

A STUDY OF THE HILL-FUNCTION SOLUTION
TO PROBLEMS OF PROPAGATION IN STRATIFIED MEDIA

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by

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I. INTRODUCTION

The problem of finding the electromagnetic fields due to a plane wave incident upon a plane-stratified medium reduces to the problem of solving a second-order linear differential equation with variable coefficients. There exists a limited number of specific layer variations which yield a differential equation solvable in terms of known functions [1]-[9]. If the dielectric profile is not one of these, then the differential equation may be solved by various approximate methods [10]. At high frequencies, that is, when the dielectric variation over distances of the order of a wavelength is small, the WKB (Wentzel-Kramers-Brillouin) or phase-integral solution may be used [1, Ch. 4], [2]. One may also construct an approximate differential equation to obtain a solution good at either high or low frequencies. Another technique is to approximate the dielectric profile by one or more profiles for which an exact solution is known. For example, probably the most obvious approximation is a "stepped" profile consisting of homogeneous layers of arbitrary number and size.

In a recent paper by Casey [11], it was shown that the fields in a plane-stratified layer may be expressed in terms of Hill functions. This method gives formally exact solutions for a nearly arbitrary dielectric profile in the layer. We shall in this paper be concerned with finding the fields outside the inhomogeneous region using the Hill function approach. More precisely, we are interested in the rate of convergence of certain infinite determinants associated with the Hill function method. The rate of convergence will be seen to depend upon frequency and other parameters of the problem. Additionally, we shall be interested in the effect of certain dielectric profile approximations which may conveniently be made using the

Hill-function method of solution. For this study, we shall consider a dielectric profile for which a known-function solution may be obtained and compare numerically the known-function solution and the Hill-function solution.

II. FORMULATION OF THE PROBLEM

The geometry of the problem is shown in Figure 1. A plane wave is incident upon a stratified dielectric layer of thickness d and permittivity $\epsilon(z)$. On either side of the transition layer are homogeneous regions of permittivity ϵ_1 and ϵ_2 . The permeability in all regions is μ_0 . We shall consider only the case of a plane wave polarized perpendicular to the plane of incidence, as shown in Fig. 1. Also, $e^{-i\omega t}$ time dependence will be assumed for all field quantities and suppressed throughout.

In region (1), $z < 0$, the total electric field is the sum of the incident and reflected fields,

$$E_y^{(1)} = \psi_1 e^{ik_1 \sin \theta_i x} \quad (1)$$

with

$$\psi_1 = e^{ik_1 \cos \theta_i z} + R e^{-ik_1 \cos \theta_i z}. \quad (2)$$

In region (2), $z > d$, the transmitted field is

$$E_y^{(2)} = \psi_2 e^{ik_2 \sin \theta_t (z-d)} \quad (3)$$

with

$$\psi_2 = T e^{ik_2 \cos \theta_t (z-d)}. \quad (4)$$

R and T are respectively the reflection and transmission coefficients to be determined, and $k_{1,2} = \omega \sqrt{\mu_0 \epsilon_{1,2}}$

Now, the following conditions describe the transition region;

- 1) linear, 2) isotropic, 3) time-invariant, 4) inhomogeneous, 5) source-free. For this case the wave equation for the electric field \bar{E} is

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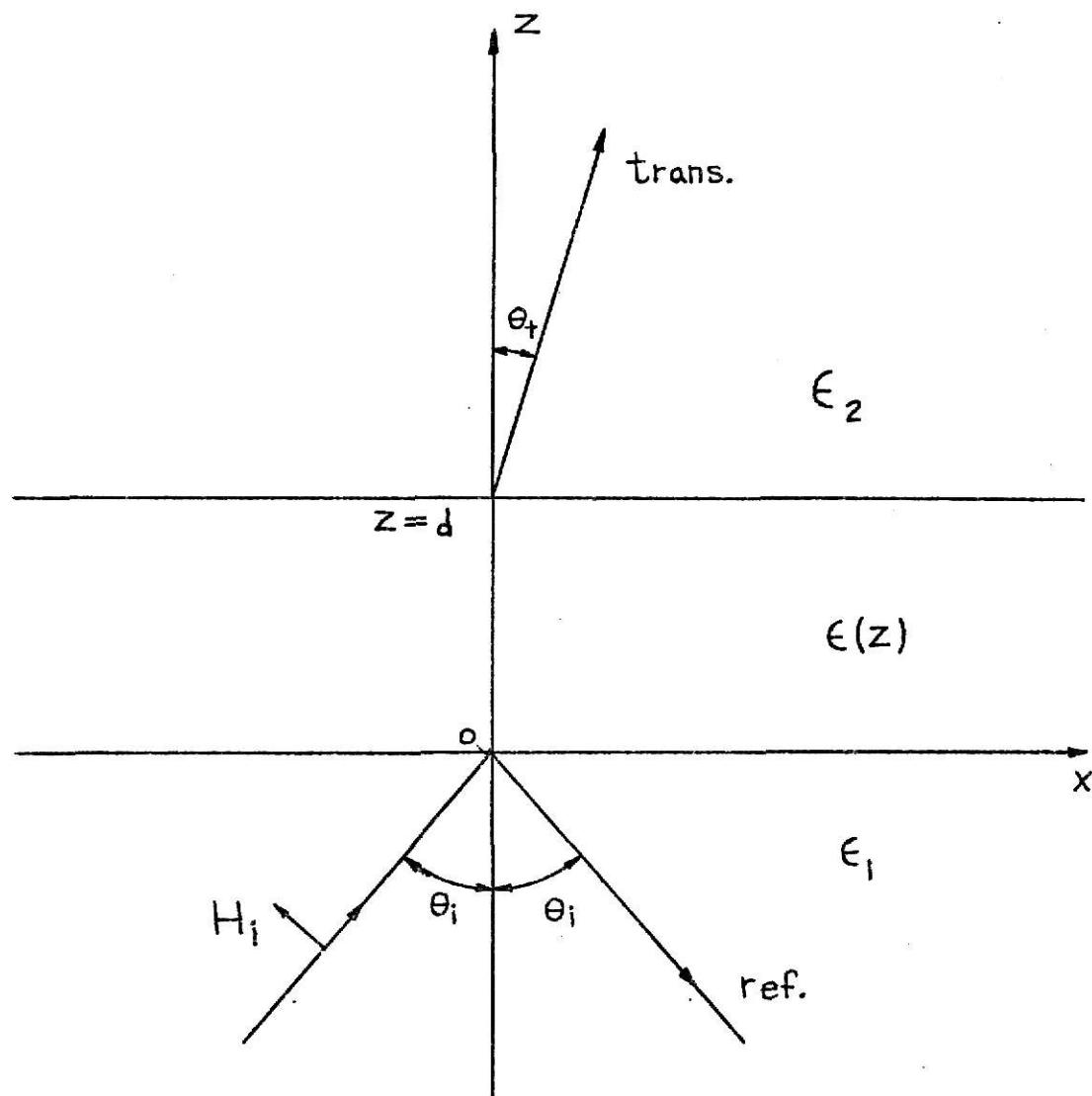


Fig. 1. Geometry of the problem of a plane wave incident upon an inhomogeneous dielectric layer.

$$\nabla^2 \bar{E} + \nabla \left(\bar{E} \cdot \frac{\nabla \epsilon}{\epsilon} \right) + k^2 \bar{E} = 0 . \quad (5)$$

We may specialize this further by noting that for our case the field will have only a y component, with no variation in y, and that the permittivity is a function of z only; or $E_x = E_z = 0$, $\frac{\partial E}{\partial y} = 0$, and $\epsilon = \epsilon(z)$. The wave equation thus becomes

$$[\nabla_{xz}^2 + k^2(z)]E_y = 0 . \quad (6)$$

Using the method of separation of variables, we seek solutions of the form

$$E_y = \psi(z) X(x) . \quad (7)$$

Substituting (7) into (6) and dividing by ψX gives

$$\frac{\psi''}{\psi} + k^2 = - \frac{X''}{X} . \quad (8)$$

Since x and z are independent variables, each must be constant. If this constant value is γ^2 , then we obtain the two equations

$$X'' + \gamma^2 X = 0 \quad (9)$$

$$\psi'' + (k^2 - \gamma^2)\psi = 0 . \quad (10)$$

Solutions to (9) are

$$X = e^{\pm i \gamma x} . \quad (11)$$

A boundary condition at $z = 0$ is the continuity of E_y so the x variation of the fields in the layer must be that of the fields in region (1). This requires that

$$\gamma = k_1 \sin \theta_i \quad (12)$$

and

$$X = e^{ik_1 \sin \theta_i x} . \quad (13)$$

Note that the variation in x is not a function of k - since of course k is a function of z - and so does not depend upon the dielectric profile.

We have now reduced the problem to that of solving (10), a second-order, linear differential equation with a non-constant coefficient. In order to numerically investigate the resulting fields, we choose a linear dielectric profile as shown in Fig. 2. We have in the layer

$$\epsilon(z) = \epsilon_1(1 + az) \quad (14)$$

where

$$a = (\epsilon_r - 1)/d \quad (15)$$

and

$$\epsilon_r = \epsilon_2/\epsilon_1 = k_2^2/k_1^2 . \quad (16)$$

The equation to be solved may now be written as

$$\psi'' + k_1^2(\cos^2 \theta_i + az)\psi = 0 . \quad (17)$$

Subsequent application of boundary conditions at $z = 0$ and $z = d$ will complete the solution.

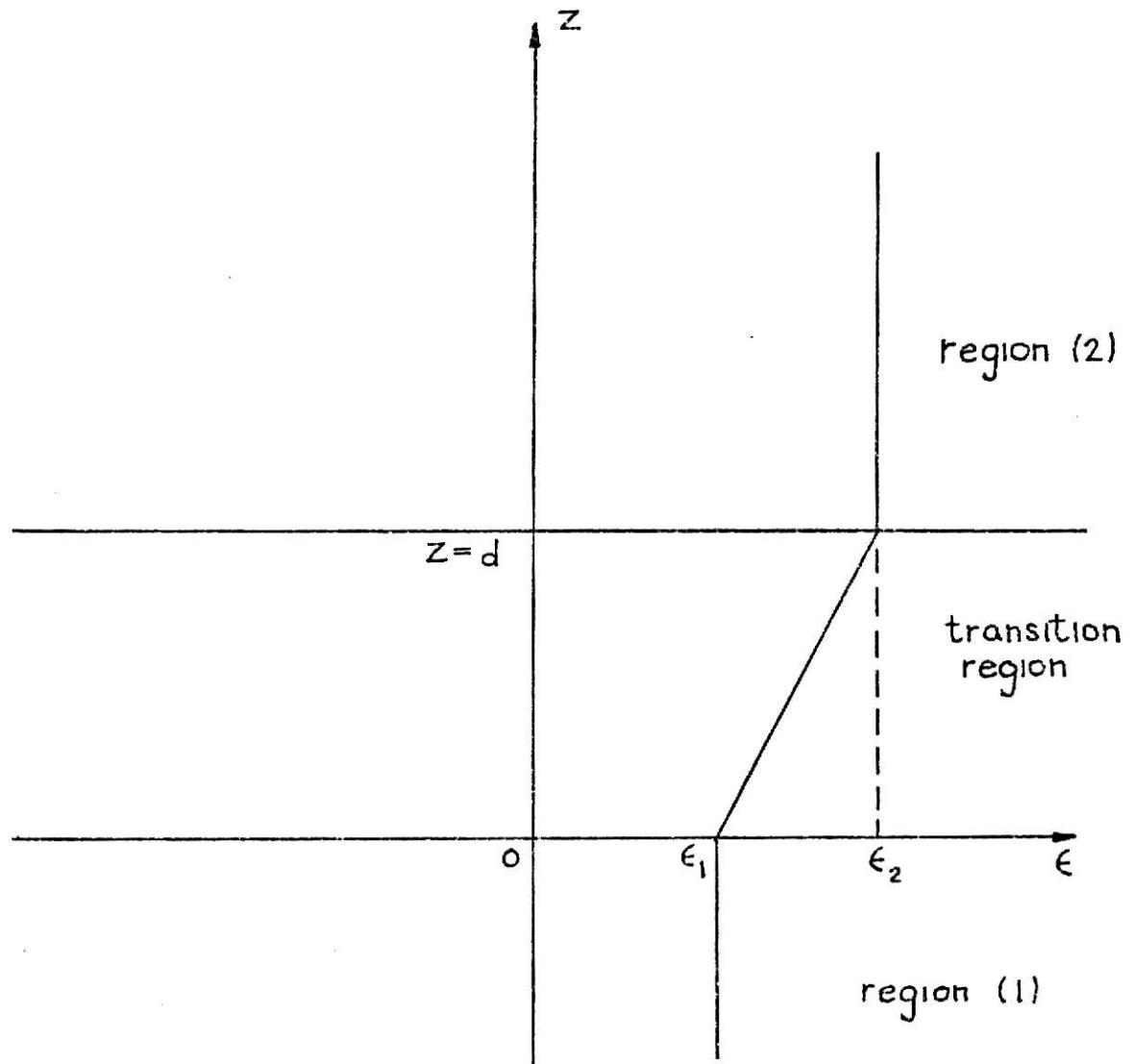


Fig. 2. Transition region with linear dielectric profile.

III. KNOWN-FUNCTION SOLUTION

To put (17) into a familiar form, let

$$w = \frac{2k_1}{3a} (az + \cos^2 \theta_i)^{3/2} \quad (18)$$

$$\psi = w^{1/3} v(w) \quad (19)$$

to obtain

$$w' = k_1 (az + \cos^2 \theta_i)^{1/2} \quad (20)$$

$$w'' = \frac{ak_1}{2} (az + \cos^2 \theta_i)^{-1/2} \quad (21)$$

$$\begin{aligned} \psi''' &= w''' \left(\frac{w^{-2/3} v}{3} + w^{1/3} v' \right) \\ &+ w'^2 [v''' w^{1/3} + \frac{2}{3} w^{-2/3} v' - \frac{2}{9} w^{-5/3} v] . \end{aligned} \quad (22)$$

By substituting these expressions into (17), a change of variables is accomplished and (17) becomes

$$w''' \left(\frac{w^{-2/3} v}{3} + w^{1/3} v' \right) + w'^2 [w^{1/3} v''' + \frac{2}{3} w^{-2/3} v' + (w^{1/3} - \frac{2}{9} w^{-5/3}) v] = 0 . \quad (23)$$

Noting that $\frac{w'^2}{w'''} = 3w$, we divide (23) by $w''' w^{4/3}$ and obtain Bessel's equation of order $1/3$,

$$v''' + \frac{v'}{w} + (1 - \frac{1}{9w^2}) v = 0 . \quad (24)$$

The solution to (24) may be written as

$$v = A J_{1/3}(w) + B J_{-1/3}(w) \quad (25)$$

where $J_{\pm 1/3}$ denotes the Bessel function of the first kind of order $\pm 1/3$ and

A and B are constants to be evaluated. The fields in the layer are now

$$E_y = \psi e^{ik_1 \sin \theta_i x} \quad (26)$$

where

$$\psi = w^{1/3} [A J_{1/3}(w) + B J_{-1/3}(w)] . \quad (27)$$

Now we require that the tangential components of the electric and magnetic fields be continuous across the boundaries, that is at $z = 0$

$$E_y^{(1)} = E_y \quad (28)$$

$$H_x^{(1)} = H_x \quad (29)$$

and at $z = d$

$$E_y^{(2)} = E_y \quad (30)$$

$$H_x^{(2)} = H_x . \quad (31)$$

H_x in all regions is given by the field equation

$$H_x = \frac{i}{\omega \mu_0} \frac{\partial E_y}{\partial z} . \quad (32)$$

Substituting the expressions for the fields, given by (1), (3), (26), and (32), into (28)-(31) yields four simultaneous linear equations in A , B , R , T :

$$\begin{bmatrix} (-i w_o^{1/3} J_{-2/3}^0) & (i w_o^{1/3} J_{2/3}^0) & 1 & 0 \\ (w_o^{1/3} J_{1/3}^0) & (w_o^{1/3} J_{-1/3}^0) & -1 & 0 \\ (-i w_d^{1/3} J_{-2/3}^d) & (i w_d^{1/3} J_{2/3}^d) & 0 & -1 \\ (w_d^{1/3} J_{1/3}^d) & (w_d^{1/3} J_{-1/3}^d) & 0 & -1 \end{bmatrix} \begin{bmatrix} A \\ B \\ R \\ T \end{bmatrix} = \begin{bmatrix} 1 \\ 1 \\ 0 \\ 0 \end{bmatrix} . \quad (33)$$

Here we have introduced the notation of subscript o and d indicating the argument w evaluated at z = 0 and z = d and superscript o and d meaning the function J evaluated at z = 0 and z = d. From these equations it is found that

$$A = \frac{2w_o^{-1/3} (i J_{2/3}^d - J_{-1/3}^d)}{D} \quad (34)$$

$$B = \frac{2w_o^{-1/3} (J_{1/3}^d + i J_{-2/3}^d)}{D} \quad (35)$$

$$R = \frac{i J_{2/3}^d J_{1/3}^o - J_{-1/3}^d J_{1/3}^o + J_{1/3}^d J_{-1/3}^o + i J_{-2/3}^d J_{-1/3}^o}{- J_{2/3}^d J_{-2/3}^o - i J_{-1/3}^d J_{-2/3}^o - i J_{1/3}^d J_{2/3}^o + J_{-2/3}^d J_{2/3}^o} \quad (36)$$

$$T = \frac{i 2\sqrt{3} w_o^{-1/3} w_d^{-2/3}}{\pi D} , \quad (37)$$

where

$$D = (i J_{2/3}^d - J_{-1/3}^d) J_{1/3}^o + (J_{1/3}^d + i J_{-2/3}^d) J_{-1/3}^o + (J_{2/3}^d + i J_{-1/3}^d) J_{-2/3}^o + (i J_{1/3}^d - J_{-2/3}^d) J_{2/3}^o . \quad (38)$$

This completes the solution for the fields in all regions.

As stated earlier, we are interested in the field outside the transition region which requires that we know R and T. It will be seen later that the Hill-function approach requires the evaluation of the same infinite determinants to find either R or T. Therefore it will be necessary to consider only one of these quantities. We will use R.

Now, for computational purposes the reflection coefficient R will be expressed in terms of the Airy functions, $Ai(\cdot)$ and $Bi(\cdot)$. By substituting into (36) the relations [12]

$$J_{\pm 1/3}(w) = \frac{1}{2} \sqrt{3/u} [\sqrt{3} Ai(-u) \mp Bi(-u)] \quad (39)$$

$$J_{\pm 2/3}(w) = \sqrt{3}/2u [\pm \sqrt{3} Ai'(-u) + Bi'(-u)] , \quad (40)$$

where

$$u = (3/2 w)^{2/3} , \quad (41)$$

we obtain

$$R = \frac{-(Ai^d, Bi^o, - Bi^d, Ai^o) + \sqrt{u_d u_o} (Ai^d Bi^o - Bi^d Ai^o)}{(Ai^d, Bi^o, - Bi^d, Ai^o) + \sqrt{u_d u_o} (Ai^d Bi^o - Bi^d Ai^o)} + \frac{i[-\sqrt{u_d} (Ai^d Bi^o, - Bi^d Ai^o) + \sqrt{u_o} (-Ai^d, Bi^o + Bi^d, Ai^o)]}{(Ai^d, Bi^o, - Bi^d, Ai^o) + \sqrt{u_d u_o} (Ai^d Bi^o - Bi^d Ai^o)} + \frac{i[\sqrt{u_o} (-Ai^d, Bi^o + Bi^d, Ai^o) + \sqrt{u_d} (Ai^d Bi^o, - Bi^d Ai^o)]}{(Ai^d, Bi^o, - Bi^d, Ai^o) + \sqrt{u_d u_o} (Ai^d Bi^o - Bi^d Ai^o)} \quad (42)$$

It should be noted that the arguments of the Airy functions in this expression are $-u_o$ and $-u_d$. If we let

$$v_o = e^{-i2\pi/3} u_o \quad (43)$$

$$v_d = e^{-i2\pi/3} u_d , \quad (44)$$

then it can be shown that

$$R = \frac{-(Ai^d, Bi^o, - Bi^d, Ai^o) + \sqrt{v_o v_d} (Ai^d Bi^o - Bi^d Ai^o)}{(Ai^d, Bi^o, - Bi^d, Ai^o) + \sqrt{v_o v_d} (Ai^d Bi^o - Bi^d Ai^o)} + \frac{i[\sqrt{v_d} (Ai^d Bi^o, - Bi^d Ai^o) - \sqrt{v_o} (-Ai^d, Bi^o + Bi^d, Ai^o)]}{(Ai^d, Bi^o, - Bi^d, Ai^o) + \sqrt{v_o v_d} (Ai^d Bi^o - Bi^d Ai^o)} + \frac{i[-\sqrt{v_o} (Bi^d, Ai^o - Ai^d, Bi^o) - \sqrt{v_d} (Ai^d Bi^o, - Bi^d Ai^o)]}{(Ai^d, Bi^o, - Bi^d, Ai^o) + \sqrt{v_o v_d} (Ai^d Bi^o - Bi^d Ai^o)} \quad (45)$$

where the arguments are $-v_o$ and $-v_d$. In order to have the arguments of the Airy functions real, (42) is used when $\epsilon_r > 1$ and (45) when $\epsilon_r < 1$. This is done, again, for computational purposes.

IV. HILL-FUNCTION SOLUTION

Let the following change of variables be introduced into (10):

$$\zeta = \pi z / 2d \quad (46)$$

$$u(\zeta) = \psi(z) . \quad (47)$$

This gives

$$u'' + \left(\frac{2d}{\pi}\right)^2 (k^2 - \gamma^2)u = 0 . \quad (48)$$

Now if the coefficient $g(z) = \left(\frac{2d}{\pi}\right)^2 (k^2 - \gamma^2)$ is even and periodic in π , as a function of ζ , then it may be expressed as a Fourier cosine series and (48) becomes

$$u'' + (\lambda + 2 \sum_{n=1}^{\infty} g_n \cos 2n\zeta)u = 0 . \quad (49)$$

This is Hill's equation. If $\sum_{n=1}^{\infty} |g_n|$ converges, then we may proceed and write the solution in terms of Hill functions [13]-[15],

$$u = Au_1 + Bu_2 , \quad (50)$$

where A and B are constants.

As will be seen, it is not necessary to specify the functions u_1 and u_2 ; that is we need not actually solve the differential equation for the fields in the transition layer if only R or T is desired.

In the layer, then,

$$E_y = Au_1 + Bu_2 \quad (51)$$

$$H_x = \frac{i\pi}{2d\omega\mu_0} [Au'_1 + Bu'_2] \quad (52)$$

where the x dependence has been suppressed. Application of boundary conditions at $z = 0$ and $z = d$, as given in (28)-(31), gives

$$\begin{bmatrix} \frac{-i\pi}{2dk_1 \cos \theta_i} u_1'(0) & \frac{-i\pi}{2dk_1 \cos \theta_i} u_2'(0) \\ u_1(0) & u_2(0) \\ \frac{-i\pi}{2dk_2 \cos \theta_t} u_1'(\pi/2) & \frac{-i\pi}{2dk_2 \cos \theta_t} u_2'(\pi/2) \\ u_1(\pi/2) & u_2(\pi/2) \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ -1 & 0 \\ 0 & -1 \\ 0 & -1 \end{bmatrix} \begin{bmatrix} A \\ B \\ R \\ T \end{bmatrix} = \begin{bmatrix} 1 \\ 1 \\ 0 \\ 0 \end{bmatrix}. \quad (53)$$

From (53), the reflection coefficient is found to be

$$R = \frac{-(u_2^d, u_1^o, -u_1^d, u_2^o) + a_o a_d (u_1^o u_2^d - u_2^o u_1^d) + i[-a_d (u_1^d u_2^o, -u_2^d u_1^o) + a_o (u_1^o u_2^d, -u_2^o u_1^d)]}{(u_2^d, u_1^o, -u_1^d, u_2^o) + a_o a_d (u_1^o u_2^d - u_2^o u_1^d) + i[a_o (u_2^d, u_1^o - u_1^d, u_2^o) + a_d (u_1^d u_2^o, -u_2^d u_1^o)]} \quad (54)$$

where

$$a_o = L_1 \cos \theta_i \quad (55)$$

$$a_d = L_1 (\epsilon_r - \sin^2 \theta_i)^{1/2} \quad (56)$$

$$L_1 = \frac{2k_1 d}{\pi}. \quad (57)$$

Now the Hill functions may be normalized by setting

$$u_1(0) = u_2(0) = 1 \quad (58)$$

$$u_1'(0) = u_2'(0) = 0 \quad (59)$$

and we have

$$R = \frac{u_1^d + a_o a_d u_2^d - i a_d u_1^d + i a_o u_2^d}{-u_1^d + a_o a_d u_2^d + i a_d u_1^d + i a_o u_2^d}. \quad (60)$$

So to find the reflection coefficient we must find $u_1(\pi/2)$, $u_2(\pi/2)$, $u_1'(\pi/2)$, $u_2'(\pi/2)$.

In principle, the Hill functions may be found if the coefficients of the Fourier series in (49) are specified. Thus, we might expect that the value at $\xi = \pi/2$ would be expressible somehow in terms of these coefficients. This has indeed been found to be so [15, p. 34] and the results are

$$u_1(\pi/2) = \cos\left(\frac{\pi}{2}\sqrt{\lambda}\right) c_1 \quad (61)$$

$$u_2(\pi/2) = \frac{\sin\left(\frac{\pi}{2}\sqrt{\lambda}\right)}{\sqrt{\lambda}} s_0 \quad (62)$$

$$u_1'(\pi/2) = -\sqrt{\lambda} \sin\left(\frac{\pi}{2}\sqrt{\lambda}\right) c_0 \quad (63)$$

$$u_2'(\pi/2) = \cos\left(\frac{\pi}{2}\sqrt{\lambda}\right) s_1 . \quad (64)$$

s_0, s_1, c_0, c_1 are the one-sided infinite determinants

$$s_0 = \left| \delta_{n,m} + \frac{g_{n-m} - g_{n+1}}{\lambda - 4n^2} \right|_1^\infty \quad (65)$$

$$s_1 = \left| \delta_{n,m} + \frac{g_{n-m} - g_{n+m+1}}{\lambda - (2n+1)^2} \right|_0^\infty \quad (66)$$

$$c_0 = \left| \delta_{n,m} + \frac{(g_{n-m} + g_{n+m})(1 + \operatorname{sgn} n \operatorname{sgn} m)}{\sqrt{\epsilon_n \epsilon_m} (\lambda - 4n^2)} \right|_0^\infty \quad (67)$$

$$c_1 = \left| \delta_{n,m} + \frac{g_{n-m} + g_{n+m+1}}{\lambda - (2n+1)^2} \right|_0^\infty \quad (68)$$

where

$$\epsilon_n = \begin{cases} 1 & n = 0 \\ 2 & n > 0 \end{cases} \quad (69)$$

$$g_{-n} = \begin{cases} 0 & n = 0 \\ g_n & n > 0 \end{cases} \quad (70)$$

$$\delta_{n,m} = \begin{cases} 1 & n = m \\ 0 & n \neq m \end{cases} \quad (71)$$

$$\operatorname{sgn} n = \begin{cases} 0 & n = 0 \\ 1 & n > 0 \end{cases} \quad (72)$$

The Fourier coefficients λ and g_n are

$$g_n = \frac{L_1^2}{d} \int_0^d [\epsilon_r(z) - \sin^2 \theta_i] \cos \frac{n\pi z}{d} dz \quad (73)$$

$$\lambda = \frac{L_1^2}{d} \int_0^d [\epsilon_r(z) - \sin^2 \theta_i] dz, \quad (74)$$

where $\epsilon_r(z)$ is the relative permittivity in the layer, $\epsilon(z)/\epsilon_1$. Some insight as to the structure of the determinants may be gained by writing them as

$$S_0 = \left| \begin{array}{cccc} 1 - \frac{g_2}{\lambda - 4} & \frac{g_1 - g_3}{\lambda - 4} & \frac{g_2 - g_4}{\lambda - 4} & \frac{g_3 - g_5}{\lambda - 4} \\ \frac{g_1 - g_3}{\lambda - 16} & 1 - \frac{g_4}{\lambda - 16} & \frac{g_1 - g_5}{\lambda - 16} & \frac{g_2 - g_6}{\lambda - 16} \\ \frac{g_2 - g_4}{\lambda - 36} & \frac{g_1 - g_5}{\lambda - 36} & 1 - \frac{g_6}{\lambda - 36} & \frac{g_1 - g_7}{\lambda - 36} \\ \frac{g_3 - g_5}{\lambda - 64} & \frac{g_2 - g_6}{\lambda - 64} & \frac{g_1 - g_7}{\lambda - 64} & 1 - \frac{g_8}{\lambda - 64} \end{array} \right| \quad (75)$$

$$S_1 = \begin{vmatrix} 1 - \frac{g_1}{\lambda - 1} & \frac{g_1 - g_2}{\lambda - 1} & \frac{g_2 - g_3}{\lambda - 1} & \frac{g_3 - g_4}{\lambda - 1} & \cdot \\ \frac{g_1 - g_2}{\lambda - 9} & 1 - \frac{g_3}{\lambda - 9} & \frac{g_1 - g_4}{\lambda - 9} & \frac{g_2 - g_5}{\lambda - 9} & \cdot \\ \frac{g_2 - g_3}{\lambda - 25} & \frac{g_1 - g_4}{\lambda - 25} & 1 - \frac{g_5}{\lambda - 25} & \frac{g_1 - g_6}{\lambda - 25} & \cdot \\ \frac{g_3 - g_4}{\lambda - 49} & \frac{g_2 - g_5}{\lambda - 49} & \frac{g_1 - g_6}{\lambda - 49} & 1 - \frac{g_7}{\lambda - 49} & \cdot \\ \cdot & \cdot & \cdot & \cdot & \end{vmatrix} \quad (76)$$

$$C_0 = \begin{vmatrix} 1 & \frac{\sqrt{2} g_1}{\lambda} & \frac{\sqrt{2} g_2}{\lambda} & \frac{\sqrt{2} g_3}{\lambda} & \cdot \\ \frac{\sqrt{2} g_1}{\lambda - 4} & 1 + \frac{g_2}{\lambda - 4} & \frac{g_1 + g_3}{\lambda - 4} & \frac{g_2 + g_4}{\lambda - 4} & \cdot \\ \frac{\sqrt{2} g_2}{\lambda - 16} & \frac{g_1 + g_3}{\lambda - 16} & 1 + \frac{g_4}{\lambda - 16} & \frac{g_1 + g_5}{\lambda - 16} & \cdot \\ \frac{\sqrt{2} g_3}{\lambda - 36} & \frac{g_2 + g_4}{\lambda - 36} & \frac{g_1 + g_5}{\lambda - 36} & 1 + \frac{g_6}{\lambda - 36} & \cdot \\ \cdot & \cdot & \cdot & \cdot & \end{vmatrix} \quad (77)$$

$$C_1 = \begin{vmatrix} 1 + \frac{g_1}{\lambda - 1} & \frac{g_1 + g_2}{\lambda - 1} & \frac{g_2 + g_3}{\lambda - 1} & \frac{g_3 + g_4}{\lambda - 1} & \cdot \\ \frac{g_1 + g_2}{\lambda - 9} & 1 + \frac{g_3}{\lambda - 9} & \frac{g_1 + g_4}{\lambda - 9} & \frac{g_2 + g_5}{\lambda - 9} & \cdot \\ \frac{g_2 + g_3}{\lambda - 25} & \frac{g_1 + g_4}{\lambda - 25} & 1 + \frac{g_5}{\lambda - 25} & \frac{g_1 + g_6}{\lambda - 25} & \cdot \\ \frac{g_3 + g_4}{\lambda - 49} & \frac{g_2 + g_5}{\lambda - 49} & \frac{g_1 + g_6}{\lambda - 49} & 1 + \frac{g_7}{\lambda - 49} & \cdot \\ \cdot & \cdot & \cdot & \cdot & \end{vmatrix} \quad (78)$$

One should note that approximating the dielectric profile by truncating the Fourier series does not result in finite determinants. However, truncating the determinants then requires a knowledge of only a finite number of Fourier coefficients for their evaluation. The more basic question then is what size determinants to use in finding the reflection coefficient. In practice, however, one must approximate the actual permittivity profile so we shall study independently the effect upon the reflection coefficient due to truncating the Fourier series and truncating the infinite determinants. Later, some analytical details will be presented with regard to convergence. Presently, we continue with the Hill-function solution for the linear dielectric profile.

The coefficient in (48) for the linear profile is

$$g(z) = L_1^2 (az + \cos^2 \theta_i) , \quad 0 \leq z \leq d . \quad (79)$$

Fig. 3 shows the even periodic extension of this which is the function given by the Fourier series in Hill's equation. The two horizontal axes illustrate the change of variables in (46). The Fourier coefficients for this function are

$$\lambda = \left(\frac{\epsilon_r - 1}{2} + \cos^2 \theta_i \right) L_1^2 \quad (80)$$

$$g_n = \begin{cases} \frac{-2(\epsilon_r - 1)}{n^2 \pi^2} L_1^2 & n \text{ odd} \\ 0 & n \text{ even} . \end{cases} \quad (81)$$

Since $\sum_{n=1}^{\infty} \frac{1}{n^2}$ is absolutely convergent, the infinite determinants converge and the reflection coefficient may be found. This completes the Hill-function solution.

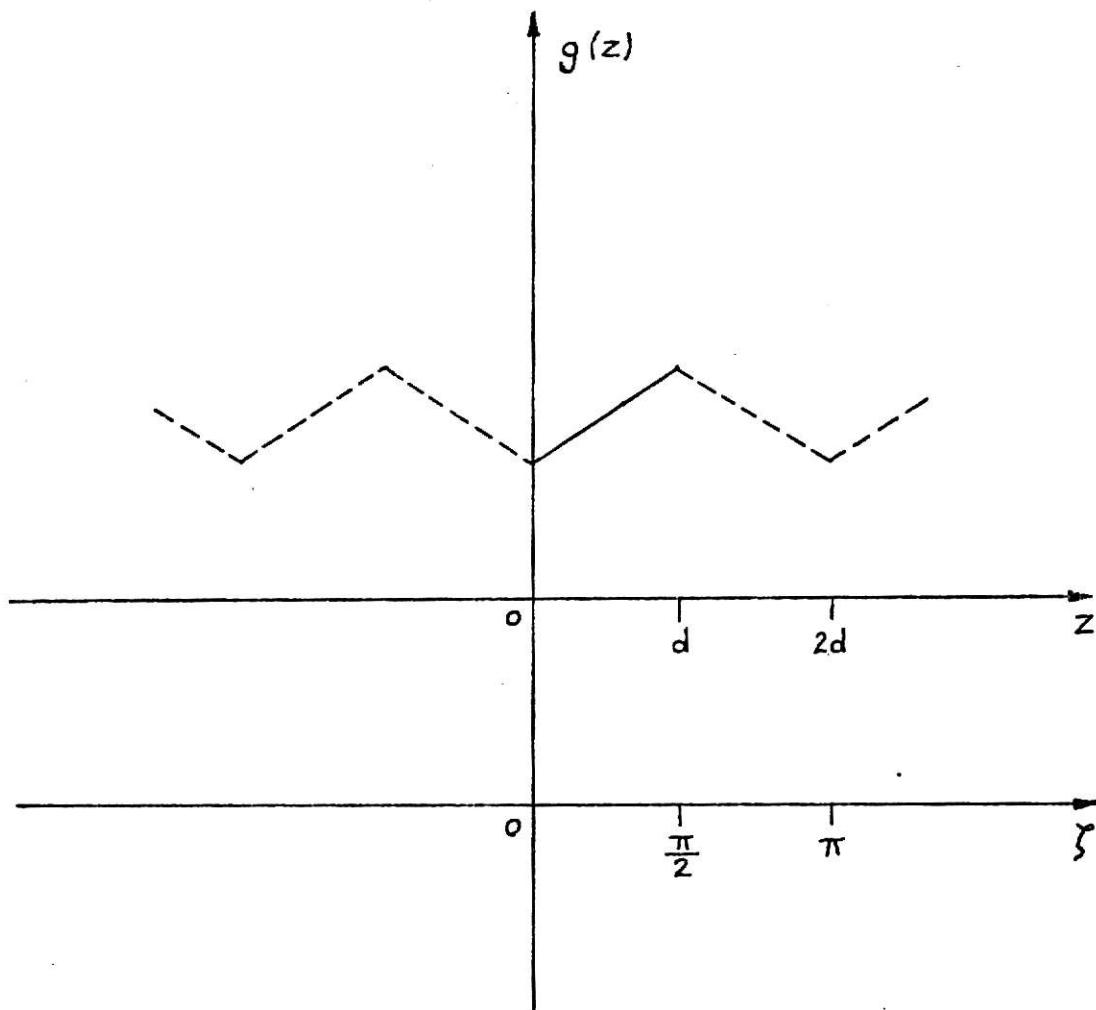


Fig. 3. Coefficient in Hill's equation represented by Fourier series.

V. NUMERICAL CALCULATION OF THE REFLECTION COEFFICIENT

From the Fourier series for $g(z)$ we obtain the series for the linear relative permittivity profile as

$$\epsilon_r(z) = \frac{\epsilon_r + 1}{2} - \frac{4}{\pi^2} (\epsilon_r - 1) \sum_{n=1}^{\infty} \frac{1}{(2n-1)^2} \cos \frac{(2n-1)\pi z}{d}. \quad (82)$$

Figs. (4a), (4b) show profiles approximating the linear profile obtained by truncating the series in (82). The superscript n in $\epsilon_r^{(n)}(z)$ gives the number of terms retained. Thus, $\epsilon_r^{(n)}(z)$ is the actual dielectric profile that is solved if one truncates the series for $g(z)$ in n non-zero terms.

In Figs. (5)-(7) the magnitude of the reflection coefficient is given as a function of either L_1 or θ_1 with ϵ_r and L_1 or θ_1 constant. Each figure consists of a family of five curves. One is the exact reflection coefficient obtained from the Airy-function solution. The remaining four are computed from the Hill-function solution using various degrees of profile approximation, namely $\epsilon_r^{(n)}(z)$ with $n = 1, 2, 3, 4$. For each curve using the Hill-function solution, the computing procedure was as follows: The infinite determinants were truncated to size 3x3 and the reflection coefficient calculated. The determinant size was incremented by one until the computed reflection coefficient differed by less than 0.5% from the one preceding. The final values of the reflection coefficient R and the determinant size NS were recorded. For each successive data point, the initial determinant size was set at one less than the final size of the determinant for the preceding point. For Fig. (5), calculations were begun at $L_1 = 0.2$, and for Figs. (6), (7), at $\theta_1 = 0$.

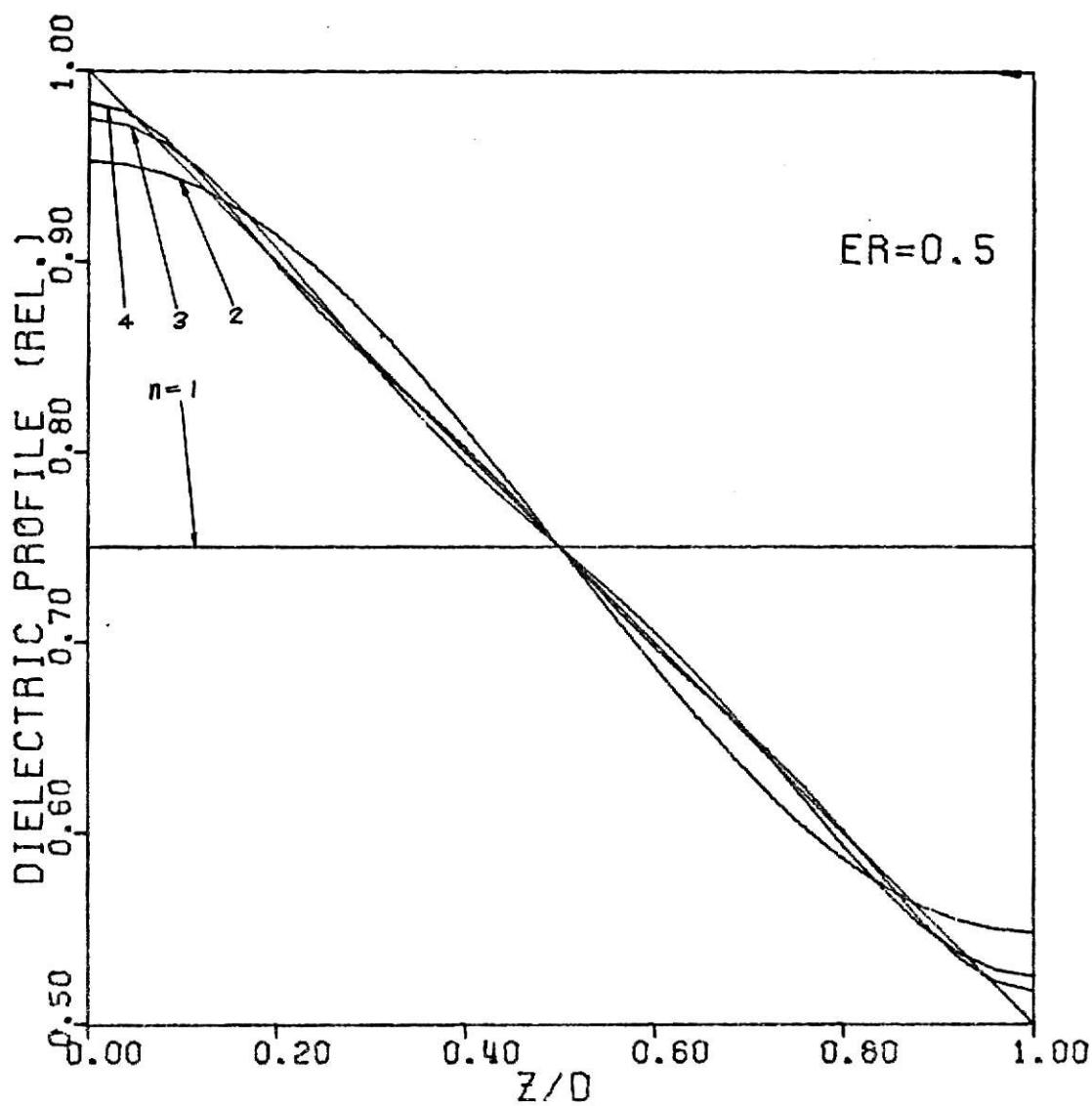


Fig. 4a. Dielectric profile represented by truncated Fourier series, $\epsilon_r = 0.5$.

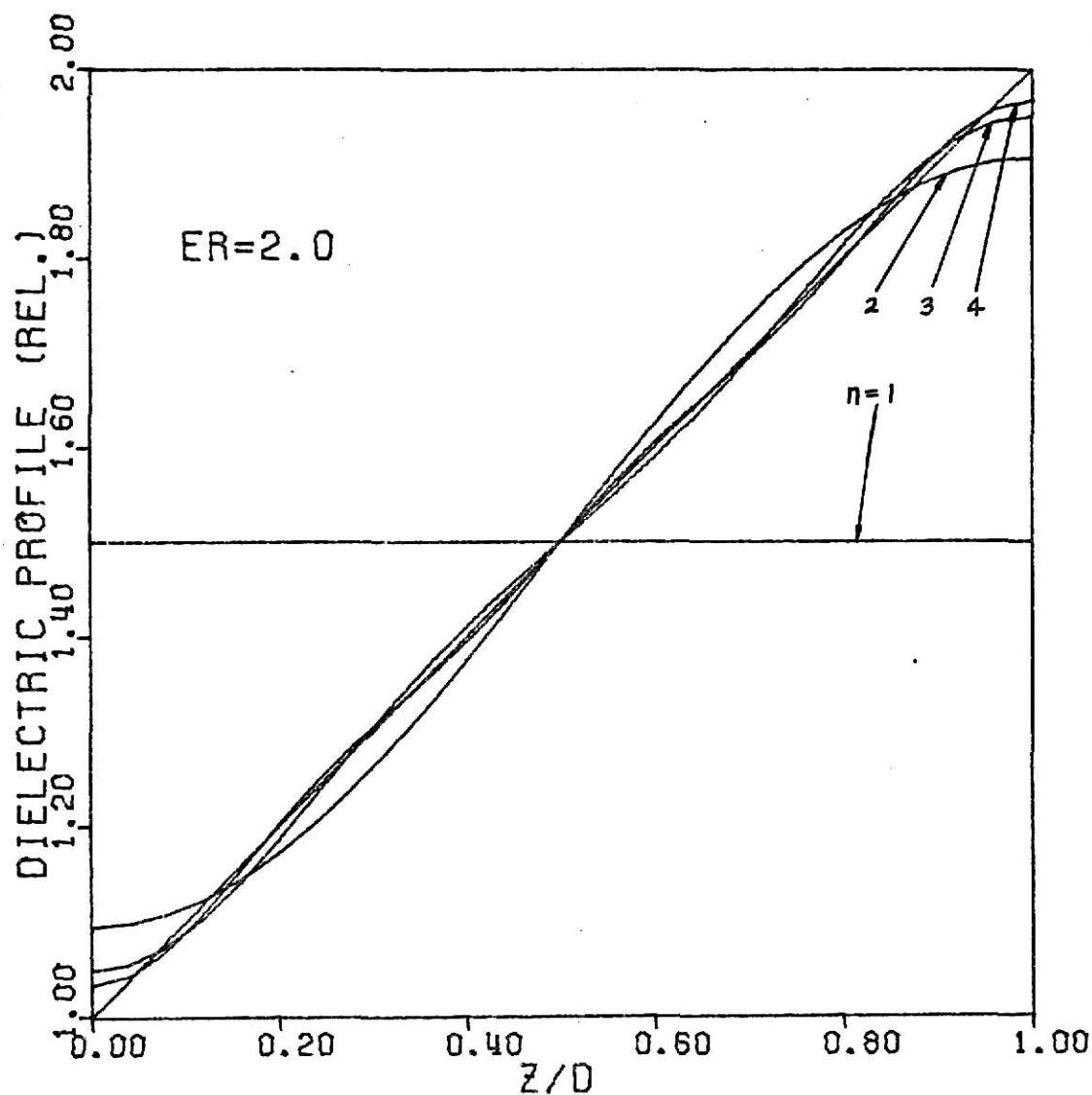


Fig. 4b. Dielectric profile represented by truncated Fourier series, $\epsilon_r = 2.0$.

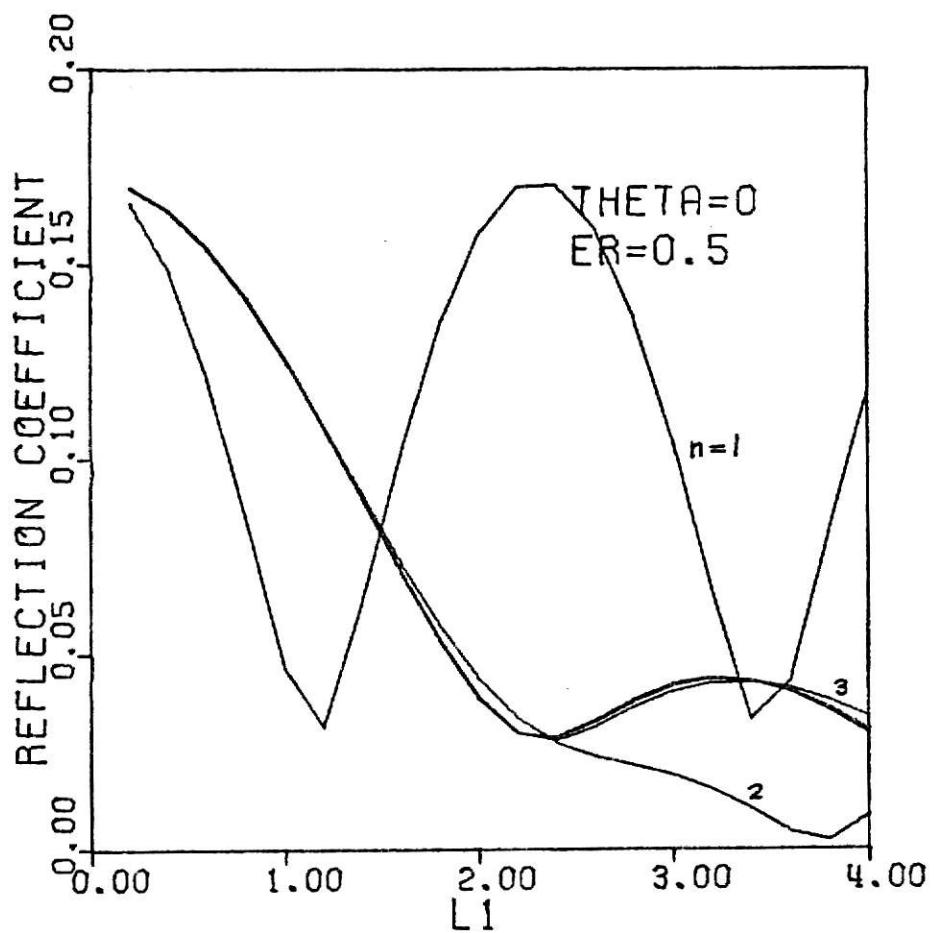


Fig. 5a. Reflection coefficient vs. L_1 , $\theta_i = 0$, $\epsilon_r = 0.5$.

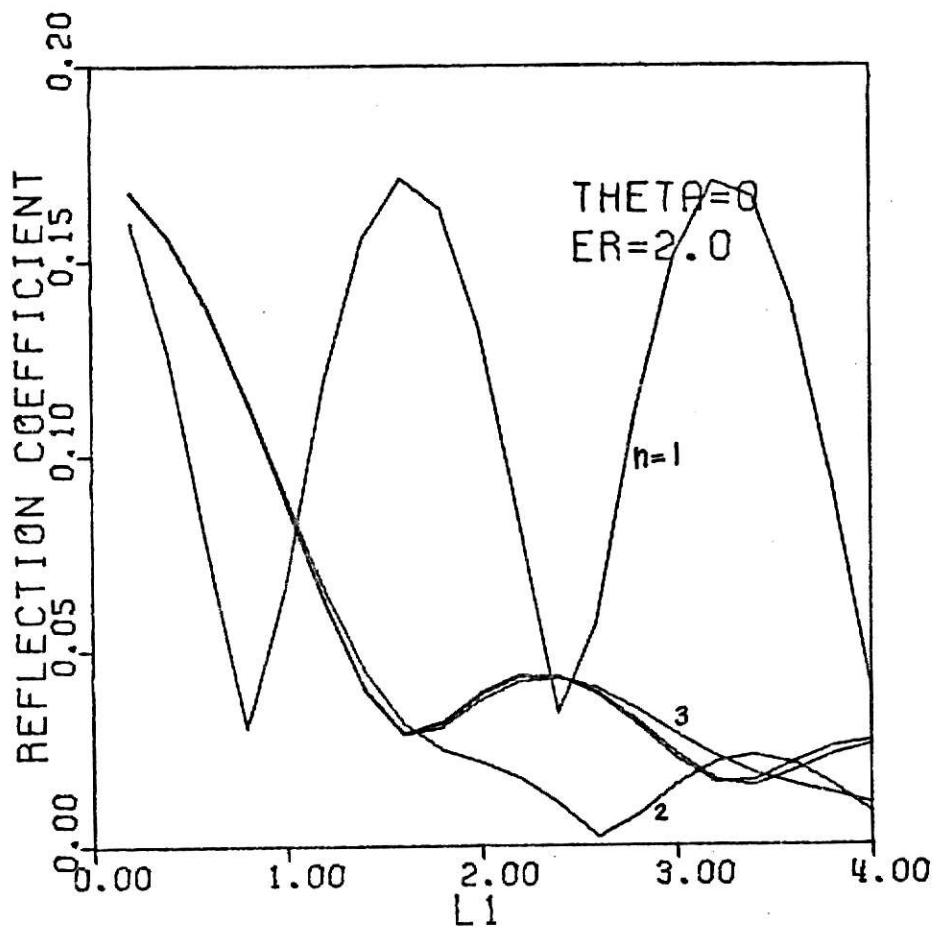


Fig. 5b. Reflection coefficient vs. L_1 , $\theta_i = 0$,
 $\epsilon_r = 2.0$.

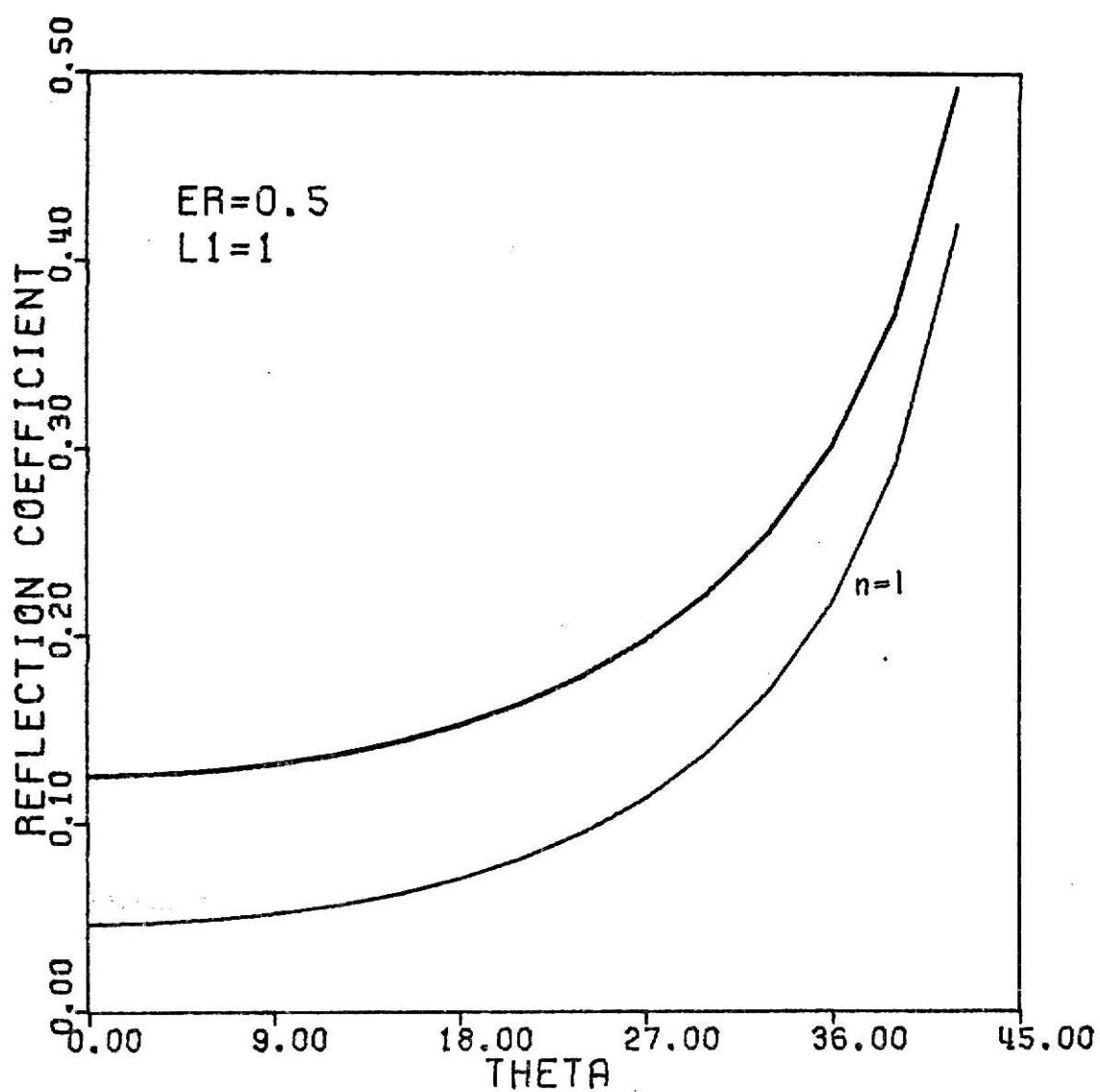


Fig. 6a. Reflection coefficient vs. θ_i , $\epsilon_r = 0.5$, $L_1 = 1$.

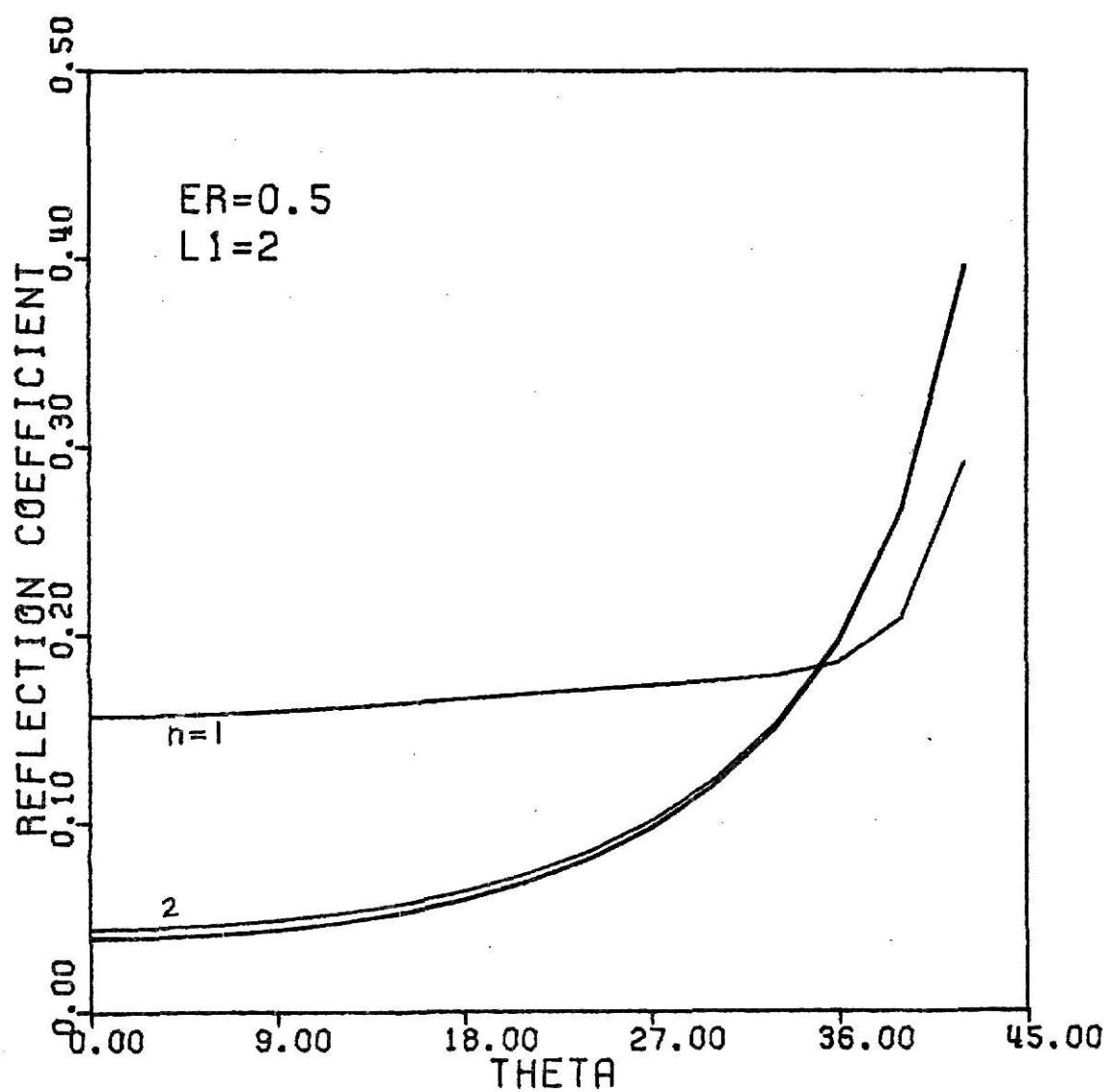


Fig. 6b. Reflection coefficient vs. θ_i , $\epsilon_r = 0.5$, $L_1 = 2$.

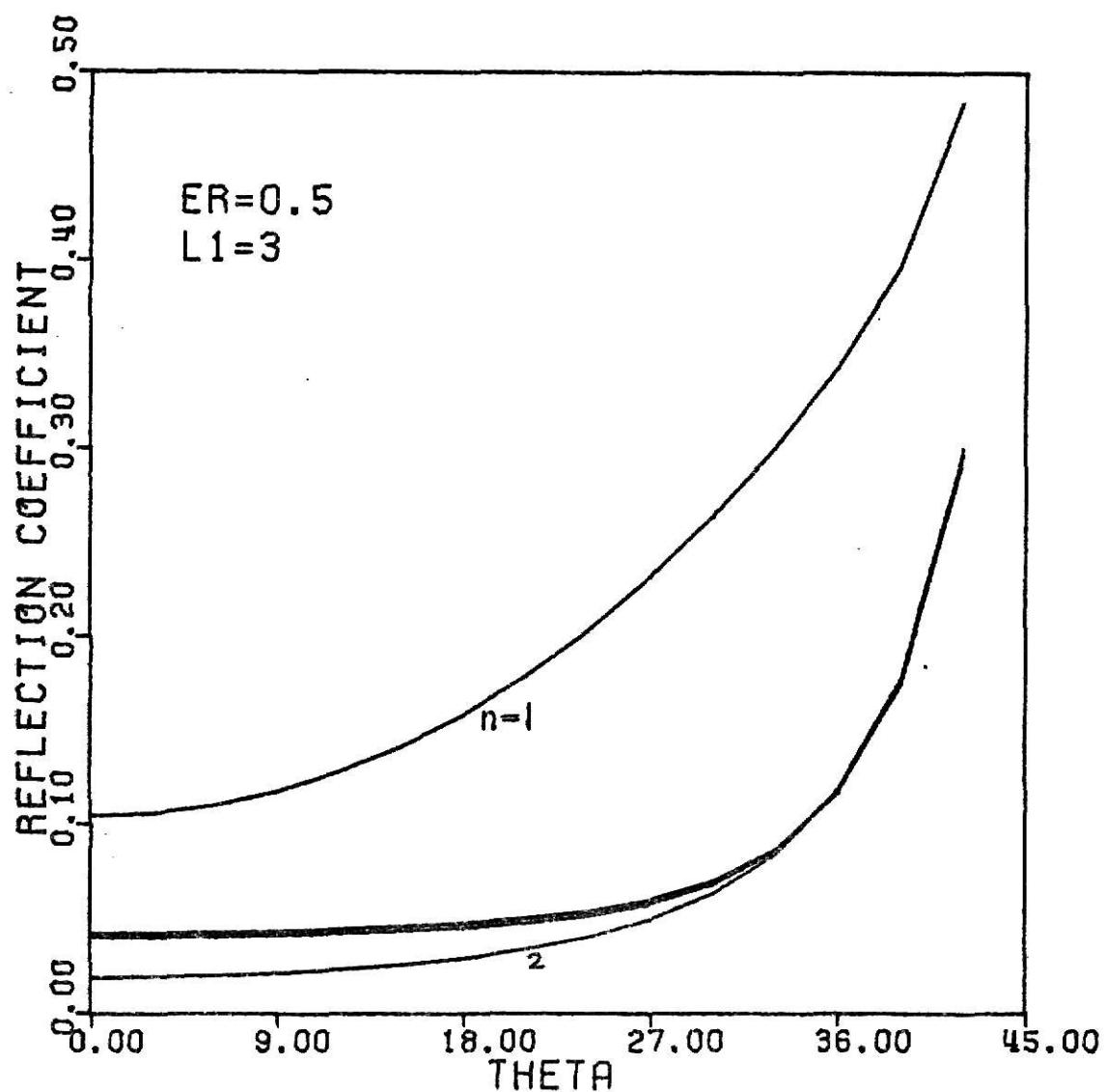


Fig. 6c. Reflection coefficient vs. θ_1 , $\epsilon_r = 0.5$, $L_1 = 3$

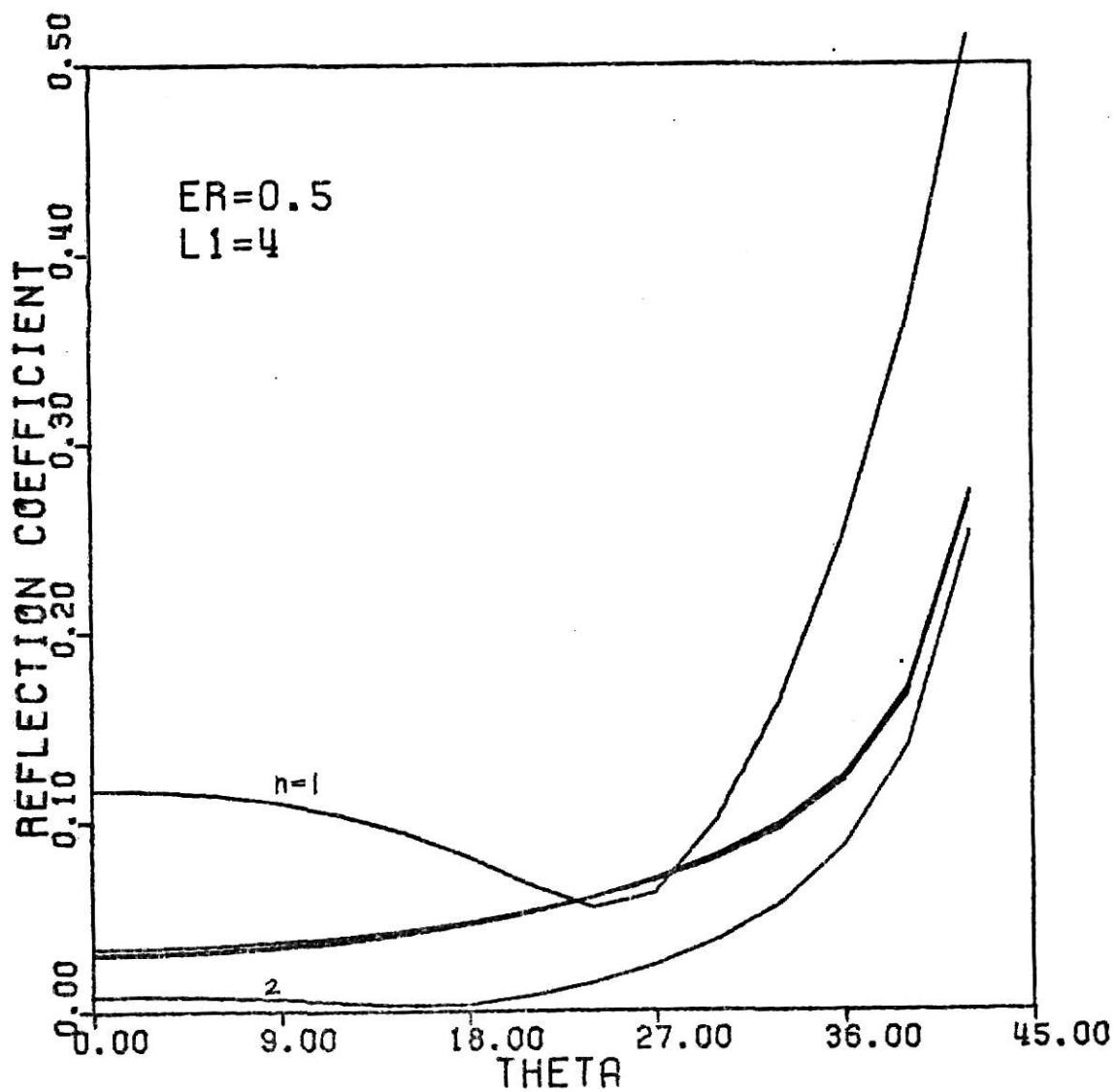


Fig. 6d. Reflection coefficient vs. θ_i , $\epsilon_r = 0.5$, $L_1 = 4$.

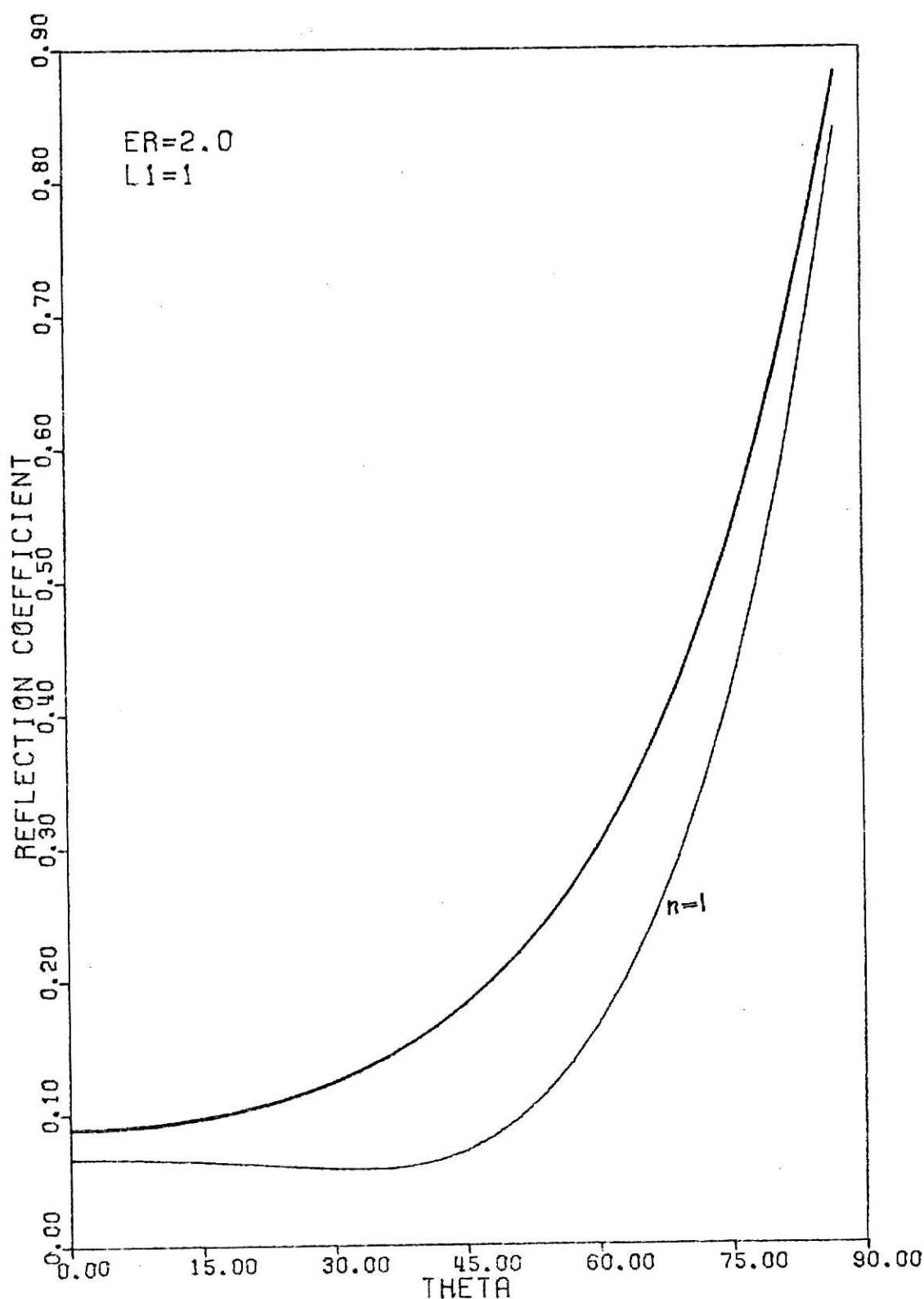


Fig. 7a. Reflection coefficient vs. θ_1 , $\epsilon_r = 2.0$, $L_1 = 1$.

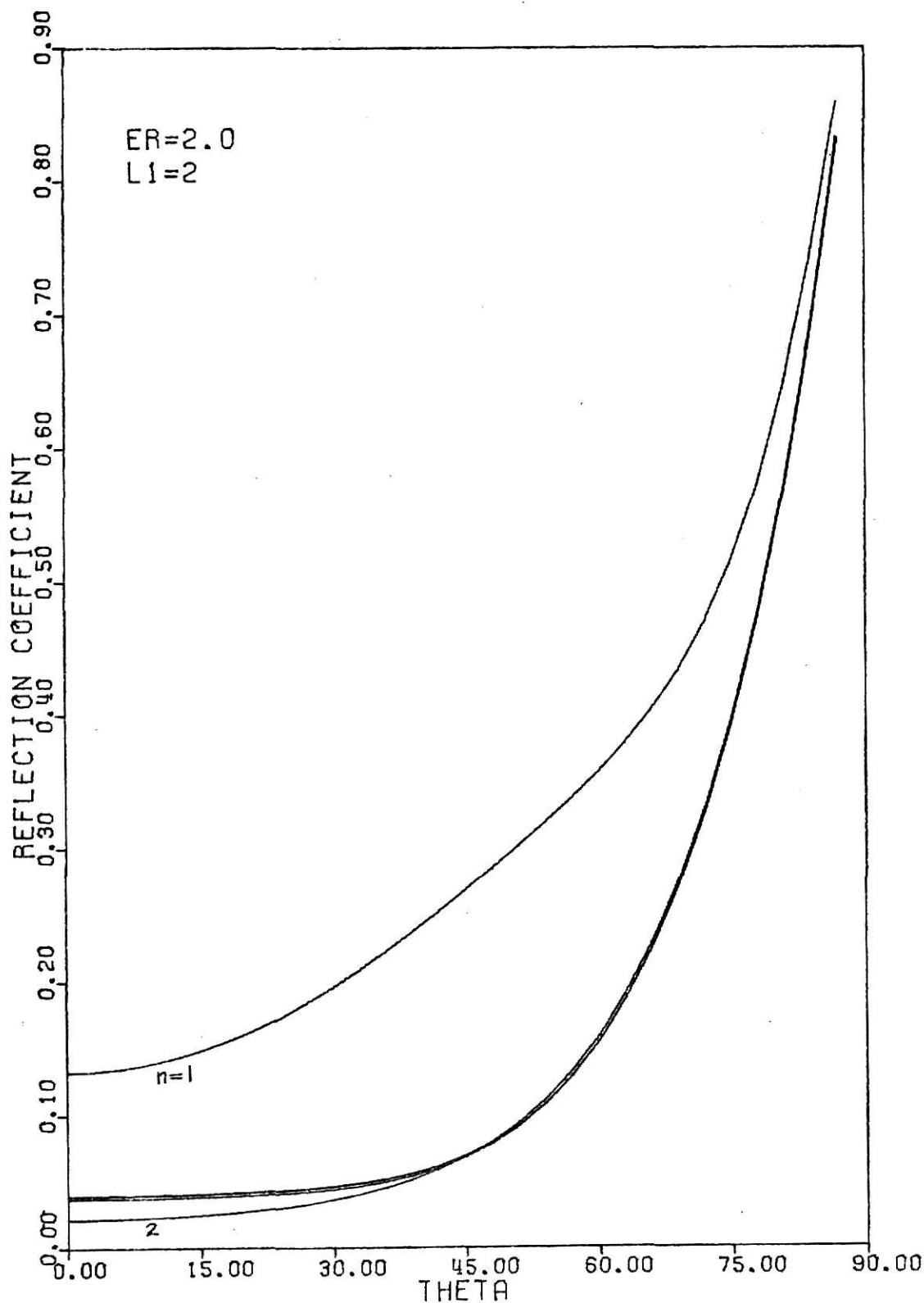


Fig. 7b. Reflection coefficient vs. θ_i , $\epsilon_r = 2.0$, $L_1 = 2$.

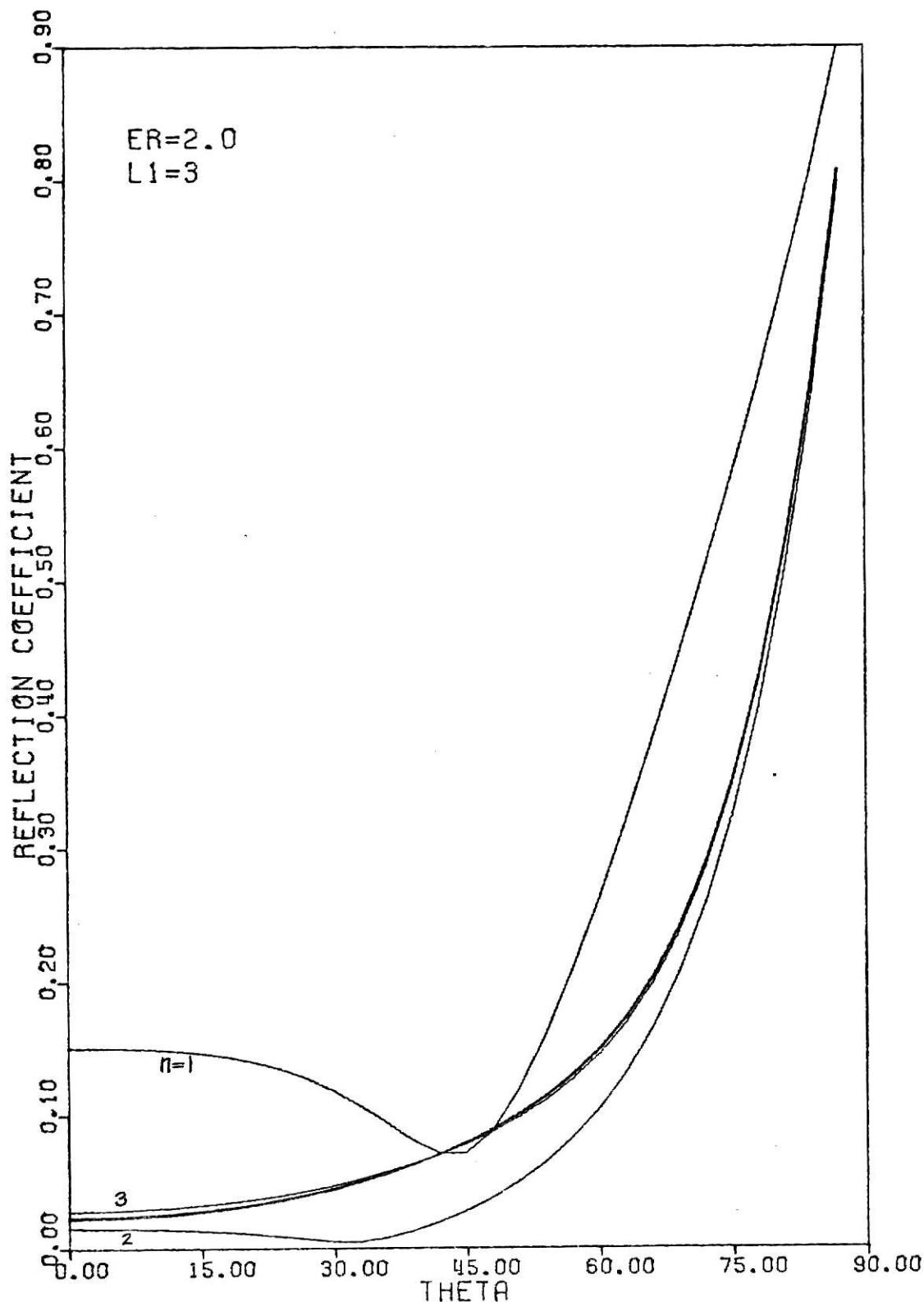


Fig. 7c. Reflection coefficient vs. θ_i , $\epsilon_r = 2.0$, $L_1 = 3$.

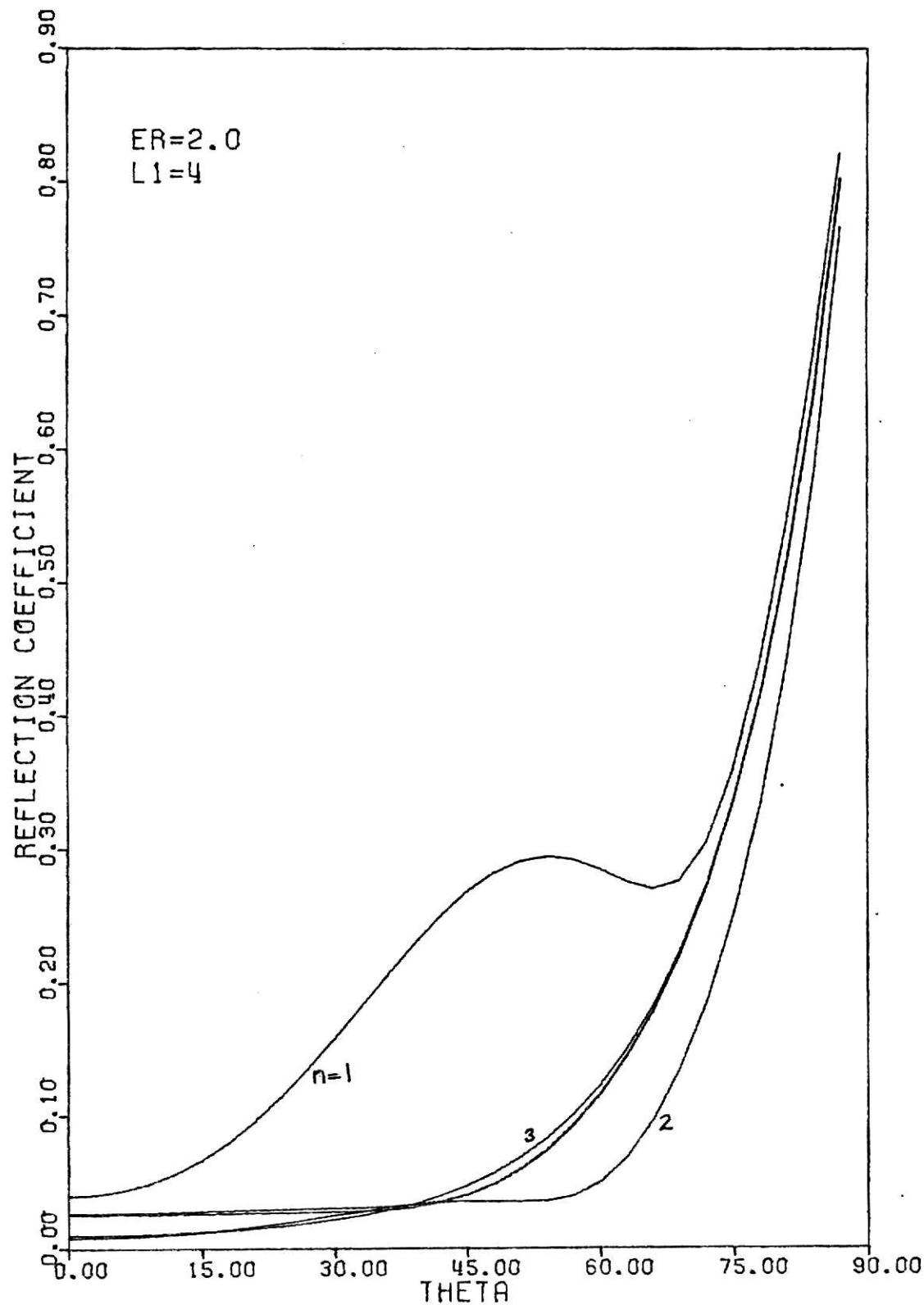


Fig. 7d. Reflection coefficient vs. θ_1 , $\epsilon_r = 2.0$, $L_1 = 4$.

The criterion of 0.5% change in R gives no clue as to the accuracy of the final R. The idea here was that if the determinants have converged close to their exact value for a given NS, then so has R, and therefore R cannot change by more than some small amount if NS is increased. The problem is that the change in R can also be small if R is converging slowly, with a value perhaps far from the exact value. The error can therefore vary over a wide range, and this is evident when comparing the reflection coefficient for $n = 4$ to the exact R. Fortunately, however, the error was small in all but a few places, a notable one being for $\epsilon_r = 2$, $L_1 = 4$, and $\theta_i = 30^\circ$ (Fig. 7d). This was roughly the point of maximum error, about 10%. Here the size of the determinant would have to be made greater by one or two to reduce the error to a more typical value of 0.5%. The range of the error as obtained from reflection coefficient data (not shown) is about 0.1% - 10%.

Since the error was reasonably uniform, the behavior of convergence with respect to the parameters θ_i , L_1 , and ϵ_r is clear. In Tables Ia, Ib, we show the range of final determinant size in the computations for each combination of L_1 and $\epsilon_r^{(n)}$. It is seen that larger NS is required as L_1 increases. From (80), (81), if $(\epsilon_r - 1)/2 \gg \cos^2 \theta_i$, then the determinants behave similarly as a function of the quantity $(\epsilon_r - 1)L_1^2$. Since L_1 is the length of the transition region in quarter wavelengths of region (1), we may state that the convergence is most rapid when the change in permittivity is small and occurs through a region thin with respect to wavelength. Also, Tables Ia, Ib indicate that convergence is not strongly dependent upon θ_i since most combinations of L_1 and n have a single value for the determinant sizes.

Now, consider an off-diagonal element of S_o or C_o of the form $(g_p + g_q)/(\lambda - 4n^2)$, as in (75), (77), where n is the row of the element. For the linear profile this is

$n \backslash L_1$	1	2	3	4
n				
1	4	4	4	4
2	4	4	5	7-9
3	4	4	4-5	6
4	4	4	4-5	6

(a)

$n \backslash L_1$	1	2	3	4
n				
1	4	4	4	4
2	4	5	6-9	9
3	4	4	6	9
4	4	5	6	6-7

(b)

TABLE I. Range of final determinant size in the Hill-function computations vs. L_1 and n ; a) $\varepsilon_r = 0.5$, b) $\varepsilon_r = 2.0$.

$$\frac{g_p + g_q}{\lambda - 4n^2} = \frac{-2(\epsilon_r - 1)L_1^2 \left(\frac{1}{p^2} + \frac{1}{q^2} \right)}{\left(\frac{(\epsilon_r - 1)}{2} + \cos^2 \theta_i \right) L_1^2 - 4n^2} \quad (83)$$

where

$$0 \leq \left(\frac{1}{p^2} + \frac{1}{q^2} \right) \leq \frac{5}{4}. \quad (84)$$

If the frequency is such that the magnitude of the L_1^2 term in the denominator is much greater than $4n^2$ then all elements of rows less than n are nearly frequency independent. As L_1 is made smaller the $4n^2$ term will dominate and more elements will go as L_1^2 . Those that have been going as L_1^2 will be small compared to the constant elements so that we may truncate the determinant by disregarding those elements for which

$$\left| \left(\frac{(\epsilon_r - 1)}{2} + \cos^2 \theta_i \right) L_1^2 \right| \ll 4n^2. \quad (85)$$

With $\epsilon_r = 2$ we have

$$\frac{1}{2} \leq \frac{(\epsilon_r - 1)}{2} + \cos^2 \theta_i \leq \frac{3}{2} \quad (86)$$

so that we write (85) as

$$L_1^2 = \frac{4n^2}{36}. \quad (87)$$

This gives the truncated determinant size as

$$NS = 3L_1. \quad (88)$$

This result is valid for S_1 and C_1 also, and could have been obtained by considering their elements. It compares well with the empirical result $NS \sim 2L_1$, obtained by inspection of Table Ib. It is significant that the required determinant size is directly proportional to L_1 . This result may be extended to an arbitrary profile by noting that the general Fourier coefficients are proportional to L_1^2 , as shown in (73), (74).

The effect upon the reflection coefficient due to truncating the Fourier series for $g(z)$ is shown by Figs. (5)-(7). It is seen that for a given number of terms retained, more accurate values of R are obtained for smaller L_1 . One should note, especially from Fig. (5), that keeping only the first term gives very poor results even for small L_1 and that a substantial improvement is effected by adding a second term. This could be expected after comparing the dielectric profile approximations in Fig. (4).

If the infinite determinants are truncated to size NS , then the highest Fourier coefficient needed for their evaluations is g_{2NS} . It follows then that for small L_1 the dielectric profile may be approximated with fewer terms of the Fourier series. However, the converse is not true. That is, the determinants for a profile which can accurately be represented by few terms of the series do not necessarily converge faster than if the profile requires a relatively larger number of terms. This is because keeping a finite number of series terms still gives infinite determinants whose rate of convergence depends upon the size and placement of the non-zero elements. Examples of faster convergence with more series terms may be found in Tables Ia, Ib.

Actually there is no reason to be concerned with truncating the Fourier series since one may always obtain the number of terms required to evaluate the truncated determinants. This is obvious if the profile, or the coefficient $g(z)$, is such that the integrals in (73), (74) can be evaluated directly. If not, then graphical or numerical integration may be used. An alternate approach is the discrete Fourier transform (DFT). The DFT requires $n+1$ values of the function at equally-spaced points in order to obtain $n+1$ values approximating the first $n+1$ Fourier coefficients. These values converge to the actual Fourier coefficients for large n . In our case then, we require the value of $\epsilon_r(z)$ at the points $z = md/n$, $m = 0, 1, 2, \dots, n$, to find the

Fourier coefficients λ through g_n . Using the fast Fourier transform (FFT) algorithm allows one to compute the DFT coefficients very economically for large n .

Although the Hill-function solution for the reflection coefficient is formally exact, the accuracy of results obtained by its implementation is limited by economics. In Table IIa, we give a summary of the cost and time required for computing the Hill-function data used for this paper. It should be noted that we have tabulated the total number of reflection coefficient computations. This is slightly greater than twice the number of points used to construct Figs. (5)-(7) due to the computing procedure described earlier. In Table IIb the cost and execution time for the Airy-function computations is summarized. A comparison shows that for the accuracy obtained with the Hill-function solution and the range of L_1 used, the Hill-function method cost about one-fifth as much per reflection coefficient as did the Airy-function solution. At higher frequencies, or larger L_1 , larger determinants will have to be evaluated, and since the work required for determinant evaluation goes up as the cube of the size [16], the cost will rise sharply. This limits the usefulness of the Hill-function method, as applied here, to transition regions with a thickness of no more than a wavelength or so.

A method of reducing the cost of finding R that was not investigated numerically, but would appear significant, is now described. We break up the transition region into n layers as shown in Fig. 8, with the solution for the m^{th} layer written as

$$u^{(m)} = A_m u_1^{(m)} + B_m u_2^{(m)} . \quad (89)$$

The application of boundary conditions yields $2(n+1)$ equations for the unknowns R , T , A_m , B_m , $m = 1, 2, \dots, n$, and we may solve these for R in terms

NS	3	4	5	6	7	8	9
No. of R's	677	830	318	195	52	113	107
Total No. of R's	2292						
Execution time	6.78 minutes						
Cost	20.34 dollars						

(a)

Total No. of R's	280
Execution time	4.08 minutes
Cost	12.24 dollars

(b)

TABLE II. Computations summary for a) Hill-function solution,
b) Airy-function solution.

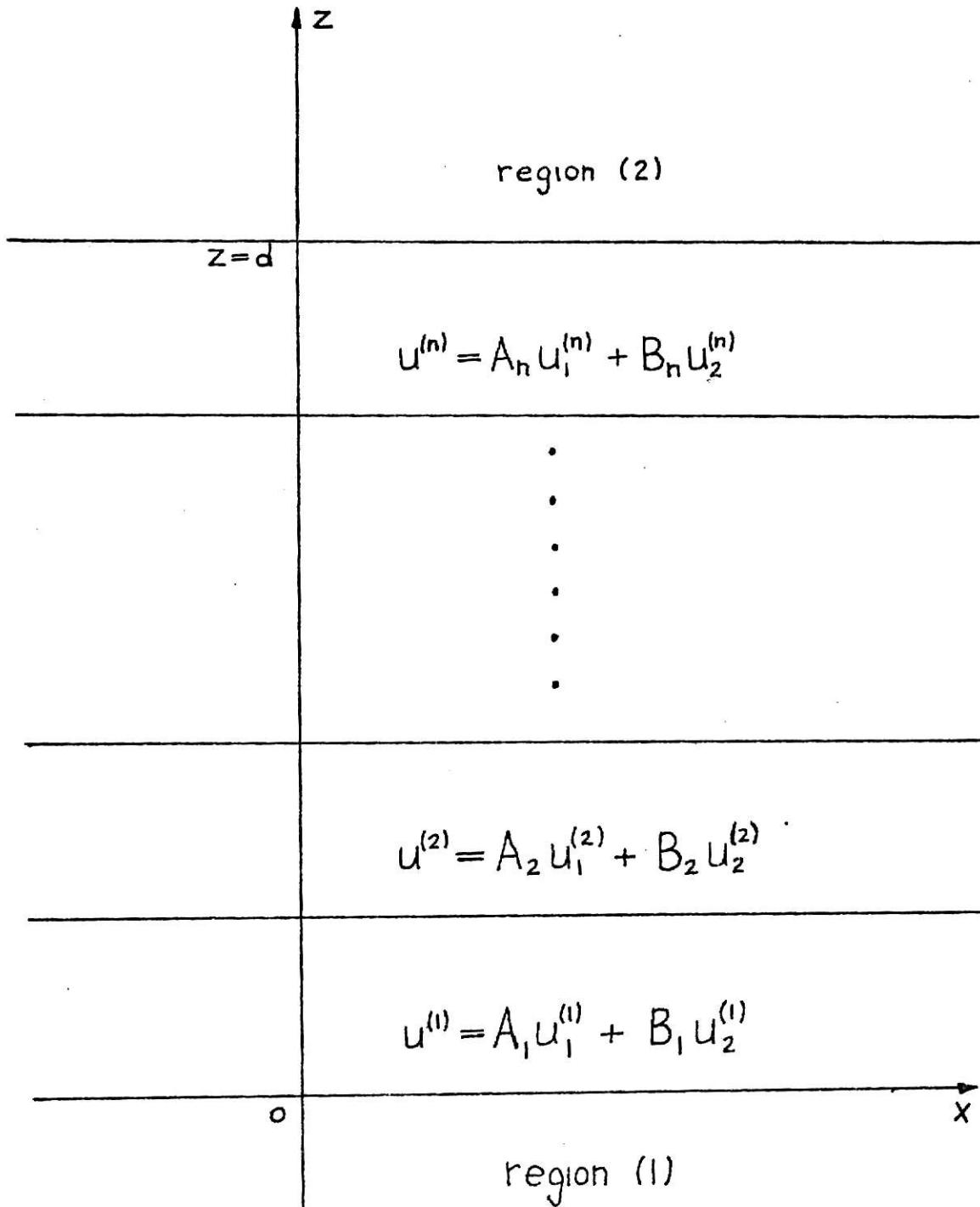


Fig. 8. Geometry of transition region divided into n layers, with the respective Hill function for each layer.

of $u_1^{(m)}(\pi/2)$, $u_2^{(m)}(\pi/2)$, $u_1'^{(m)}(\pi/2)$, $u_2'^{(m)}(\pi/2)$, $m = 1, 2, \dots, n$. We now compare the cost for finding R by using a single transition region of thickness L_1 and by dividing the layer into n regions of thickness L_1/n . From (88), the necessary truncated determinant size is $3L_1$ so that for the single transition region we must compute four determinants of size $3L_1$. For the n layers, we must compute $4n$ determinants of size $3L_1/n$. Remembering that the work required for determinant computation goes as the size cubed, we find the cost ratio of the two methods is

$$\frac{4(3L_1)^3}{4n\left(\frac{3L_1}{n}\right)^3} = n^2. \quad (90)$$

Thus, by dividing the transition layer into n equal parts, we reduce the cost of evaluating the necessary determinants by a factor of $1/n^2$. We expect a point of diminishing returns, however, since one must solve $2n + 2$ simultaneous equations for R . Also, more work is required in setting up the problem, for example finding the Fourier coefficients for each of the n regions.

The computer subroutines will not be examined in detail, but are presented in the Appendix along with a brief description.

VI. ANALYTICAL CONSIDERATIONS

If L_1 is small, we may approximate the Bessel differential equation, (24), as

$$v'' + \frac{v'}{w} - \frac{v}{9w^2} = 0 . \quad (91)$$

Its solution is

$$v = C\left(\frac{w}{2}\right)^{1/3} + D\left(\frac{w}{2}\right)^{-1/3} , \quad (92)$$

which gives

$$\psi = w^{1/3} [C\left(\frac{w}{2}\right)^{1/3} + D\left(\frac{w}{2}\right)^{-1/3}] . \quad (93)$$

After applying boundary conditions we obtain

$$R = \frac{w_d^{2/3} - w_o^{2/3} + i 2/3(w_d^{-1/3} - w_o^{-1/3})}{w_d^{2/3} - w_o^{2/3} + i 2/3(w_d^{-1/3} + w_o^{-1/3})} \quad (94)$$

$$= \frac{(\epsilon_r - 1)(\epsilon_r - \sin^2 \theta_i)^{1/2} \cos \theta_i + i \frac{2(\epsilon_r - 1)}{L_1 \pi} [\cos \theta_i - (\epsilon_r - \sin^2 \theta_i)^{1/2}]}{(\epsilon_r - 1)(\epsilon_r - \sin^2 \theta_i)^{1/2} \cos \theta_i + i \frac{2(\epsilon_r - 1)}{L_1 \pi} [\cos \theta_i + (\epsilon_r - \sin^2 \theta_i)^{1/2}]} . \quad (95)$$

If the frequency is low enough the higher power terms may be dropped giving

$$R = \frac{w_d^{-1/3} - w_o^{-1/3}}{w_d^{-1/3} + w_o^{-1/3}} = \frac{\cos \theta_i - (\epsilon_r - \sin^2 \theta_i)^{1/2}}{\cos \theta_i + (\epsilon_r - \sin^2 \theta_i)^{1/2}} , \quad (96)$$

the abrupt boundary reflection coefficient. To see how good the approximation in (94) is we expand the Bessel functions and form the products in (36).

Keeping the first two terms of the series expansion gives generally,

$$J_u^d J_v^o = \frac{1}{u! v!} \left(\frac{w_d}{2}\right)^u \left(\frac{w_o}{2}\right)^v - \frac{1}{u! (1+v)!} \left(\frac{w_d}{2}\right)^u \left(\frac{w_o}{2}\right)^{2+v} - \frac{1}{v! (1+u)!} \left(\frac{w_o}{2}\right)^v \left(\frac{w_d}{2}\right)^{2+u} . \quad (97)$$

Substituting this into (36) gives all terms containing the first four powers of L_1 . Keeping only those of -1 and 0 power yields

$$\begin{aligned}
 & -w_d^{-1/3} w_o^{1/3} + w_d^{1/3} w_o^{-1/3} + i 2/3 w_d^{-2/3} w_o^{-1/3} - 1/2 w_d^{2/3} w_o^{-2/3} \\
 R = & \frac{-i 2/3 w_d^{-1/3} w_o^{-2/3} + 1/2 w_d^{-2/3} w_o^{2/3}}{-w_d^{-1/3} w_o^{1/3} + w_d^{1/3} w_o^{-1/3} + i 2/3 w_d^{-2/3} w_o^{-1/3} + 1/2 w_d^{2/3} w_o^{-2/3}} \quad (98) \\
 & + i 2/3 w_d^{-1/3} w_o^{-2/3} - 1/2 w_d^{-2/3} w_o^{2/3} \\
 & - (\epsilon_r - \sin^2 \theta_i)^{1/2} \cos^3 \theta_i + (\epsilon_r - \sin^2 \theta_i)^{3/2} \cos \theta_i - 1/2 (\epsilon_r - \sin^2 \theta_i)^2 \\
 & + 1/2 \cos^4 \theta_i + \frac{i 2(\epsilon_r - 1)}{L_1 \pi} [\cos \theta_i - (\epsilon_r - \sin^2 \theta_i)^{1/2}] \\
 = & \frac{-(\epsilon_r - \sin^2 \theta_i)^{1/2} \cos^3 \theta_i + (\epsilon_r - \sin^2 \theta_i)^{3/2} \cos \theta_i + 1/2 (\epsilon_r - \sin^2 \theta_i)^2}{-(\epsilon_r - \sin^2 \theta_i)^{1/2} \cos^3 \theta_i + (\epsilon_r - \sin^2 \theta_i)^{3/2} \cos \theta_i + 1/2 (\epsilon_r - \sin^2 \theta_i)^2} \\
 & - 1/2 \cos^4 \theta_i + \frac{i 2(\epsilon_r - 1)}{L_1 \pi} [\cos \theta_i + (\epsilon_r - \sin^2 \theta_i)^{1/2}] \quad (99)
 \end{aligned}$$

If the 0 power terms are dropped here, we again have the abrupt boundary reflection coefficient, so that (98) may be considered to contain the first-order correction to the abrupt boundary coefficient. Since this expression is somewhat more complex than (94), we may assume that the reflection coefficient obtained by simplifying the differential equation directly is a rather crude approximation.

A first-order approximation identical to (98) may be easily obtained from the Hill-function formulation. Using (61)-(64) we have for the reflection coefficient of (60)

$$R = \frac{\tan(\frac{\pi}{2} \sqrt{\lambda}) \left(\frac{S_o}{\sqrt{\lambda}} a_o a_d - \sqrt{\lambda} c_0 \right) + i L_1 [\cos \theta_i s_1 - (\epsilon_r - \sin^2 \theta_i)^{1/2} c_1]}{\tan(\frac{\pi}{2} \sqrt{\lambda}) \left(\frac{S_o}{\sqrt{\lambda}} a_o a_d + \sqrt{\lambda} c_0 \right) + i L_1 [\cos \theta_i s_1 + (\epsilon_r - \sin^2 \theta_i)^{1/2} c_1]} \quad (100)$$

Let $\tan \left(\frac{\pi}{2} \sqrt{\lambda} \right) \approx \frac{\pi}{2} \sqrt{\lambda}$ and recall that $a_0, a_d, \sqrt{\lambda} \propto L_1$. In order to avoid higher powers of L_1 in R, we set the four infinite determinants equal to one, their limit as $L_1 \rightarrow 0$. Under these conditions, (100) may be shown to be identical to the first-order Bessel function approximation in (98).

To get a better approximation with higher powers of L_1 , we may expand (75)-(78) and obtain

$$S_0 = 1 - \sum_{n=1}^{\infty} \frac{g_{2n}}{\lambda - 4n^2} + \dots \quad (101)$$

$$S_1 = 1 - \sum_{n=1}^{\infty} \frac{g_{2n-1}}{\lambda - (2n-1)^2} + \dots \quad (102)$$

$$C_0 = 1 + \sum_{n=1}^{\infty} \frac{g_{2n}}{\lambda - 4n^2} - \frac{2}{\lambda} \sum_{n=1}^{\infty} \frac{g_n^2}{\lambda - 4n^2} + \dots \quad (103)$$

$$C_1 = 1 + \sum_{n=1}^{\infty} \frac{g_{2n-1}}{\lambda - (2n-1)^2} + \dots \quad (104)$$

At low frequencies λ is small so L_1^2 may be factored from the numerator of the series shown explicitly. Terms with higher powers of L_1 may be found by factoring subsequent infinite-series terms in (101)-(104). These further terms are products of g_n 's and are complicated due to the many combinations possible.

VII. SUMMARY

The problem of finding the electromagnetic fields outside an inhomogeneous, stratified layer for TE plane-wave incidence has been solved in terms of Hill functions. For a region with a linear dielectric profile we have demonstrated the utility of the Hill-function formulation by direct comparison with the Bessel function solution. For a transition region with thickness on the order of a wavelength or less, the Hill-function method was substantially cheaper for obtaining accurate numerical results for the reflection coefficient. At higher frequencies, convergence of infinite determinants becomes slower and limits the usefulness of the Hill-function solution. However, it was shown that breaking up the transition region into subregions can reduce the cost of finding R at higher frequencies. At low frequencies analytical approximations for R were readily obtained by simply expanding the suitably truncated determinants.

APPENDIX

The subroutines used for numerical evaluation of R by the Hill-function method are now presented. The subroutines HY1, HY1P, HY2, HY2P are used to compute $u_1(\pi/2)$, $u'_1(\pi/2)$, $u_2(\pi/2)$, $u'_2(\pi/2)$ respectively. These quantities are denoted by the variables Y1, Y1P, Y2, Y2P. ALAM is the Fourier coefficient λ , NS is the determinant size, and G is a vector consisting of the Fourier coefficients g_1, g_2, \dots, g_{60} . The subroutines CDET and CBIG are called by the above subroutines and actually perform the determinant evaluation by the Gauss elimination method.

For evaluating the known-function solution, three subroutines are used. The subroutine AIRY computes $Ai(z)$, $Ai'(z)$, $Bi(z)$, $Bi'(z)$ for the real argument z. In the IO list for this subroutine we have the argument Z for the above functions, respectively, AI, AIP, BI, BIP. The subroutines BESI, BESJ are called by AIRY.

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DOCUMENT(S) IS OF
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$JOB          HY1
SUBROUTINE HY1(ALAM,G,NS,Y1)
C   THIS SUBROUTINE COMPUTES THE ARRAY CMAT
C   CORRESPONDING TO THE DETERMINANT D1.
COMPLEX CSRL,CFACT
COMPLEX CSRL,CFACT
DIMENSION CMAT(31,31),G(60)
NCHK=0
PI=3.141593
C   CONSIDER THE N-TH ROW OF THE DETERMINANT.
DO 1 N=1,NS
FREAL=ALAM-(2*N-1)**2
ELEMENTS OF THE N-TH ROW HAVE A COMMON POLE.
CHECK TO SEE IF THE ELEMENTS ARE NEAR THE POLE.
IF SO, MULTIPLY THIS ROW BY THE APPROPRIATE
FUNCTION, WHICH WILL HAVE A ZERO HERE.
IF(ABS(FN).LE.1.0E-04) GO TO 2
DEFINE THE ELEMENT IN THE M-TH COLUMN
OF THE N-TH ROW.
DO 5 M=1,NS
NM=ABS(M-9)
IF(NM.EQ.0)CMAT(N,M)=1.+G(2*N-1)/FN
IF(NM.NE.0)CMAT(N,M)=(G(NM)+G(N+M-1))/FN
5 CONTINUE
GO TO 1
2 FACN=PI*COS(PIN)/(8*N-4)
NCHK=1
DO 4 M=1,NS
MM=ABS(M-9)
IF(MM.EQ.0)CMAT(N,M)=FACN*G(2*N-1)
IF(MM.NE.0)CMAT(N,M)=FACN*(G(MM)+G(M+M-1))
4 CONTINUE
1 CONTINUE
CALL SUBROUTINE CDET TO EVALUATE THE DETERMINANT.
CALL CDET(Y1,CMAT,NS)
IF NO ELEMENTS WERE NEAR A POLE, NOW MULTIPLY THE
DETERMINANT BY THE APPROPRIATE FUNCTION TO OBTAIN Y1.
IF(NCHK.LT.1) GO TO 3
RETURN
3 CSRL=CSORT(COMPLEX(ALAM,0.))
CFACT=CCOS(PI*CSRL/2.)
Y1=Y1*REAL(CFACT)
RETURN
END
$ENTRY
$STOP
/*

```

```

$JOB          YIP
              SUBROUTINE YIP(ALAM,GNS,YIP)
C   THIS SUBROUTINE CONSTRUCTS THE ARRAY CMAT
C   CORRESPONDING TO THE DETERMINANT GO.
C   COMPLEX CMPLX,CSORT,CSTH,CCOS
C   COMPLEX CSRL,CFACT
C   DIMENSION CMAT(31,31),G(60)
C   NCHK=0
C   PI=3.141593
C   CONSIDER THE N-TH ROW OF THE DETERMINANT.
C   DO 1 M=1,NS
C     ALAM=M-4.*(N-1)**2
C     ELEMENTS OF THE N-TH ROW HAVE A COMMON POLE.
C     CHECK TO SEE IF THE ELEMENTS ARE NEAR THE POLE.
C     IF SO, MULTIPLY THIS ROW BY THE APPROPRIATE
C     FUNCTION, WHICH WILL HAVE A ZERO HERE.
C     IF ABS(FM).LE.1.DE-04 GO TO 2
C     DEFINE THE ELEMENT IN THE M-TH COLUMN
C     OF THE N-TH ROW.
C   DO 5 M=1,NS
C     MM=IABS(M-M)
C     IF(M.EQ.1.AND.N.EQ.1)CMAT(N,M)=1.
C     IF(M.EQ.1.AND.N.NE.1)CMAT(N,M)=1.414214*G(N-1)/FN
C     IF(M.NE.1.AND.N.EQ.1)CMAT(N,M)=1.414214*G(M-1)/FN
C     IF(M.EQ.N.AND.N.NE.1)CMAT(N,M)=1.+C(2*N-2)/FN
C     IF(M.NE.N.AND.N.NE.1)CMAT(N,M)=(G(M)+G(N+M-2))/FN
C   5 CONTINUE
C   GO TO 1
C 2 FACN=PI*CSIN(N*PT)/4.
C   NCHK=1
C   DO 4 M=1,NS
C     MM=IABS(N-M)
C     IF(M.EQ.1.AND.N.EQ.1)CMAT(N,M)=0.
C     IF(M.EQ.1.AND.N.NE.1)CMAT(N,M)=1.414214*G(N-1)*FACN
C     IF(M.NE.1.AND.N.EQ.1)CMAT(N,M)=1.414214*G(M-1)*(-PI/2.)
C     IF(M.EQ.N.AND.N.NE.1)CMAT(N,M)=FACN*G(2*N-2)
C     IF(M.NE.N.AND.N.NE.1.AND.M.NE.1)CMAT(N,M)=(G(M)+G(N+M-2))*FACN
C   4 CONTINUE
C 1 CONTINUE
C   CALL SUBROUTINE COET TO EVALUATE THE DETERMINANT.
C   CALL COET(YIP,CMAT,NS)
C   IF NO ELEMENTS WERE NEAR A POLE, NOW MULTIPLY THE
C   DETERMINANT BY THE APPROPRIATE FUNCTION TO OBTAIN YIP.
C   IF(NCHK.LT.1) GO TO 3
C   RETURN
C 3 CSRL=CSORT(CMPLX(ALAM,0.))
C   CFACT=-CSRL*CSTH(PT*CSRL/2.)
C   YIP=YIP*REAL(CFACT)
C   RETURN
C   END
$ENTRY
$STOP
/*

```

```

$JOB          HY2
SUBROUTINE HY2(ALAM,G,NS,Y2)
C   THIS SUBROUTINE CONSTRUCTS THE ARRAY SMAT
C   CORRESPONDING TO THE DETERMINANT SO.
COMPLEX CMPLX,CSORT,CSTN,COS
COMPLEX CSRL,CFACT
DIMENSION SMAT(31,31),G(60)
NCHK=0
PI=3.141593
C   CONSIDER THE N-TH ROW OF THE DETERMINANT.
DO 1 N=1,NS
  MM=ALAM-4.*PI*PI
C   ELEMENTS OF THE N-TH ROW HAVE A COMMON POLE.
C   CHECK TO SEE IF THE ELEMENTS ARE NEAR THE POLE.
C   IF SO, MULTIPLY THIS ROW BY THE APPROPRIATE
C   FUNCTION, WHICH WILL HAVE A ZERO HERE.
IF(ABS(G(N)).LE.1.0E-04) GO TO 2
C   DEFINE THE ELEMENT IN THE M-TH COLUMN
C   OF THE N-TH ROW.
DO 5 M=1,NS
  MM=ABS(M-N)
  IF(MM.EQ.0) SMAT(N,M)=1.-G(2*MM)/PI
  IF(MM.NE.0) SMAT(N,M)=(G(MM)-G(N+M))/PI
5 CONTINUE
GO TO 1
2 FACT=PI*COS(PI*N)/(16.*NN)
NCHK=1
DO 4 M=1,NS
  MM=ABS(M-N)
  IF(MM.EQ.0) SMAT(N,M)=-FACT*G(2*MM)
  IF(MM.NE.0) SMAT(N,M)=FACT*(G(MM)-G(M+N))
4 CONTINUE
1 CONTINUE
C   CALL SUBROUTINE COET TO EVALUATE THE DETERMINANT.
CALL COET(Y2,SMAT,NS)
C   IF NO ELEMENTS WERE NEAR A POLE, NOW MULTIPLY THE
C   DETERMINANT BY THE APPROPRIATE FUNCTION TO OBTAIN Y2.
IF(NCHK.LT.1) GO TO 3
RETURN
3 IF(ABS(ALAM).LE.1.0E-04) GO TO 6
CSRL=CSORT(CMPLX(ALAM,0.))
CFACT=CSTN(PI*CSRL/2.)/CSRL
Y2=Y2*RREAL(CFACT)
RETURN
6 Y2=Y2*PI/2.
RETURN
END

$ENTRY
$STOP
/*

```

```

$JNB          HY2P
C   SUBROUTINE HY2P(ALAM,G,NS,Y2P)
C   THIS SUBROUTINE CONSTRUCTS THE ARRAY SMAT
C   CORRESPONDING TO THE DETERMINANT SI.
C   COMPLEX CMPLX,CSORT,CSTN,CCOS
C   COMPLEX CSRL,CFACT
C   DIMENSION SMAT(31,31),G(60)
C   NCHK=0
C   PI=3.141593
C   CONSIDER THE N-TH ROW OF THE DETERMINANT.
C   DO 1 N=1,NS
C   FN=ALAM-(2*N-1)**2
C   ELEMENTS OF THE N-TH ROW HAVE A COMMON POLE.
C   CHECK TO SEE IF THE ELEMENTS ARE NEAR THE POLE.
C   IF SO, MULTIPLY THIS ROW BY THE APPROPRIATE
C   FUNCTION, WHICH WILL HAVE A ZERO HERE.
C   IF(ABS(FN).LE.1.0E-04) GO TO 2
C   DEFINE THE ELEMENT IN THE N-TH COLUMN
C   OF THE N-TH ROW.
C   DO 5 M=1,NS
C   MM=IABS(N-M)
C   IF(MM.EQ.0) SMAT(N,M)=1.-G(2*N-1)/FN
C   IF(MM.NE.0) SMAT(N,M)=(G(MM)-G(N+M-1))/FN
C   5 CONTINUE
C   GO TO 1
C   2 FACN=PI*CCOS(PI*N)/(8*N-4)
C   NCHK=1
C   DO 4 M=1,NS
C   MM=IABS(N-M)
C   IF(MM.EQ.0)SMAT(N,M)=-FACN*G(2*N-1)
C   IF(MM.NE.0)SMAT(N,M)=FACN*(G(MM)-G(N+M-1))
C   4 CONTINUE
C   1 CONTINUE
C   CALL SUBROUTINE CDET TO EVALUATE THE DETERMINANT.
C   CALL CDET(Y2P,SMAT,NS)
C   IF NO ELEMENTS WERE NEAR A POLE, NOW MULTIPLY THE
C   DETERMINANT BY THE APPROPRIATE FUNCTION TO OBTAIN Y2P.
C   IF(NCHK.LT.1) GO TO 3
C   RETURN
C   3 CSRL=CSORT(CMPLX(ALAM,0.))
C   CFACT=CCOS(PI*CSRL/2.)
C   Y2P=Y2P*REAL(CFACT)
C   RETURN
C   END
SENTRY
$STOP
/*

```

FORTRAN IV G LEVEL 21

CDET

DATE = 73135

```
0001      SUBROUTINE CDET(OUT,A,N)
0002      DIMENSION A(31,31)
0003      OUT=1.
0004      L=1
0005      9 CALL CRIG(A,N,IR,JB,L)
0006      IF(IR.EQ.L.AND.JB.EQ.L) GO TO 10
0007      IF(IR.EQ.L.OR.JB.EQ.L) GO TO 11
0008      GO TO 12
0009      11 OUT=-OUT
0010      12 DO 1 J=L,N
0011      B=A(IR,J)
0012      A(IR,J)=A(L,J)
0013      1 A(L,J)=B
0014      DO 2 I=L,N
0015      B=A(I,JB)
0016      A(I,JB)=A(I,L)
0017      2 A(I,L)=B
0018      10 IF(ABS(A(L,L)))20,21,20
0019      21 OUT=0.
0020      RETURN
0021      20 M=L+1
0022      DO 5 I=M,N
0023      DO 5 J=M,N
0024      5 A(I,J)=A(I,J)-A(L,J)*A(I,L)/A(L,L)
0025      OUT=OUT*A(L,L)
0026      IF(N-L-1)8,8,7
0027      7 L=L+1
0028      GO TO 9
0029      8 OUT=OUT*A(N,N)
0030      RETURN
0031      END
```

FORTRAN IV G LEVEL 21

CBIG

DATE = 73135

```
0001      SUBROUTINE CBIG(A,N,IBIG,JBIG,L)
0002      DIMENSION A(31,31)
0003      BIG=A(L,L)
0004      IBIG=L
0005      JBIG=L
0006      DO 5 I=L,N
0007      DO 5 J=L,N
0008      IF(ABS(A(I,J))-ABS(BIG)>5,5,2
0009      2 BIG=A(I,J)
0010      IBIG=I
0011      JBIG=J
0012      5 CONTINUE
0013      RETURN
0014      END
```

FORTRAN IV G LEVEL 21

AIRY

DATE = 73135

```

0001      SUBROUTINE AIRY(Z,AI,AIP,BI,BIP)
0002      IMPLICIT REAL*8(A-H,O-Z)
0003      RREAL*8 ARR(100)
0004      GNU1=.333333333333333
0005      GNU2=.666666666666667
0006      ZETA=GNU2*DABS(Z)**1.5D0
0007      SR3=DSQRT(3.0D0)
0008      SRZ=DSQRT(DABS(Z))
0009      PI=3.141592653589793
0010      IF(Z)1,3,2
0011      2 DZETP=DEXP(ZETA)
0012      CALL BESI(ZETA,GNU1,0,ARR,100)
0013      AI1=ARR(1)*DZETP
0014      CALL BESI(ZETA,GNU2,0,ARR,100)
0015      AI2=ARR(1)*DZETP
0016      CALL BESK(ZETA,GNU1,0,ARR,100)
0017      AK1=ARR(1)/DZETP
0018      CALL BESK(ZETA,GNU2,0,ARR,100)
0019      AK2=ARR(1)/DZETP
0020      AI=SRZ*AK1/(SR3*PI)
0021      AIP=-Z*AK2/(SR3*PI)
0022      BI=2.0D0*SRZ*AI1/SR3+SRZ*AK1/PI
0023      BIP=2.0D0*Z*AI2/SR3+Z*AK2/PI
0024      RETURN
0025      1 CALL BESJ(ZETA,GNU1,0,ARR,100)
0026      AJ1=ARR(1)
0027      CALL BESJ(ZETA,GNU2,0,ARR,100)
0028      AJ2=ARR(1)
0029      CALL BESY(ZETA,GNU1,0,ARR,100)
0030      AY1=ARR(1)
0031      CALL BESY(ZETA,GNU2,0,ARR,100)
0032      AY2=ARR(1)
0033      AI=(SRZ/2.0D0)*(AJ1-AY1/SR3)
0034      AIP=(-Z/2.0D0)*(AJ2+AY2/SR3)
0035      BI=(-SRZ/2.0D0)*(AJ1/SR3+AY1)
0036      BIP=(-Z/2.0D0)*(AJ2/SR3-AY2)
0037      RETURN
0038      3 AI=.355028053887817
0039      BI=AI*SR3
0040      AIP=-.258819403792807
0041      BIP=-AIP*SR3
0042      RETURN
0043      END

```

FORTRAN IV G LEVEL 21

BESI

DATE = 73135

```

0001      SUBROUTINE RESI (X,V,N,A,LDIM1)
0002      IMPLICIT REAL*8 (A-H,O-Z)
0003      REAL*8 X,V,A
0004      DIMEN$ION A(1),GA(8),GB(7)
0005      DATA EPPS/1.0-14/
0006      DATA ACM/Z4000000000000001/
0007      DATA THRSHX/2.0100/
0008      DATA THRSHV/.78D-3/
0009      DATA SD720P /Z41978676F65064F6/
0010      DATA PI   /Z413243F6A8885A31/
0011      DATA S2   /Z411A51A625307D3/
0012      DATA S3   /Z41133RA004F00621/
0013      DATA S4   /Z411151322AC7D848/
0014      DATA S5   /Z411097418ECA7CCE/
0015      DATA EC   /Z4093C467E37DB0C8/
0016      DATA TWOD3 /Z40AAAAAAAAB/
0017      DATA TWOD15 /Z4022222222222222/
0018      DATA GA(1) /Z408E6F3D4EECC38B/
0019      DATA GA(2) /ZC1F7F3D3AE838B52/
0020      DATA GA(3) /ZC13A078933933909/
0021      DATA GA(4) /ZC1B455348B2EBC07/
0022      DATA GA(5) /Z418B39FCBD8ED5E7/
0023      DATA GA(6) /ZC1281211E1F357A2/
0024      DATA GA(7) /Z4022567952E7902C/
0025      DATA GA(8) /ZC1642B8A52A6B47D/
0026      DATA GR(1) /Z41B5FRA40A55B259/
0027      DATA GB(2) /Z422FFC46988A114F/
0028      DATA GB(3) /Z416166FEFAC29673/
0029      DATA GB(4) /Z42698F4E7FREA26F/
0030      DATA GB(5) /ZC139F9E38A8B156C/
0031      DATA GB(6) /Z4167695A73381AC4/
0032      DATA GB(7) /Z407850306D234C9C/
0033      DATA OND3/Z40555555555555555555/
0034      COSH(EXPX)=.500*(FXPX+1./EXPX)
0035      VKL(Y)=DEXP(X*(1.00-COSH(DEXP(Y))))
0036      WR0NSK(AIVP1,AIV,AKV)=(.500*TWODX-AIVP1*AKV)/AIV
0037      LMB=N
0038      ASSIGN 450 TO IK1
0039      KV=1
0040      IF (V.NE.0) KV=2

C      ENTRY FOR RESI CALCULATES (THE I-BESSEL FUNCTION
C      OF X FOR ORDERS V,1+V,2+V,...,N+V)*EXP(-X)
C      R(K)=I(K+1,X)/I(K,X) IS STORED IN A(K+2)
C      TEMPORARILY. THIS ELIMINATES OVERFLOW PROBLEMS

0041      100 CONTINUE
0042      IF (X.LE.0.00.OR.V.LT.0.00.OR.V.GE.1.00.OR.N.LT.0)
0043          X      GO TO 995
0044          TWODX=2.000/X
0045          FK=0.00
0046          FKPI=1.00
0047          K=LMB
0048          TEMP=-(DFLOAT(LMB)+V)
0049          TEMP=TEMP*TWODX
0050
0051      120 K=K+1
          FKM1=FK
          FK=FKPI

```

FORTRAN IV G LEVEL 21

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```

0052      TEMP=TEMP-TWO0X
0053      FKPI=TEMP*FK+FKM1
0054      IF ((FKPI+1.0D0).NE.FKPI) GO TO 120
C
C      THE POINT OF RECURSION K IS NOW WELL DETERMINED.
C
0055      IF ((K+2).GT.LDIMA) GO TO 995
0056      KP=K+2
0057      A(KP)=0.0D0
0058      DR=DFLOAT(K)+V
0059      DR=DR+DR
0060      DO 150 M=1,K
0061      KP=KP-1
0062      A(KP)=X/(DR+X*A(KP+1))
0063      DR=DR-2.0D0
0064      150 CONTINUE
C
C      NORMALIZE I'S BY SUMMATION FORMULA
C
0065      BGAM=1.0D0
0066      Q2DXPV=1.0D0
0067      IF (KV.EQ.1) GO TO 200
0068      Q2DXPV=(TWO0X)**V
0069      TEMP=V+2.0D0
0070      BGAM=GA(1)+GB(1)/(TEMP+
2      GA(2)+GR(2)/(TEMP+
3      GA(3)+GB(3)/(TEMP+
4      GA(4)+GR(4)/(TEMP+
5      GA(5)+GR(5)/(TEMP+
6      GA(6)+GR(6)/(TEMP+
7      GA(7)+GR(7)/(TEMP+GA(8)))))))
0071      BGAM=BGAM/(1.0D0+V)
0072      200 CONTINUE
0073      A(1)=1.0D0
0074      TEMP=Q2DXPV*BGAM
0075      SIG=TEMP
0076      EN1=V+1.0D0
0077      TEMP=TEMP*(EN1+EN1)
0078      EN2=V+V
0079      D1=1.0D0
0080      D2=V
0081      DO 250 M=1,K
0082      A(M+1)=A(M)*A(M+1)
0083      SIG=SIG+TEMP*A(M+1)
0084      EN1=EN1+1.0D0
0085      EN2=EN2+1.0D0
0086      D1=D1+1.0D0
0087      D2=D2+1.0D0
0088      TEMP=TEMP*(EN1/D1)*(EN2/D2)
0089      250 CONTINUE
0090      A(1)=1.0D0/SIG
0091      DO 300 M=1,K
0092      A(M+1)=A(M+1)*A(1)
0093      300 CONTINUE
0094      GO TO 1K1,(450,1100)
0095      450 RETURN
0096      ENTRY BESK(X,V,N,A,LDIMA)
0097      LM8=1

```

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C ENTRY BESK CALCULATES EXP(X)*THE K-BESSEL
FUNCTIONS FOR ORDERS V,V+1,V+2,...,V+N.

```

0098      ASSIGN 1100 TO IK1
0099      VX=V
0100      KV=1
0101      IF (V.EQ.0.0D0) GO TO 100
0102      IF (V.LT.THRSHV) GO TO 1020
0103      TEMP=V-1.0D0
0104      IF ((TEMP+THRSHV).GT.0.0D0) GO TO 1030
0105      KV=3
0106      GO TO 100
0107      1020 CONTINUE
0108      KV=2
0109      GO TO 100
0110      1030 CONTINUE
0111      VX=TEMP
0112      KV=4
0113      GO TO 100
0114      1100 CONTINUE
0115      IF (X.LT.THRSHX) GO TO 1500

```

C COMPUTE K(V,X) BY INTEGRATION

C

```

0116      H=DLOG((X+38.7)/(X*.5D0-1.0D0))
0117      1110 CGTINUE
0118      VKM=VKL(H)
0119      IF (KV.NE.1) VKM=VKM*COSH(DEXP(V*H))
0120      IF (VKM.GT.ACML) GO TO 1150

```

```

0121      H=.5D0*H
0122      GO TO 1110
0123      1150 CONTINUE

```

```

C124      LMR=2
0125      SIG=.5D0+VKM
0126      S=SIG*H

```

```

0127      1190 CONTINUE
0128      HP=.5D0*H
0129      XP=-HP
0130      SIG1=0.0D0
0131      DO 1200 M=1,LMB
0132      XP=XP+H

```

```

0133      VKM=VKL(XP)
0134      IF (KV.NE.1) VKM=VKM*COSH(DEXP(V*XP))
0135      SIG1=SIG1+VKM

```

```

0136      1200 CONTINUE
0137      SIG=SIG+SIG1
0138      SP=S

```

```

0139      S=SIG*HP
0140      IF (DABS(SP-S).LE.(EPPS*S)) GO TO 1497
0141      H=HP

```

```

0142      LMB=LMB+LMB
0143      GO TO 1190
0144      1497 CONTINUE

```

```

0145      GO TO 1800
0146      1500 CONTINUE
0147      EXPOX=DEXP(X)

```

C COMPUTE K(VX,X) BY SUMMATION OF I'S

0148 LMR=3

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```

0149      IF(KV.E0.3) GO TO 1525
0150      AA=BG-DBDX(TWODX)
0151      1510 GO TO (1520,1540,1510,1530),KV
0152      1520 CONTINUE
0153      DZ=-AA
0154      GO TO 1550
0155      1525 CONTINUE
0156      DZ=((BGAM*Q2DXPV)**2-PI**V/DSIN(PI*V))*5/V
0157      GO TO 1550
0158      1530 CONTINUE
C
C      FUNCTIONS OF V MUST BE TRANSFORMED TO FUNCTIONS
C      OF VX=V-1.
C
0159      BGAM=BGAM/V
0160      Q2DXPV=Q2DXPV/TWODX
0161      LMB=2
0162      1540 CONTINUE
C
C      COMPUTE DZ BY SERIES APPROXIMATION
C
0163      AA2=AA*AA
0164      AA3=AA2*AA
0165      AA4=AA3*AA
0166      AA5=AA4*AA
0167      S=-VX
0168      DZ=(TWOD15*AA5+S5+TWOD3*(AA3*S2+AA2*S3)+.5D0*AA *
X      (S4+S2**2)+OND3*S2*S3)*S+
X      .25D0*(S4+S2**2)+TWOD3*AA*S3+AA2*S2+
X      OND3*AA4-SD723P
0169      DZ=((DZ*S+(OND3*S3+AA*S2+TWOD3*AA3))*S+
X      AA2)*S+AA
0170      DZ=-DZ
0171      IF (KV.E0.2) GO TO 1550
0172      SIG1=A(1)+TWODX*V*A(1)
0173      SIG=DZ*SIG1*EXPDX
0174      GO TO 1560
0175      1550 CONTINUE
0176      SIG=DZ*A(1)*EXPDX
0177      LMR=3
0178      1560 CONTINUE
C
C      COMPUTE D1
C
0179      TEMP=((BGAM*Q2DXPV)**2*(VX+2.0D0)/(1.0D0-VX))*EXPDX
0180      EN1=VX+2.0D0
0181      EN2=VX-VX
0182      EN3=V X
0183      D2=EN3
0184      D3=1.0D0-D2
0185      D1=1.0D0
0186      GO 1570 M=LMB,K,2
0187      SIG=SIGHA(M)*TEMP
0188      EN1=EN1+2.0D0
0189      EN2=EN2+1.0D0
0190      EN3=EN3+1.0D0
0191      D1=D1+1.0D0
0192      D2=D2+2.0D0

```

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```
0193      D3=D3+1.0D0
0194      TEMP=TEMP*(EN1/01)*(EN2/02)*(EN3/D3)
0195      1570 CONTINUE
C
C      SIG CONTAINS K(VX,X)
C
0196      S=EXPDX*SIG
0197      IF (KV.NE.4) GO TO 1800
C
C      USE WRONSKIAN TO GET K(V,X)*EXP(X)
C
0198      S=WRONSK(A(1),SIC1,S)
0199      1800 CONTINUE
C
C      NOW USE WRONSKIAN TO GET K(V+1,X)*EXP(X)
C      S CONTAINS K(V,X)*EXPX
SP=WRONSK(A(2),A(1),S)
A(2)=SP
A(1)=S
0203      DR=TWODX*V
0204      DO 1900 M=2,N
0205      DR=DR+TWODX
0206      A(M+1)=DR*A(M)+A(M-1)
0207      1900 CONTINUE
0208      RETURN
0209      995 PRINT 1,X,V,N,K,LDIM
0210      1 FORMAT(10ERROR IN BES I/K X=*,E14.5,1 V=*,E14.5,* N=*,I5,* K=*,I5,
           X* LDIMA=*,I5)
0211      RETURN
0212      END
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0001      SUBROUTINE BESJ(X,V,N,A,LDIMA)
0002      IMPLICIT REAL*8 (A-H,D-Z)
0003      REAL*8 X,A,V
C      ENTRY RESJ OF SUBROUTINE CALCULATES THE J-BESSEL
C      FUNCTION OF X FOR DROPS V,1+V,2+V,...,N+V
C      R(K) IS STORED IN A(K+2) (TEMPORARILY)
C      J(V+K) IS STORED INTO A(K+1)
C      X>0
C      0<=V<1.00
C      N>=0
C      DIMENSION OF ARRAY A MUST BE AT LEAST MAX(X,N)+16
C      LDIMA IS THE DIMENSION OF A SUPPLIED BY THE
C      USER.
C      **** ERROR RETURNS ****
C      FOR X, V, OR N OUT OF RANGE OR IF DIMENSION
C      OF ARRAY FURNISHED IS TOO SMALL.
C      ERROR RETURN INFORMATION INCLUDES
C      X, V, N, MU, AND LDIMA WHERE MU IS THE
C      SIZE-2 OF THE ARRAY A NEEDED (MU IS
C      MEANINGLESS IF ANOTHER PARAMETER IS
C      OUT OF RANGE, E.G., X=-5.)
C
0004      DIMENSION A(1)
0005      DIMENSION GA(8),GR(7)
0006      DATA PI/3.1415926535897932D0/
0007      DATA GA/5.5636964340807375D-1,-1.5497028047270494D1,
X-3.6268398299612859D0,-1.1271046203309880D1,
X8.7016417889220691D0,-2.6919115854447927D0,
X1.3413198732411108D-1,-6.2606299618306871D0/
0008      DATA GB/1.1373935737962844D1,4.7985452207286753D1,
X6.087645111805654D0,1.0555975154972231D2,
X-3.6235080158088192D0,6.4632210255382896D0,
X4.8169233955859386D-1 /
0009      DATA PI2/1.5707963267948966D0/
0010      EQUIVALENCE (GD,GOLRGV,GOSMLV)
0011      DATA THRSHV/.789D-3/
0012      DATA EC /74093C467E77D80C8/
0013      DATA S2/1.6449340668482264/,_
1   S2D2SQ/Z4JD2BF5ABCE72C3/,_
2   S5/1.036927755143370D0/,_
3   S33/.4006856343865314/,_
4   S4/1.0823232337111381D0/
0014      DATA PI4D72/Z4115A57EB579CE55/
0015      DATA THSJ/.12D0/
0016      DIMENSION GOVSMD(4)
0017      ASSIGN 510 TO JY1
0018      ASSIGN 110 TO JY4
0019      80 CONTINUE
0020      IF(X.LE.0.OR.V.LT.0.OR.V.GE.1.0.OR.N.LT.0) GO TO 995
C      SET OVERFLOW INDICATOR OFF
C
C      87 CALL OVERFL(MU)
C      VP1=V+1.0D0
0021

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0022      IX=IFIX(SNGL(X))
0023      LMR=IX+1
0024      MU=LMR
0025      GO TO JY4,(110,120)
0026      110 CONTINUE
0027      IF(LMR.LT.N) MU=N
0028      120 CONTINUE

C      LMB IS THE VALUE OF I FOR WHICH WE ARE ASSURED
C      THAT J(V+I,X) IS ON THE TAIL OF THE FUNCTION,
C      I.E., LMB=IFIX(X+1)
C      LET LL=MAX(LMR,N)...
C      DEFINE MU TO BE THE POINT FROM WHICH WE
C      MUST RECUR TO ASSURE THAT J(V+LL,X)
C      IS ACCURATELY DETERMINED
C

C      FOR I=LL,LL+1,...,MU
C      WE STORE R(I)=J(I+1,X)/J(I,X) INTO A(I+2).
C      THIS AVOIDS THE PROBLEM OF OVERFLOW
C      INHERENT IN THE GROWTH OF THE J-FUNCTION
C      WHEN RECURRING BACKWARD ON ITS TAIL
C      NOTE THAT R(I-1)=X/(2*(V+I))-X*R(I)
0029      TWODX=2.0D0/X
0030      DR=TWODX*(V+DFLOAT(MU))
0031      FKPI=1.0D0
0032      FK=0.0D0
0033      180 CONTINUE

C      ITERATE UNTIL FKPI IS GREATER THAN REGISTER ACCURACY
C

0034      MU=1+MU
0035      DR=DR+TWODX
0036      FKMI=FK
0037      FK=FKPI
0038      FKPI=DR*FK-FKMI
0039      IF ((FKPI+1.0D0).NE.FKPI) GO TO 180

C      THE VALUE OF MU IS NOW WELL DETERMINED
C

0040      MU=MU+1
0041      M=MU
0042      IF((M+2).GT.LDIMA) GO TO 995
0043      A(M+2)=0.0D0
0044      200 CONTINUE
0045      IF (M.EQ.LMB) GO TO 250
0046      DR=2*(M+V)
0047      M=M-1
0048      A(M+2)=X/(DR-X*A(M+3))
0049      GO TO 200

C      STORE JBAR(I) INTO A(I+1) FOR I=0,1,2,...,LMB+1
C

0050      250 CONTINUE
0051      A(M+1)=1.0DC/A(M+2)
0052      A(M+2)=1.0D0
C      RECUR BACKWARD TO GET JBAR'S
C

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0053      280 CONTINUE
0054      DR=2*(M+V)/X
0055      A(M)=DR*A(M+1)-A(M+2)
0056      M=M-1
0057      IF (M.GT.0) GO TO 280
C
C      CLEAR R'S  UNDERFLOW MAY OCCUR HERE
C
0058      LMB=LMB+1
0059      DO 290 M=LMB,MU
0060      A(M+2)=A(M+2)*A(M+1)
C
C      287 CALL OVERFL(I)
C      288 IF (I.EQ.3) A(M+2)=0.0D
0061      290 CONTINUE
C
C      NORMALIZE SEQUENCE OF JBARS BY SUMMATION
C
0062      IF (V.EQ.0.0D0) GO TO 305
0063      VX=V+2.0D0
0064      BGAM=GA(1)+GB(1)/(VX+
X      GA(2)+GB(2)/(VX+
X      GA(3)+GR(3)/(VX+
X      GA(4)+GR(4)/(VX+
X      GA(5)+GR(5)/(VX+
X      GA(6)+GR(6)/(VX+
X      GA(7)+GR(7)/(VX+GA(8))))))
0065      RGAM=BGAM/VP1
0066      BGAM0=BGAM*RGAM
0067      Q2DXPV=(2.0D0/X)**V
0068      305 CONTINUE
C
C      SUMMATION
0069      ALPHA=A(1)
0070      PHI=2.0D0
0071      IF (V.EQ.0.0D0) GO TO 320
0072      D1=1.0D0
0073      D2=V
0074      EN2=V
0075      EN1=V+2.0D0
0076      PHI=Q2DXPV*RGAM
0077      ALPHA=PHI*ALPHA
0078      320 CONTINUE
0079      DO 350 M=1, MU,2
0080      IF (V.EQ.0.0D0) GO TO 330
0081      PHI=PHI*(EN2/D2)*(EN1/D1)
0082      D2=D2+2.0D0
0083      EN1=EN1+2.0D0
0084      D1=1.0D0+D1
0085      EN2=EN2+1.0D0
0086      330 CONTINUE
0087      ALPHA=ALPHA+PHI*A(M+2)
0088      350 CONTINUE
C      A(1),A(2),...,A(LMB+2) CONTAIN JBAR(0),JBAR(1),...,JBAR(LMB+1)
C      A(LMB+3),...,A(MU+2) CONTAIN R(LMB+1),...,R(MU)
C
C      NORMALIZE JBARS
C

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0089      M=MU+2
0090      DO 460 I=1,M
0091      457 A(I)=A(I)/ALPHA
0092      460 CONTINUE
0093      500 CONTINUE
0094      GO TO JY1,(510,1100)
0095      510 CONTINUE
0096      RETURN
0097      995 PRINT 1,X,V,N,MU,LDI MA
0098      RETURN
0099      1 FORMAT('0ERROR IN BESL X=',E14.5,' V=',E14.5,' N=',I,15,' MU=',I,15,
X' LDI MA=',I5)
0100      RETURN
0101      ENTRY BESY(X,V,N,A,LDI MA)
0102      ASSIGN 1100 TO JY1
0103      ASSIGN 120 TO JY4
0104      GO TO 80
0105      1100 CONTINUE
0106      VX=V
0107      TJ=DABS(A(1))-THSJ
C
C      COMPUTE GO AND G1
0108      IF (V.LT.THRSHV) GO TO 1200
0109      VX=V-1.0D0
0110      IF ((VX+THRSHV).GT.0.0D0) GO TO 1190
0111      VX=V
0112      GOLRGV=DCOTAN(V*PI)-(Q2DXPV**2/PI)*BGAMSQ/V
0113      GO TO 1220
0114      1190 CONTINUE
C
C      FUNCTIONS OF V USED IN COMPUTING GO AND G1
C      MUST BE TRANSFORMED TO FUNCTIONS OF VX=1-V
C
0115      Q2DXPV=Q2DXPV/TWODX
0116      BGAMSQ=RGAMSQ/V**2
0117      1200 CONTINUE
C
C      COMPUTE GO USING EXPANSION
0118      Z=EC+DLOG(X/2.0D0)
0119      GO= Z/PI2
0120      IF (V.NE.0.0D0) GO TO 1210
0121      G1=2.0D0/PI2
0122      BGAMSQ=1.0D0
0123      Q2DXPV=BGAMSQ
0124      GO TO 1230
0125      1210 CONTINUE
0126      ASQ=Z**Z
0127      ACUR=Z*ASQ
0128      AFORTH=Z*ACUR
0129      AFIFTH=AFORTH*Z/15.0D0
0130      G1=-VX
0131      GO SMLV=
X      (((((2.0D0*(AFIFTH+ACUR*S2/3.0D0+ASQ*S33+Z*S2D250)
X      +Z*S4/2.0D0+S2*S33+S5)*G1+
X      2.0D0*Z*S33+S2D250+ASQ*S2
X      +AFORTH/3.0D0+PI4D72)*G1+
X      S33+Z*S2+ACUR/1.5D0)*G1+
X      1.5D0*S2+ASQ)))*G1+Z)/PI2

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0132      1220 CONTINUE
          C COMPUTE G1
          G1=(Q2DXPV**2/PI2)*BGAMSQ*(2.000+VX)/(1.000-VX)
0133      1230 CONTINUE
          C COMPUTE Y0 FROM SUM(J**5) FORM
          C
0135      EN3=VX+1.000
0136      EN2=VX+EN3
0137      EN1=VX+4.000
0138      D1=2.000
0139      D2=D1-VX
0140      D3=D1+VX
0141      TJE=0
0142      IF(TJ.GE.0.D0.AND.VX.GE.0.D0) GO TO 1232
0143      TJE=1
          C
          C Y(VX+1,X) MUST ALSO BE COMPUTED BY A SUM
          C
0144      THVDX=3.000*VX/X
0145      PSIZ=-BGAMSQ*Q2DXPV**2/(PI2*X)
0146      PSI1=G0-.500*G1
0147      1232 CONTINUE
          IF (VX.LT.1.00) GO TO 1233
0148      M=3
0149      YV=G0*A(1)
0150      IF (TJ.GE.0) GO TO 1238
0151      YVP1=PSIZ*A(1)+PSI1*A(2)
0152      GO TO 1238
0153
0154      1233 CONTINUE
          Z=TWODX*V*A(1)-A(2)
          YV=G0*Z
0155      M=2
0156      YVP1=PSIZ*Z+PSI1*A(1)
0157
0158      1236 CONTINUE
          DO 1250 I=M,MU,2
          YV=G1*A(I)+YY
0159      G=G1
          G1=-G1*(EN1/D1)*(EN2/D2)*(EN3/D3)
0160      EN1=EN1+2.000
0161      EN2=EN2+1.000
0162      EN3=EN3+1.000
0163
0164      D1=D1+1.000
0165      D2=1.000+D2
0166      D3=D3+2.000
0167
0168      IF (TJE) 1240,1250,1240
0169
0170      1240 CONTINUE
          YVP1=YVP1+THVDX*G*A(1)+.5*(G-G1)*A(1+1)
0171
0172      1250 CONTINUE
          IF (VX.GE.0.001) GO TO 1260
0173
0174      Z=YVP1
0175      YVP1=V*Z*TWODX-YY
0176      YY=Z
0177
0178      GO TO 1400
0179      1260 CONTINUE
          IF (TJ.LT.0.D0) GO TO 1400
0180
0181      1270 CONTINUE
          C
          C NOW COMPUTE Y(V+1)

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C WRONSKIAN PROVIDED NOT NEAR A ZERO OF J
C
0182 YVP1=(YY*A(2)-1.0D0/(X*PI2))/A(1)
0183 1400 CONTINUE
C
C RECUR FORWARD TO GET Y'S (WISE?)
C
0184 A(1)=YY
0185 A(2)=YVP1
0186 G=V*TWODX
0187 DO 1500 I=2,N
C
0188 G=G+TWODX
C OVERFLOW MAY OCCUR HERE
0189 A(I+1)=G*A(I)-A(I-1)
C1487 CALL OVERFL(LMB)
C1488 IF (LMR.F0.1) A(I+1)=1.078
0190 1500 CONTINUE
0191 RETURN
0192 END

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A STUDY OF THE HILL-FUNCTION SOLUTION
TO PROBLEMS OF PROPAGATION IN STRATIFIED MEDIA

by

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ABSTRACT

The problem of determining the reflection and transmission coefficients for a plane wave incident upon an inhomogeneous, stratified region of finite thickness is solved in terms of Hill functions. The effect that parameters of the problem have upon cost and accuracy of the solution is discussed from both analytical considerations and numerical results obtained from solving the problem for a region with a linear dielectric profile. Expressions for the reflection coefficient at low frequencies are obtained using the Hill-function formulation, as well as the familiar Airy-function formulation.