point

```

```

      xdef    putchar

```

```

*      Internal definitions

```

```

stat.reg2    equ    $012000          these locations should
data.reg2    equ    $012002          be read as words

```

```

putchar:
      put.rdy:  move    d7,-(sp)      save byte to send
                move    stat.reg2,d7  get status
                andi    #2,d7        check to see if xmit rdy
                beq     put.rdy       if not check again
                move    (sp)+,d7     pull out data
                move    d7,data.reg2  send out data
                rts                    go home!

```

```

*****
*
*      qk_fft.a68                                {subroutine?}
*
*      Causes the harmonic analyzer to sample one cycle
*      of voltage and current data. The sampled data is then
*      processed by an FFT and the results are left in the
*      same spot as the data was. No registers are changed.
*      The call is made as qk.fft
*
*      source file : qk_fft.a68
*
*      revisions:
*
*      _____
*      1.00   Norm Mortensen
*
*****

*      global label definitions
xdef qk.fft
xdef v.pntr
xdef i.pntr

*      external references
xref entry.pnt
xref fft
xref get.data
xref out

*      internal definitions
v.pntr      equ $20800      pntr to voltage data block
i.pntr      equ $20A00      pntr to current data block
d.lgth      equ 1024       # of data bytes

*      save machine status
qk.fft:     movem.l d0/a6,-(sp)

*      get the data
           bsr get.data

*      Do an FFT on the voltage data
           move.l #v.pntr,a6
           bsr fft

*      Do an FFT on the current data
           move.l #i.pntr,a6
           bsr fft

*      send data out the serial port

```

```
                                move #d.lgth-1,d0      init cntr
                                move.l #v.pntr,a6        send off byte by byte
snd.fft.data:                   move.b (a6)+,d7
                                bsr out
                                dbf d0,snd.fft.data

*                               Go back to the monior
                                movem.l (sp)+,d0/a6
                                jmp entry.pnt
```



```

*****
*
*      sample                                {programlet}
*
*      Samples the data, does an fft, stores the time
*      and desired data if the SYS.GO flag is set. The
*      time is passed in d7. No registers are altered.
*
*      source file : sample.a68
*
*      revisions:
*      _____
*      1.00 28Apr87  Norm Mortensen
*
*****

```

```

*      global definitions
*      xdef sample
*
*      external references
*      xref data.pntr
*      xref entry.pnt
*      xref fft
*      xref get.data
*      xref hex.asci
*      xref i.pntr
*      xref sample.cntr
*      xref sys.go
*      xref v.pntr

```

sample:

```

*      Save machine status
*      movem.l d0/d1/d7/a0/a1/a6,-(sp)
*
*      See if sys.go set all samples already taken
*      move sys.go,d0          If sys.go = 0 then
*      beq done                no sampling, return.
*      cmpi #168,sample.cntr   If 168 samples already
*      beq done                then no sample taken.
*
*      Get data and perform fft
*      bsr get.data
*      move.l #v.pntr,a6
*      bsr fft                  fft on voltage samples
*      move.l #i.pntr,a6
*      bsr fft                  fft on current samples

```

*	<pre> Convert the hrs digits of the time to an integer asr #8,d7          put hours in low byte andi #\$ff,d7      and clear. move d7,d0 asr #4,d0 moveq #10,d1      tens place weighting mulu dl,d0 andi #\$f,d7      integer in d7 (word length) add d0,d7  move.l data.pntr,a0  a0 is data pointer move d7,(a0)+      store time integer </pre>	
*	<pre> Store voltage data move.l #v.pntr,al  al pntrs to data to store moveq #23,d0      0-23 harmonic to store v.store.1 move.l (al)+,(a0)+ dbf d0,v.store.1 addq.l #4,al      skip 24th harmonic moveq #19,d0      25-63 od harmonics v.store.2 move.l (al)+,(a0)+ addq.l #4,al      skip even harmonics dbf d0,v.store.2 </pre>	
*	<pre> Store the current data move.l #i.pntr,al  al pntrs to data to store moveq #23,d0 i.store.1 move.l (al)+,(a0)+ dbf d0,i.store.1 addq.l #4,al      skip 24th harmonic moveq #19,d0      25-63 odd harmonics i.store.2 move.l (al)+,(a0)+ addq.l #4,al      skip even harmonics dbf d0,i.store.2  addi #1,sample.cntr  inc sample counter move.l a0,data.pntr  save data location </pre>	
done	<pre> movem.l (sp)+,d0/d1/d7/a0/al/a6 bra entry.pnt </pre>	

```

*****
*
*      send.data                                {programlet}
*
*      Sends data in one hour chunks to the Z-171. The
*      sample groups taken are sent one group at a time. The
*      transmission format is
*
*      - send number of hours worth of data as an integer
*        in Intel order
*      - Send an integer that tells what hour the sample
*        was taken
*      - send 44 harmonics voltage (2 integers/harmonic)
*      - send 44 harmonics current (2 integers/harmonic)
*      - send checksum as integer in Intel order
*      - wait for reception of a character '*'
*      - start over with time integer
*
*      All integers are two bytes in length. No registers are
*      altered by this machine.
*
*      source file:    snd_data.a68
*
*      author:         Norm Mortensen
*
*      date:           28Apr87
*
*****

*      global definitions
*      xdef send.data

*      external references
*      xref data.start
*      xref entry.pnt
*      xref sample.cntr

*      internal references
*      sreg1: equ $011001
*      dreg1: equ $011003

send.data:
    movem.l d0-d2/d7/a0,-(sp)        save status

*      Send number of samples in Intel order
    move sample.cntr,d7
    bsr snd
    asr #8,d7
    bsr snd

```

```

*           Send the data packages if any
           moveq #0,d0           clear sample counter
chk.for.end: move.l #data.start,a0   a0 data ptr
           cmp sample.cntr,d0     check for end data
           beq done

*           Send out time and data in Intel order
           moveq #0,d1           clr the checksum reg
data.out:  move #176,d2           Send 177 words of data
           move (a0)+,d7
           add d7,d1             accumulate checksum
           bsr snd
           asr #8,d7
           bsr snd
           dbf d2,data.out

*           Send the checksum in Intel order
           move d1,d7
           bsr snd             lsb of checksum
           asr #8,d7
           bsr snd             msb of checksum

*           Increment counter of samples sent and go again
           addq #1,d0
           bsr rcv             Wait for okay for next
bra chk.for.end           data package.

*           Wrap up if done
done:      movem.l (sp)+,d0-d2/d7/a0
           bra entry.pnt

```

```

*=====
*
*           snd                 {local routine}
*
*           Outputs low byte of d7 out serial port 1. It does NOT
*           expect an echo. No registers are altered.
*
*=====

```

```

snd:      move d7,-(sp)           Save data on stack.
tx.rdy:  move.b sreg1,d7         Check to see if the
           andi #2,d7           transmit register is
           beq tx.rdy           empty.
           move (sp)+,d7        get data
           move.b d7,dreg1      send data
           rts

```

```

*=====
*
*          rcv                      {local routine}
*
*      Gets an input character from serial port one and
*      returns it in the low byte of d7. No other registers
*      are altered. It does NOT provide an echo.
*
*=====

```

```

rcv:          move.b sreg1,d7          Check for reception.
              andi #1,d7              If no character continue
              beq rcv                  checking.
              move.b dreg1,d7         get data
              rts

```

\*\*\*\*\*

\*  
\* send.time {programlet}  
\*

\* Sends the time of the harmonic analyzers real time  
\* clock out the serial port in ascii coded digits.  
\* The hours and minutes are sent (4 characters).  
\* No registers are altered by this routine  
\*

\* source file: snd\_time.a68  
\*

\* author: Norm Mortensen  
\*

\* date: 30Apr87  
\*

\*\*\*\*\*

\* global definitions  
\* xdef send.time

\* external references  
\* xref clock.read  
\* xref entry.pnt  
\* xref get.time  
\* xref hex.ascii  
\* xref out

send.time:

\* Save machine status  
\* movem.l d0-d1/d7,-(sp)

\* Get time and send out  
\* bsr get.time  
\* move d7,d0  
\* move #12,d1 initialize shift reg  
send.out \* move d0,d7 get raw data  
\* asr d1,d7 shift to appropriate byte  
\* bsr hex.ascii convert to ascii  
\* bsr out send  
\* subi #4,d1 adjust shift reg  
\* bcc send.out

\* Recover machine status and return  
\* movem.l (sp)+,d0-d1/d7  
\* bra entry.pnt

```

*****
*
*      trace.a68                                {exception}
*
*      The following code is executed when a trace exception
*      occurs on the harmonic analyzer. The contents of all the
*      registers are displayed as well as the contents of the
*      pc and the ccr.
*
*
*      source file : trace.a68
*
*
*      Revisions:
*
*      _____
*      0.00          Norm Mortensen          20 Jan 87
*
*
*****

```

```

*      global label definition
*      xdef trace

```

```

*      external references
*      xref  ascii.hex
*      xref  getchar
*      xref  hex.check
*      xref  putchar
*      xref  disp.reg
*      xref  disp.string

```

```

*      Internal definitions

```

```

rmaskl      reg    d0-d7/a0-a6
CR          equ    $0d
LF          equ    $0a
trace       equ    $1000

```

```

org trace
movem.l rmaskl,-(sp)  * all register contents
                     * saved on the stack
move.l sp,a0          * copy the stack pointer

```

```

lea        newline(pc),al

```

```

        move.l  al,d7
        bsr    disp.string
        bsr    disp.string      * clear two blank lines

disp.d0:    lea.l  d0.str(pc),al
            move.l al,d7        * point to the d0 header
            bsr    disp.string  * display the header
            move.l (a0)+,d7     * copy d0 from the stack
            bsr    disp.reg

disp.d1:    lea.l  d1.str(pc),al
            move.l al,d7        * display d1
            bsr    disp.string
            move.l (a0)+,d7
            bsr    disp.reg

disp.d2:    lea.l  d2.str(pc),al
            move.l al,d7        * display d2
            bsr    disp.string
            move.l (a0)+,d7
            bsr    disp.reg

disp.d3:    lea.l  d3.str(pc),al
            move.l al,d7        * display d3
            bsr    disp.string
            move.l (a0)+,d7
            bsr    disp.reg

disp.d4:    lea.l  d4.str(pc),al
            move.l al,d7        * display d4
            bsr    disp.string
            move.l (a0)+,d7
            bsr    disp.reg

disp.d5:    lea.l  d5.str(pc),al
            move.l al,d7        * display d5
            bsr    disp.string
            move.l (a0)+,d7
            bsr    disp.reg

disp.d6:    lea.l  d6.str(pc),al

```



```

        move.l  al,d7          * display d6
        bsr    disp.string
        move.l  (a0)+,d7
        bsr    disp.reg

disp.d7:  lea.l  d7.str(pc),al
        move.l  al,d7          * display d7
        bsr    disp.string
        move.l  (a0)+,d7
        bsr    disp.reg

disp.a0:  lea.l  a0.str(pc),al
        move.l  al,d7          * display a0
        bsr    disp.string
        move.l  (a0)+,d7
        bsr    disp.reg

disp.a1:  lea.l  a1.str(pc),al
        move.l  al,d7          * display a1
        bsr    disp.string
        move.l  (a0)+,d7
        bsr    disp.reg

disp.a2:  lea.l  a2.str(pc),al
        move.l  al,d7          * display a2
        bsr    disp.string
        move.l  (a0)+,d7
        bsr    disp.reg

disp.a3:  lea.l  a3.str(pc),al
        move.l  al,d7          * display a3
        bsr    disp.string
        move.l  (a0)+,d7
        bsr    disp.reg

disp.a4:  lea.l  a4.str(pc),al
        move.l  al,d7          * display a4
        bsr    disp.string
        move.l  (a0)+,d7
        bsr    disp.reg

disp.a5:  lea.l  a5.str(pc),al
        move.l  al,d7          * display a5

```

```

        bsr      disp.string
        move.l   (a0)+,d7
        bsr      disp.reg

disp.a6:    lea.l   a6.str(pc),al
            move.l  al,d7          display a6
            bsr      disp.string
            move.l  (a0)+,d7
            bsr      disp.reg

disp.a7:    lea.l   a7.str(pc),al
            move.l  al,d7          a7 is the stack pointer
            bsr      disp.string  In order to compensate for
            move.l  a0,d7          the use of the SP
            andi.b  #$1e,ccr      in processing the trace
            addi.l  #6,d7         exception an offset
            bsr      disp.reg      is added

disp.sr:    lea.l   sr.str(pc),al
            move.l  al,d7
            bsr      disp.string
            move.l  #0,d7         clears the reg since the
            move    (a0)+,d7     SR is only a word wide
            bsr      disp.reg

disp.pc:    lea.l   pc.str(pc),al
            move.l  al,d7         program counter the next
            bsr      disp.string  long word on the stack
            move.l  (a0),d7      don't increment a0 so the
            bsr      disp.reg     sr can be reached with a
                                   predecremented read

```

\* Toggle the t bit of the status register

```

        move     -(a0),d7        get the sr
        bmi     trace.on
trace.off:  lea.l   t0.str(pc),al
            move.l  al,d7         display trace off message
            bsr      disp.string
            bra     cont.l
trace.on:   lea.l   t1.str(pc),al

```

```

                                move.l  al,d7          display trace on message
                                bsr      disp.string

cont.l:                          lea.l   t.str(pc),al
                                move.l  al,d7          display direction for
                                bsr      disp.string    trace

                                bsr      getchar       get answer
                                cmpi.b  #CR,d7        if CR then don't change
                                beq      no.toggle.t    the status of t

toggle.t:                         eori    #$8000,(a0)   toggle the t bit
no.toggle.t:                      move    (a0)+,d7     point back at the pc on
                                the stack

                                lea     newline(pc),al
                                move.l  al,d7
                                bsr     disp.string    go to a new line

```

```

*      Display the header for program counter alterations and
*      see if that is to be changed.

```

```

                                lea.l  pc.str(pc),al
                                move.l  al,d7
                                bsr     disp.string

                                move.l  #0,d0          clear pc storage reg

                                bsr     getchar
                                cmpi.b  #CR,d7
                                beq     return.trace   if CR then exit trace

                                bsr     hex.check      C=0 if the byte in d7 is
                                bcs     next.char      hex digit in ascii. If bad
                                                        wait for a new input
                                bsr     putchar       echo to the screen
                                bsr     ascii.hex     convert to hex
                                andi    #$1e,ccr     clear the carry
                                add.l   d7,d0        put in low nibble of
*
                                                        pc storage register

```

next.char:	<pre> bsr    getchar cmpi.b #CR,d7 bne    cont.2 move.l d0,(a0) bra    return.trace </pre>	<pre> If a CR put d0 into the pc's location on the stack and return from the trace </pre>
cont.2:	<pre> bsr    hex.check bcs    next.char bsr    putchar bsr    ascii.hex asl.l  #4,d0 andi   #\$1e,CCR add.l  d7,d0 bra    next.char </pre>	<pre> C=0 means good character no echo if bad char echo convert to hex shift by one hex digit clear the carry </pre>
return.trace:	<pre> lea    newline(pc),al move.l al,d7 bsr    disp.string bsr    disp.string  movem.l (sp)+,rmask1 rte </pre>	<pre> clear two blank lines  get back entry state go back </pre>

```

*****
*
*      messages                      {constant definitions}
*
*      this file countains the messages used by the trace
*      exception. It is included as part of the trace file
*
*
*      revisions:
*      -----
*      0.00          Norm Mortensen          3 Feb 87
*
*****

```

```

a0.str      db 'a0: ',0
a1.str      db 'a1: ',0
a2.str      db 'a2: ',0
a3.str      db 'a3: ',0
a4.str      db 'a4: ',0
a5.str      db 'a5: ',0
a6.str      db 'a6: ',0
a7.str      db 'a7: ',0

d0.str      db 'd0: ',0
d1.str      db 'd1: ',0
d2.str      db 'd2: ',0
d3.str      db 'd3: ',0
d4.str      db 'd4: ',0
d5.str      db 'd5: ',0
d6.str      db 'd6: ',0
d7.str      db 'd7: ',0

pc.str      db 'pc: ',0
sr.str      db 'sr: ',0

t0.str      db 'trace is disabled',CR,LF,0
t1.str      db 'trace is enabled',CR,LF,0

t.str       db 'trace: ',0

alter.pc    db 'to keep the pc the same '
            db 'type <return>',CR,LF,0
            db 'to alter type in the new '
            db 'address',CR,LF,0

newline     db CR,LF,0

```

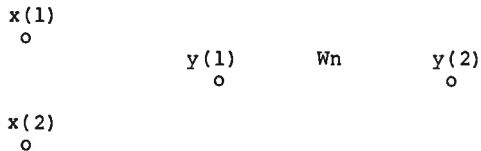
## APPENDIX C

Source code listings of `SETUP`, the Zenith Z-171 interface program. For instructions on the operation of `SETUP` see [7].

## APPENDIX D : SIGNAL FLOW GRAPHS

Signal flow graphs are a graphical way of representing the linear relationships between a set of input variables and a set of output variables. Signal flow graphs are useful in a number of areas. These include graphic representation of signal processing algorithms, and linear systems. This appendix describes how a signal flow graph is read.

In order to explain how to read a signal flow graph the simple figure below will be used.



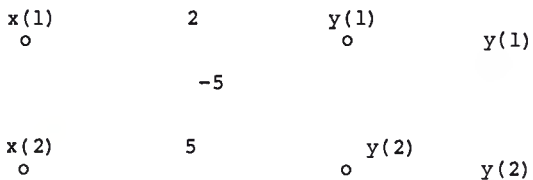
The variables  $x(1)$ ,  $x(2)$ ,  $y(1)$  and  $y(2)$  are represented by the nodes. Whenever two signals (lines) meet at a node their values are added to give the value of the node (variable). In the figure above  $y(1) = x(1) + x(2)$ . The value of a signal (line) is the value of the node it originated from. This is then multiplied by any scaling factor indicated beside the line. If no scaling factor is indicated then it is assumed to be 1. In the example above

the signal flowing from  $x(1)$  to  $y(1)$  has the value  $x(1)$ .  
 The signal flowing from  $y(1)$  to  $y(2)$  has the value  $y(1)W_n$ .  
 Since this is the only signal flowing into the  $y(2)$  node  
 $y(2) = y(1)W_n$ .

As another example the signal flow graph for the  
 following equations is shown below.

$$y(1) = 2x(1) - 5x(2)$$

$$y(2) = 5x(2) + y(1)$$





## REFERENCES

1. Motorola, 16-Bit Microprocessor User's Manual 3ed., Prentice-Hall, Englewood Cliffs, N.J., 1982.
2. J.D. Greenfield and W.C. Wray, Using Microprocessors and Microcomputers the 6800 Family, Wiley and Sons, New York, N.Y., 1981.
3. N. Ahmed and T. Natarajan, Discrete Time Signals and Systems, Reston, Reston, V.A., 1983.
4. W.D. Stanley, Digital Signal Processing, Reston, Reston, V.A., 1975.
5. N.K. Bose, Digital Filters : Theory and Applications, North-Holland, New York, N.Y., 1985.
6. P.D. Welch, "The use of Fast Fourier Transform for the Estimation of Power Spectra : A Method Based on Time Averaging Over Short Modified Periodograms," IEEE Trans. Audio and Electroacoust., Vol. AU-15, pp. 70-73, June 1967.
7. G.L. Johnson, A.N. Mortensen, A.J. Heber, Measurement of Harmonics on Kansas Utilities, Report to Kansas Electric Utilities Research Program (KEURP), September, 1987.
8. Analog Devices, Data Acquisition Handbook, Analog Devices, 1985.

AN AUTOMATED HARMONIC ANALYZER FOR USE IN STUDYING THE  
COMMERCIAL POWER SYSTEM

by

NORMAN MORTENSEN

BSEE, Kansas State University, 1986

---

AN ABSTRACT OF A MASTER'S THESIS

submitted in partial fulfillment  
of the requirements for the degree

MASTER OF SCIENCE

Department of Electrical and Computer Engineering

KANSAS STATE UNIVERSITY  
Manhattan, Kansas

1988

## Abstract

This document discusses the design of an automated harmonic analyzer for use in studying the commercial power system. This unit was developed in order to complete a study of harmonics on Kansas utilities sponsored by the Kansas Electric Utilities Research Program (KEURP). A brief introduction of the causes and possible effects of power system harmonics is presented. The physical design and operation of the harmonic analyzer are examined. The software for the harmonic analyzer's control program, ARDVARK, is explained on the flow chart level. The testing procedures used to evaluate the harmonic analyzer are explained and the results of these tests are presented.

The harmonic analyzer presented in this document has the following capabilities and features. It is a portable unit capable of independent operation for one week when properly initialized. It measures the first sixty four harmonics of 60 Hz. The analyzer's measurement error was found to be 0.1% of the magnitude of the 60 Hz fundamental waveform based on the evaluation tests described in this document.